

Project: IEEE P802.15 Working Group for Wireless Personal Area Networks (WPANs)

Submission Title: High-speed VLC modulation formats and MIMO transmission

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Abstract: In response to «Call for Proposals for OWC Channel Models» issued by 802.15.7r1, this contribution presents the PHY technologies proposal of high-speed VLC modulation formats and MIMO transmission.

Purpose: Call for Proposal Response

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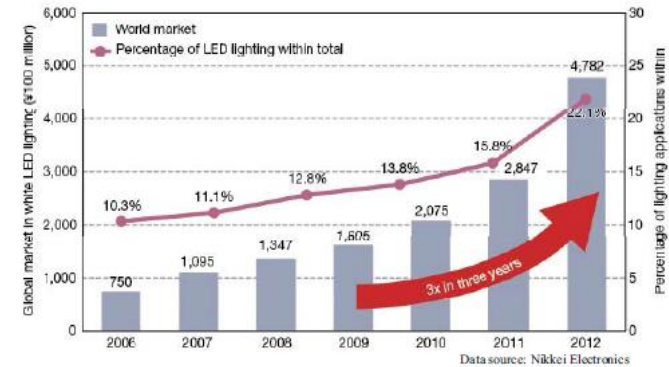
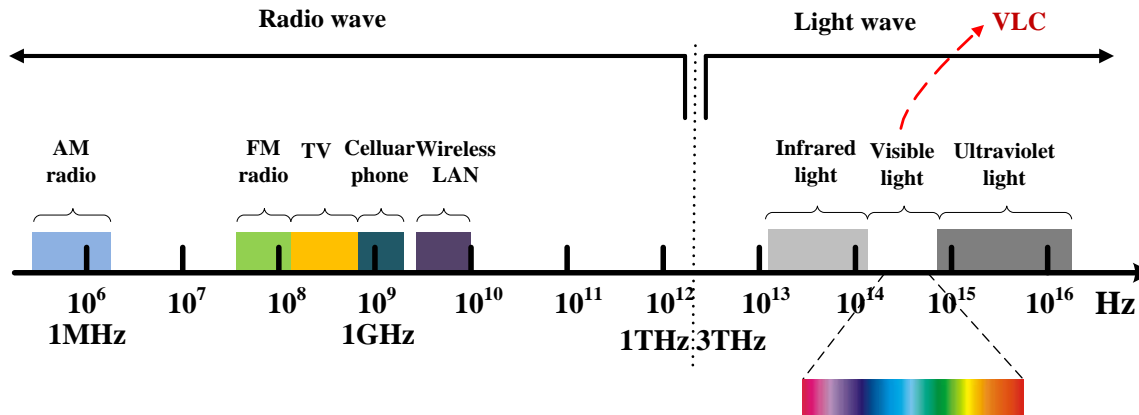
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High-speed VLC modulation formats and MIMO transmission

Outlines

- Background and Introduction
- Scenario Targets
- Description of Proposed Solutions
- Some Experiment Results
- Conclusions

Background



LED lighting in the global market

- ❑ Expand the spectrum for the next generation of broadband communications
- ❑ Combine lighting with communication, bringing unique advantages

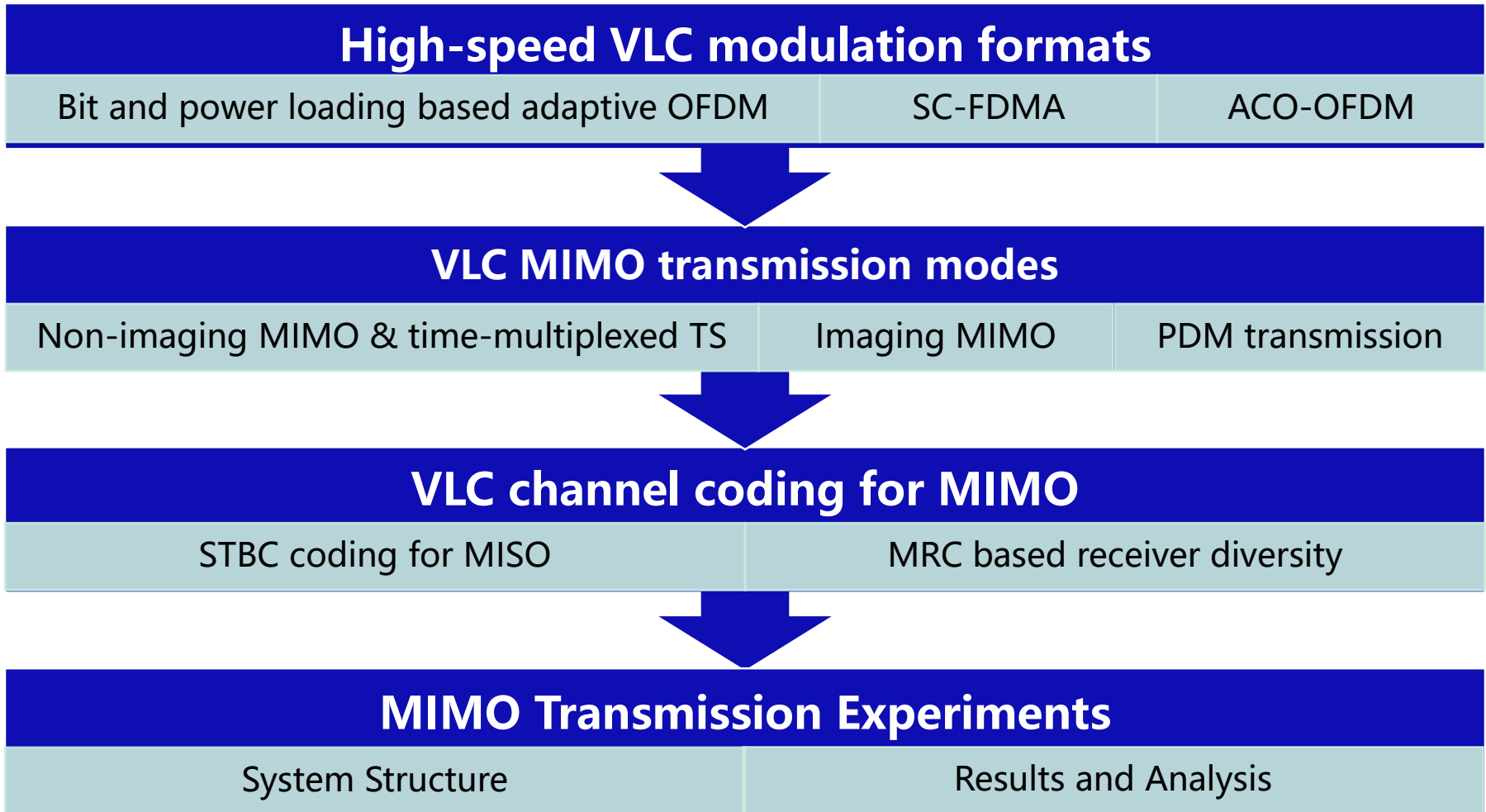
By 2018, the semiconductor lighting penetration will reach 80%. With the popularity of LED lighting process, VLC will be standing on the shoulders of giants.

Targets

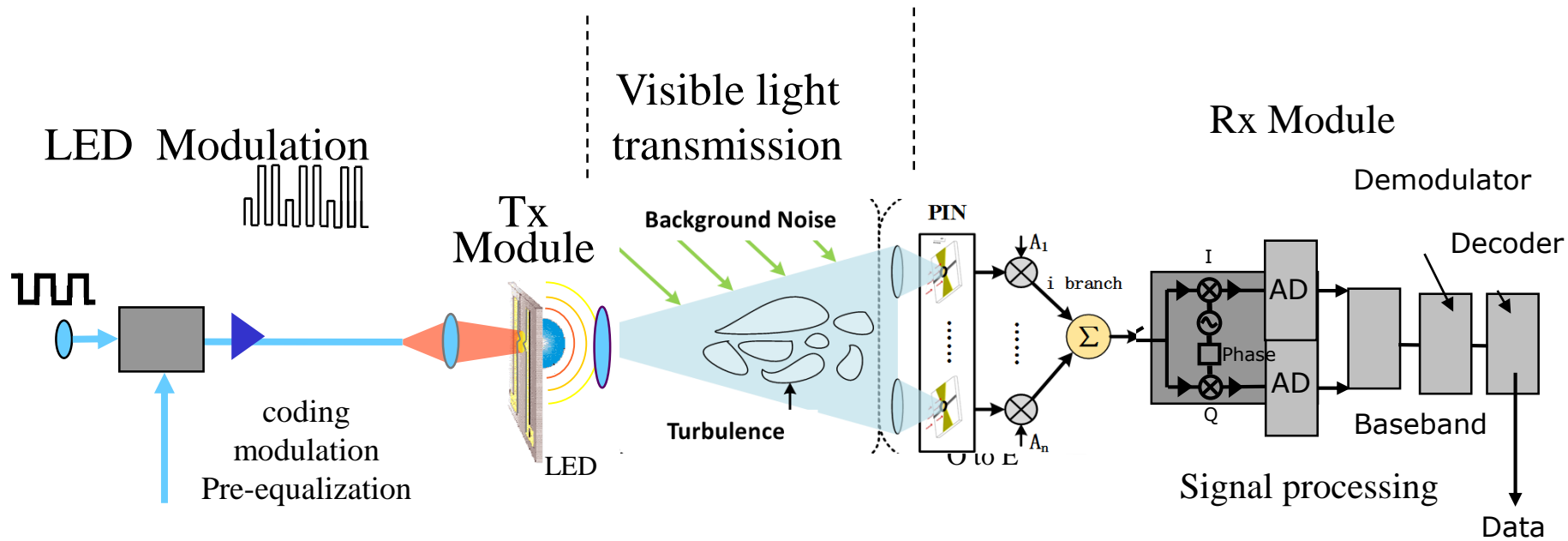
- High-speed VLC MIMO transmission
- Data Rates Speed: **~Gbps**
- Distance: **3m~5m**
- Environment: **indoor public space for multiple users**
- Link: **non-imaging MIMO, imaging MIMO**

To provide a high-speed VLC free-space link for multi-user access.

Proposal Research Route



Physical Layer of VLC system



TX :

- electronics : LED driving circuit , **coding, modulation**, Pre-equalization
- optics : transmitter antenna (**Multiple Input**)

RX :

- optics : receiver antenna (**Multiple Output**)
- electronics : signal processing (decoding, demodulation, equalization),

To achieve the high speed VLC

□ Modulation formats:

- Bit and power loading based adaptive OFDM
- Single-carrier FDMA (optional)
- Asymmetrically-clipped optical OFDM (optional)

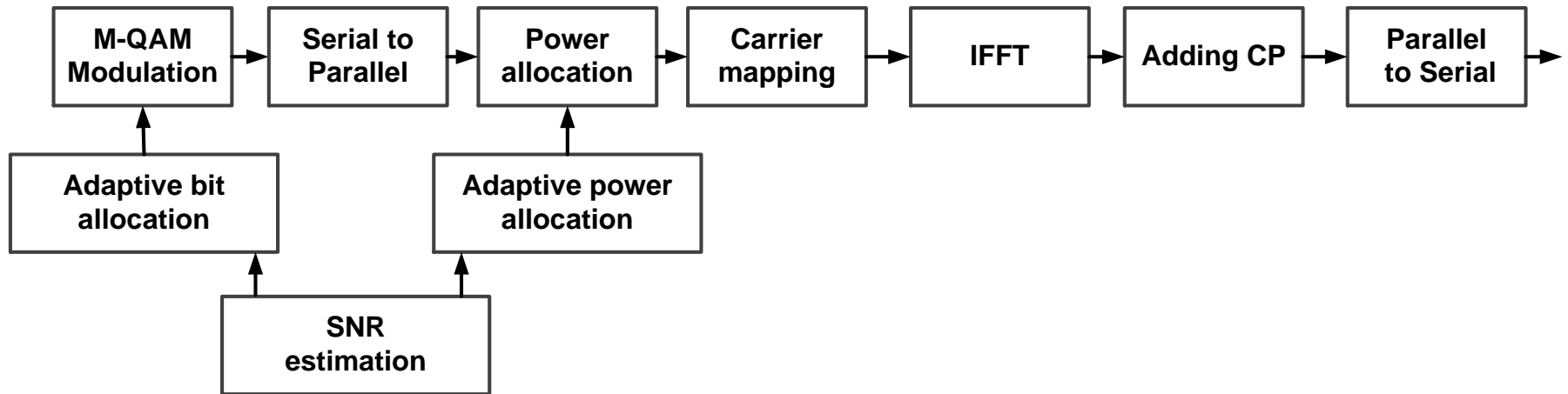
□ MIMO transmission modes

- Non-imaging MIMO with time-multiplexed training symbols
- Imaging MIMO
- Polarization division multiplexing (PDM) transmission

□ Channel coding for MIMO

- Self-adaptive space-time block coding (STBC) for MISO
- Maximal ratio combining (MRC) based receiver diversity

Bit and power loading based adaptive OFDM modulation



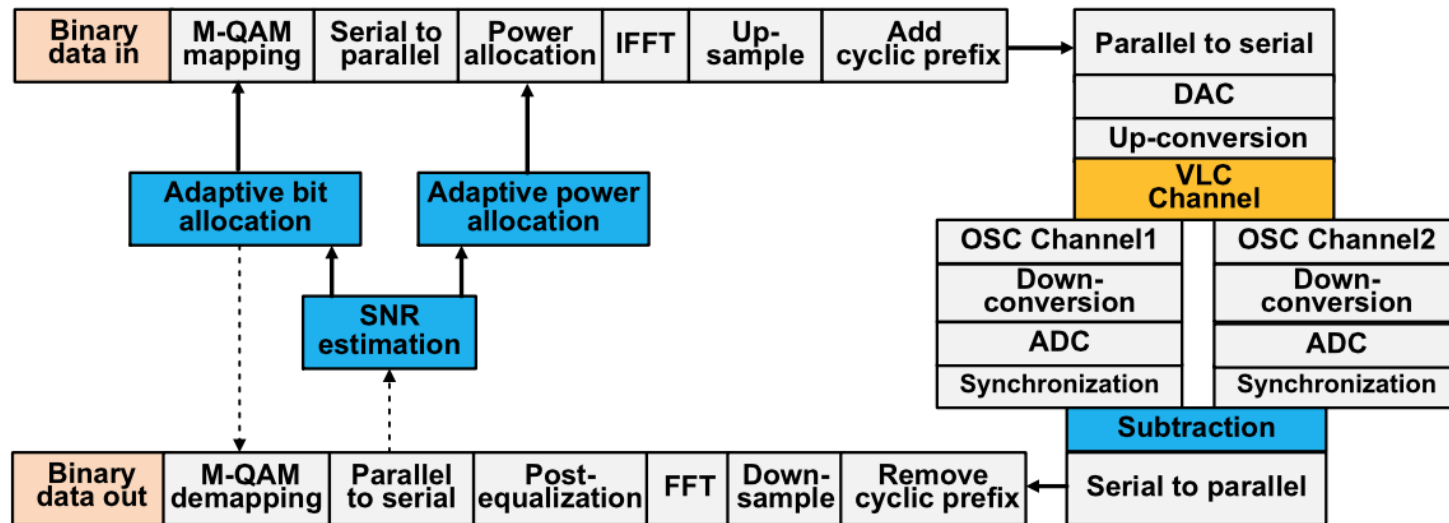
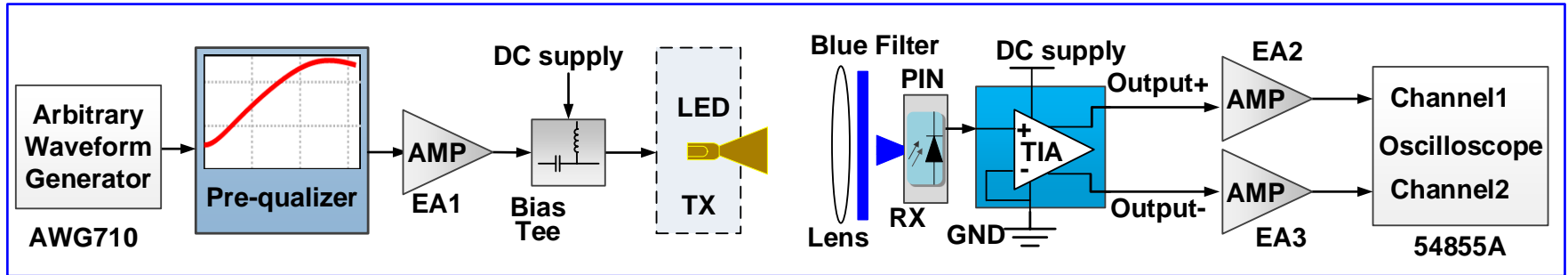
Bit and power loading is used in adaptive OFDM signal generation to maximum system capacity

$$R = \frac{B(\sum_{k=1}^N \log_2 M_k)}{N}$$

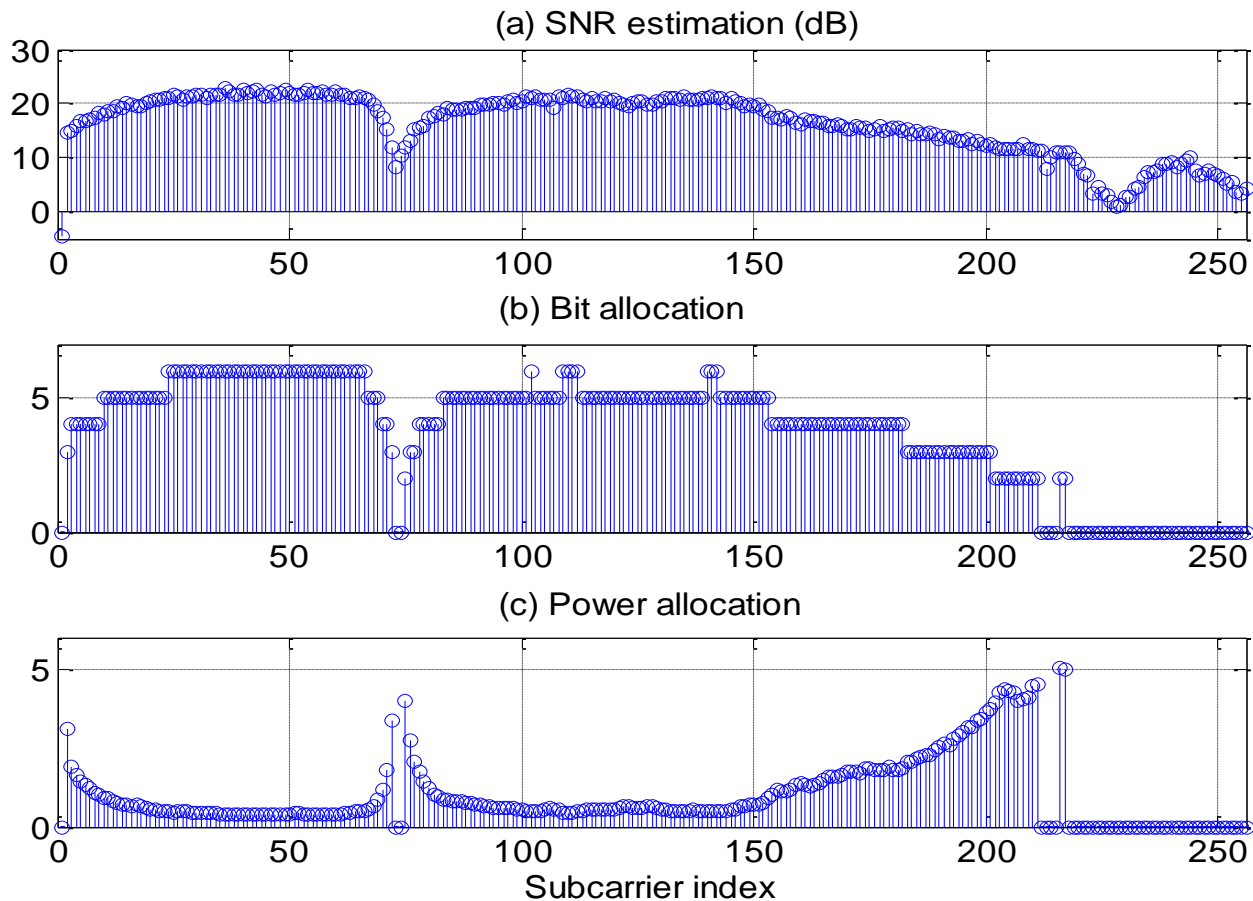
Before applying bit and power allocation, the SNR of the channel is estimated through EVM method

$$EVM = \frac{\sum_{n=1}^M |S_n - S_{0,n}|^2}{\sum_{n=1}^M |S_{0,n}|^2} \quad SNR = \frac{1}{EVM^2}$$

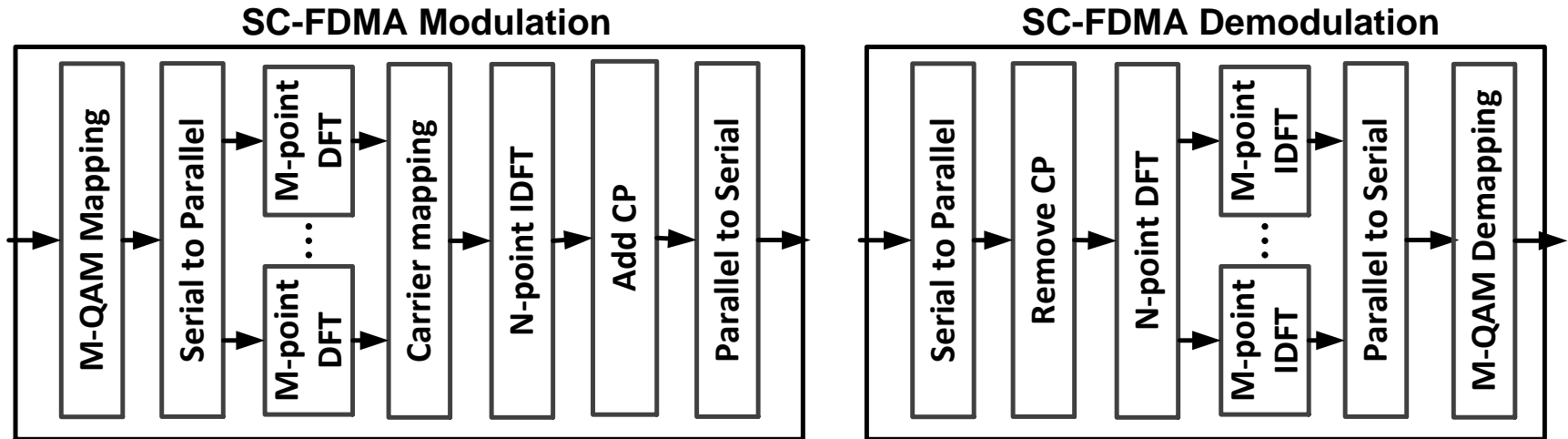
Bit and power loading based adaptive OFDM modulation



Bit and power loading based adaptive OFDM modulation

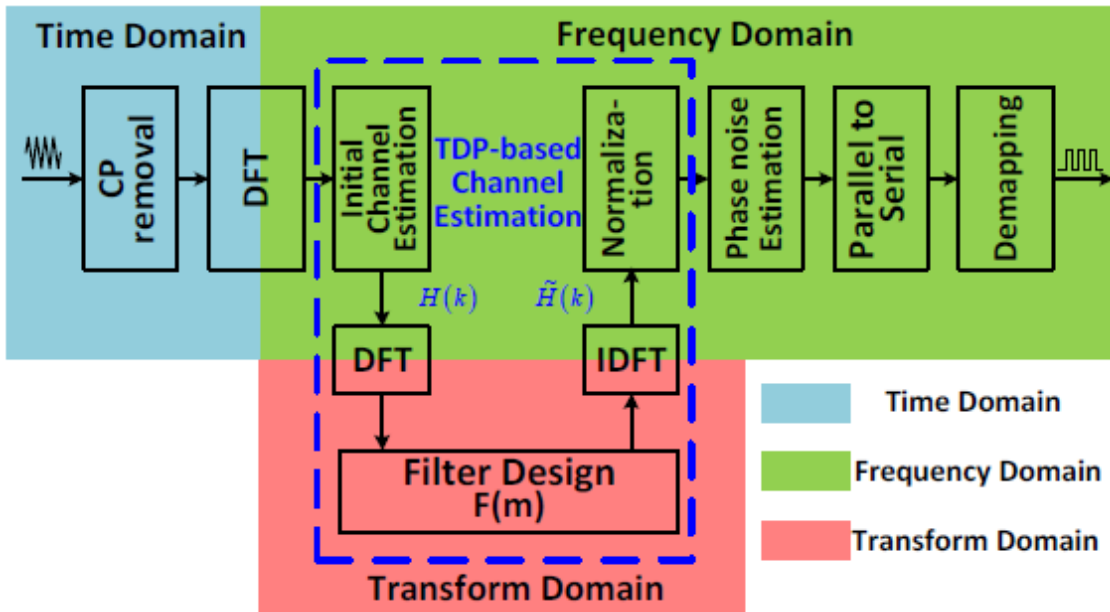


Single carrier modulation (optional): Single-carrier frequency division multiple access (SC-FDMA)



- SC-FDMA can be an optional choice for uplink, due to its advantage to provide low PAPR for the transmit waveform;
- SC-FDMA divides the transmission bandwidth into multiple parallel subcarriers, which adds a DFT block before the subcarrier mapping

Transform domain processing (TDP) based channel estimation method for SC-FDMA



Transform domain processing (TDP) can be used for channel estimation:

- improve the accuracy of channel estimation
- reduce the length of overhead

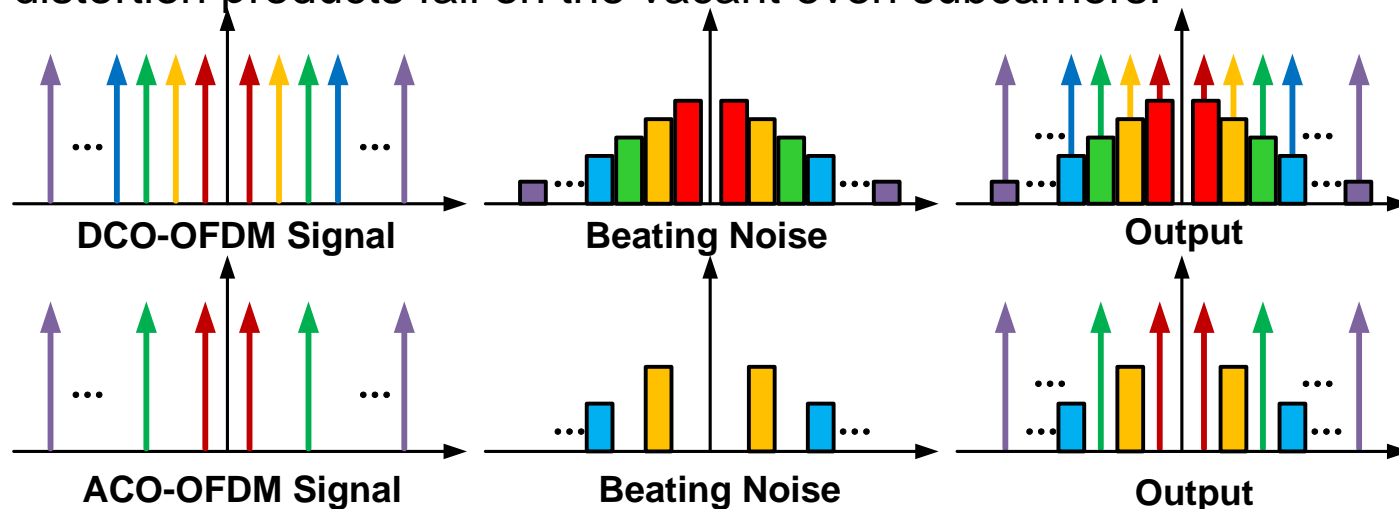
$H(k)$ is transformed to $H_T(m)$ by employing another DFT. $H_T(m)$ is in transform domain and expressed as

$$H_T(m) = \sum_{k=0}^{N-1} (H(k)) \cdot \exp(-j2\pi mk / N)$$

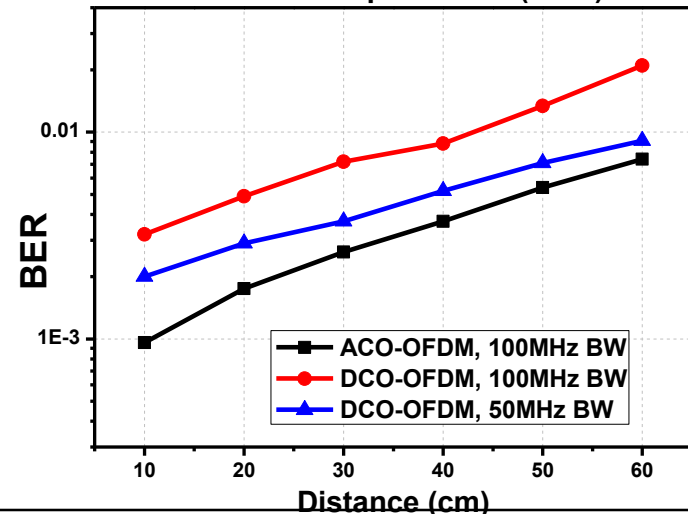
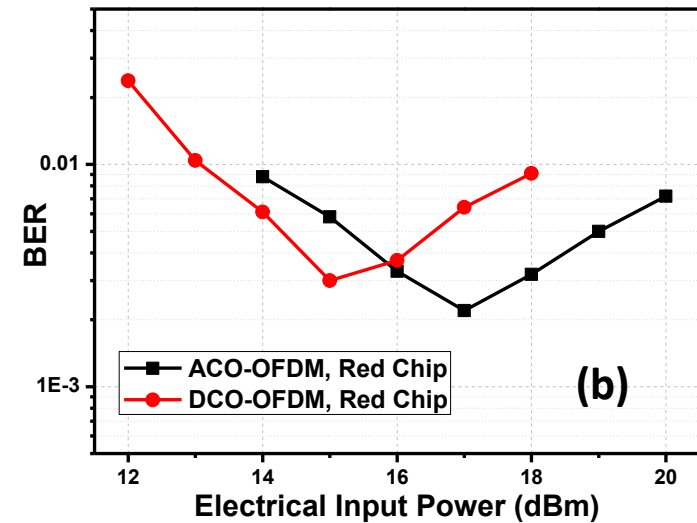
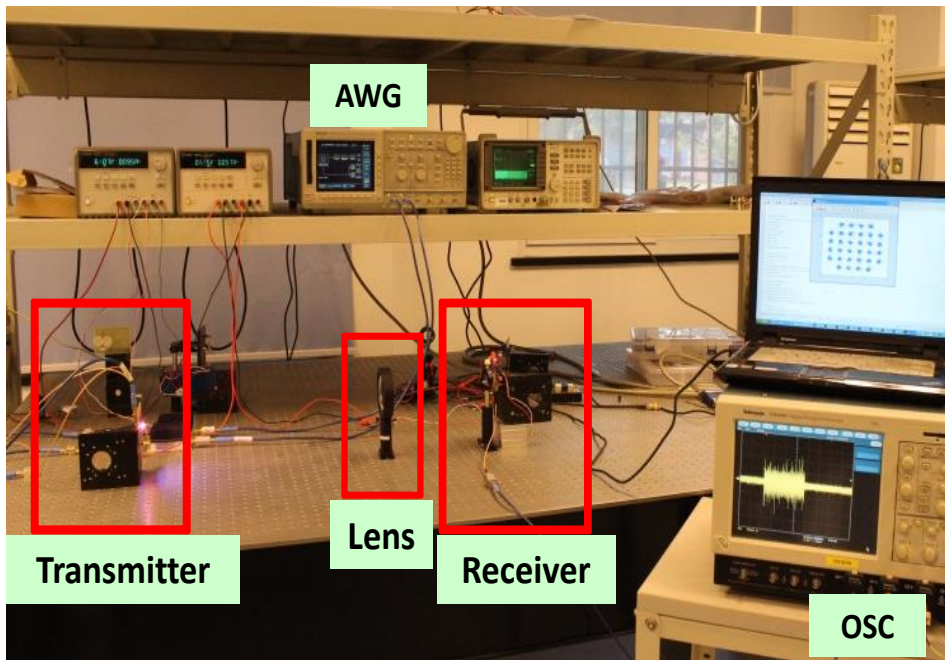
Unipolar modulation (optional): Asymmetrically-Clipped Optical OFDM (ACO-OFDM)



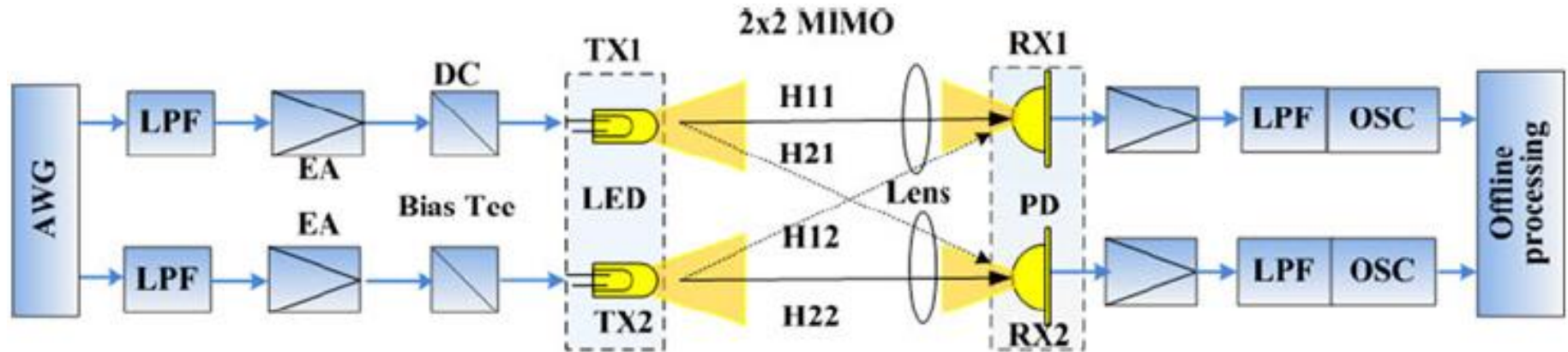
- ACO-OFDM can be used as a unipolar modulation scheme to further improve the power efficiency.
- The time domain signal is made unipolar by simply clipping the negative part at the zero level.
- Only odd subcarriers are modulated by signals, and all of the clipping distortion products fall on the vacant even subcarriers.



Unipolar modulation (optional): Asymmetrically-Clipped Optical OFDM (ACO-OFDM)



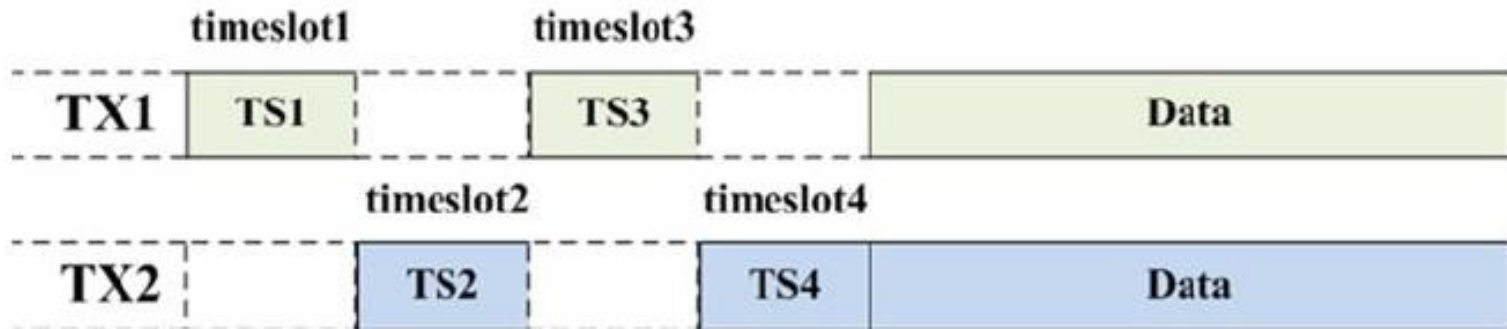
MIMO transmission modes: Non-imaging MIMO



Light from each of the LED arrays is received by all the separate receivers, but with different strengths

$$\begin{pmatrix} Y_1 \\ Y_2 \end{pmatrix} = \begin{pmatrix} H_{11} & H_{12} \\ H_{21} & H_{22} \end{pmatrix} \cdot \begin{pmatrix} X_1 \\ X_2 \end{pmatrix} + \begin{pmatrix} N_1 \\ N_2 \end{pmatrix}$$

Time-multiplexed training symbols for non-imaging MIMO de-multiplexing



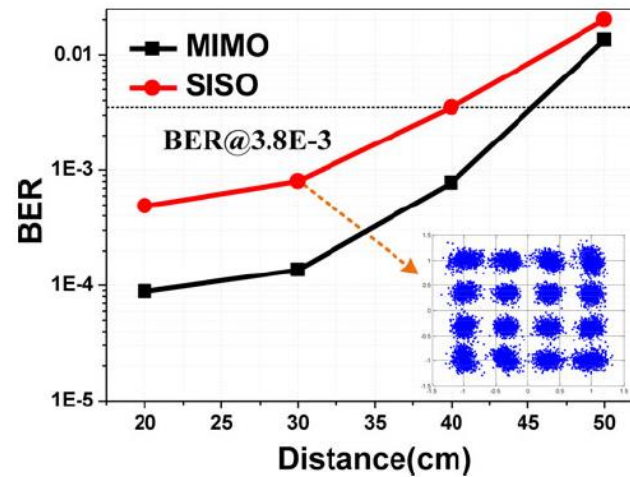
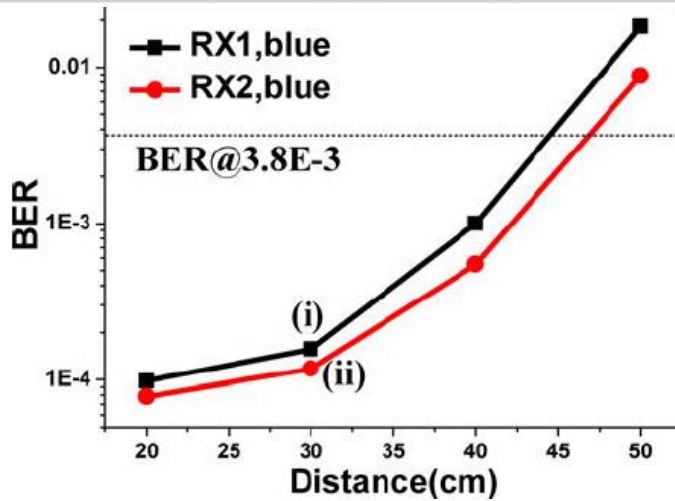
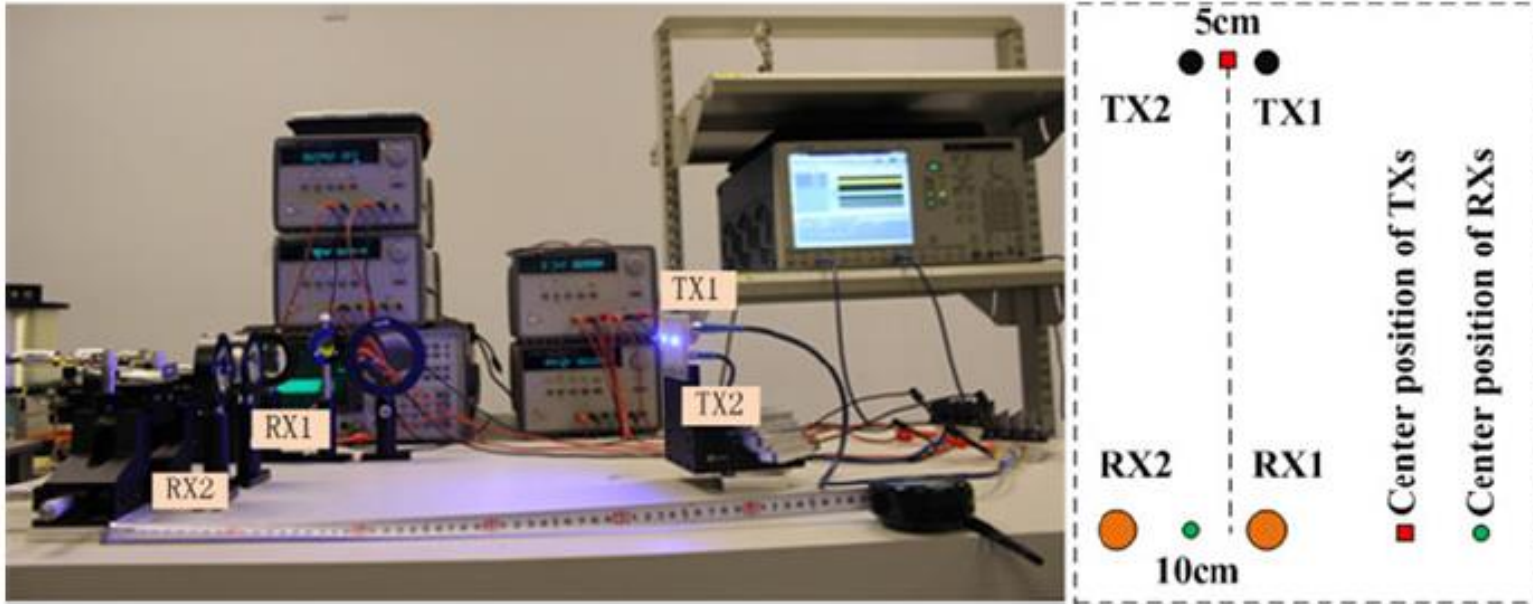
The time-multiplexed TS based frequency domain equalization can be used for MIMO channel estimation and de-multiplexing

$$T_1 = \begin{pmatrix} TS_1 \\ 0 \end{pmatrix}, T_2 = \begin{pmatrix} 0 \\ TS_2 \end{pmatrix} \xrightarrow{\text{Zero forcing}} H = \begin{pmatrix} H_{11} & H_{12} \\ H_{21} & H_{22} \end{pmatrix} = \begin{pmatrix} Y_{11} / TS_1 & Y_{12} / TS_2 \\ Y_{21} / TS_1 & Y_{22} / TS_2 \end{pmatrix}$$

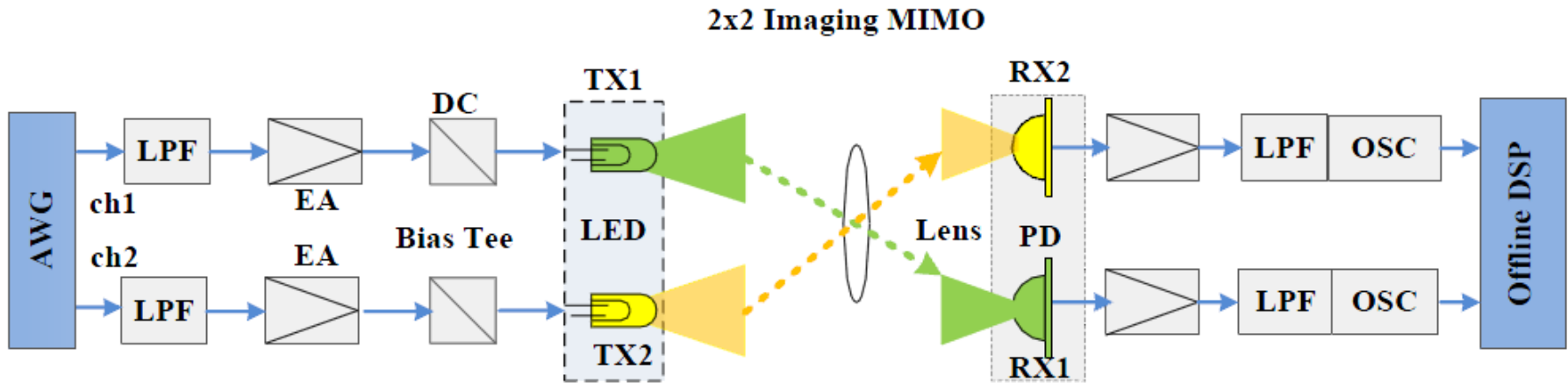
$$X_1 = (H_{22} * Y_1 - H_{12} * Y_2) / (H_{22} * H_{11} - H_{12} * H_{21})$$

$$X_2 = (H_{11} * Y_2 - H_{21} * Y_1) / (H_{22} * H_{11} - H_{12} * H_{21})$$

MIMO transmission modes: Non-imaging MIMO



MIMO transmission modes: Imaging MIMO

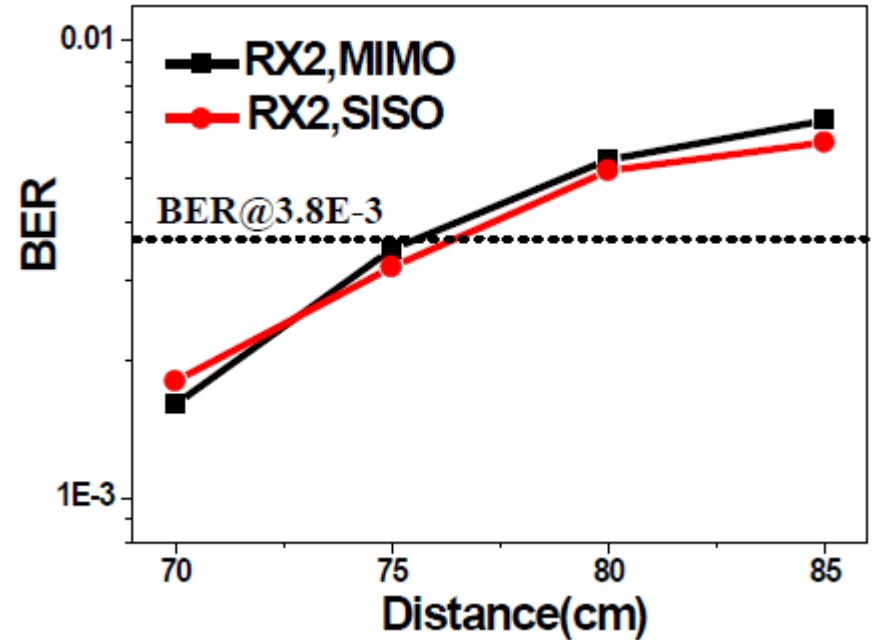
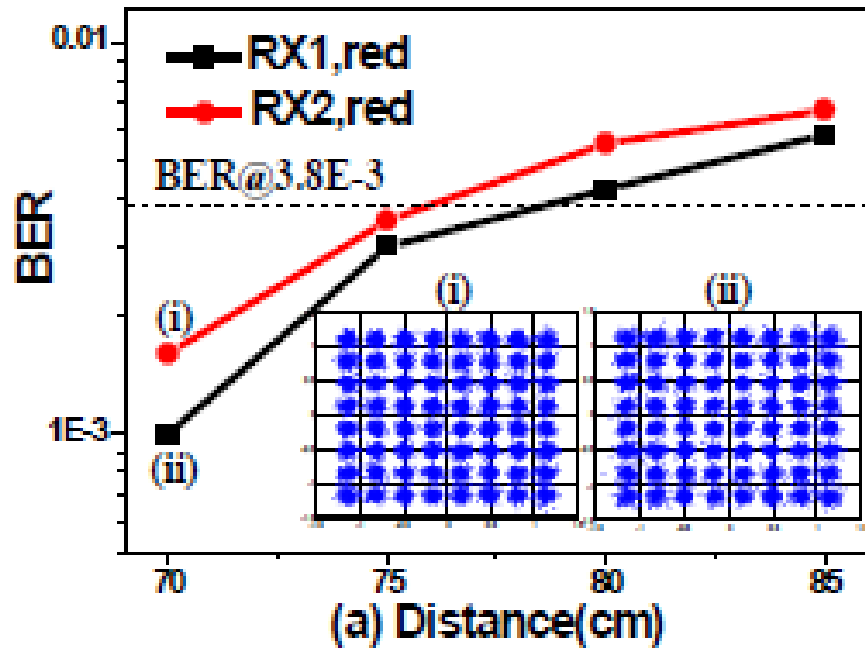


Imaging MIMO requires each LED array imaging onto a detector array, and the light propagates directly to the corresponding detector

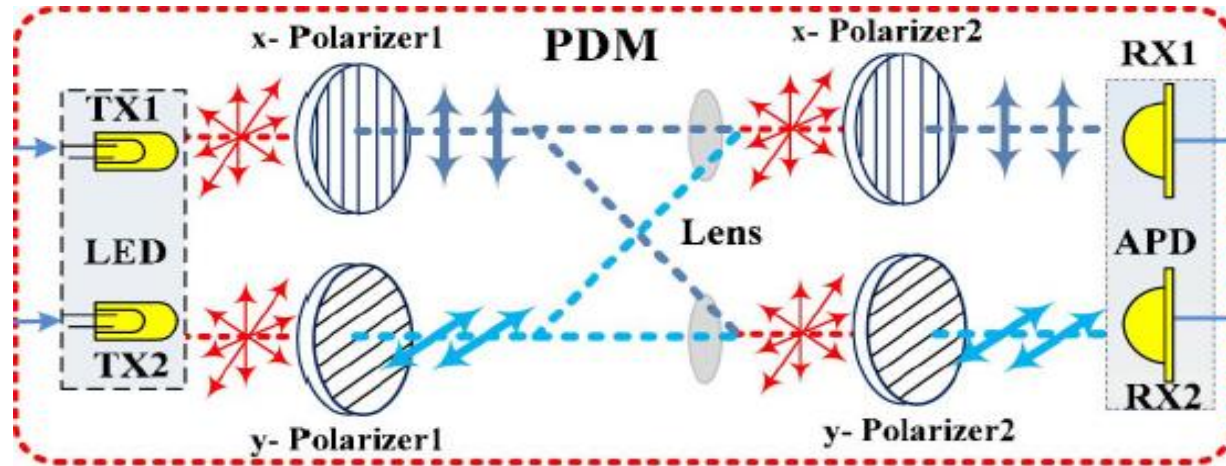
Channel crosstalk can be neglected, and the channel matrix can be simplified

$$H = \begin{pmatrix} H_{11} & 0 \\ 0 & H_{22} \end{pmatrix} = \begin{pmatrix} Y_{11} / TS_1 & 0 \\ 0 & Y_{22} / TS_2 \end{pmatrix}$$

MIMO transmission modes: Imaging MIMO



Polarization division multiplexing (PDM) transmission

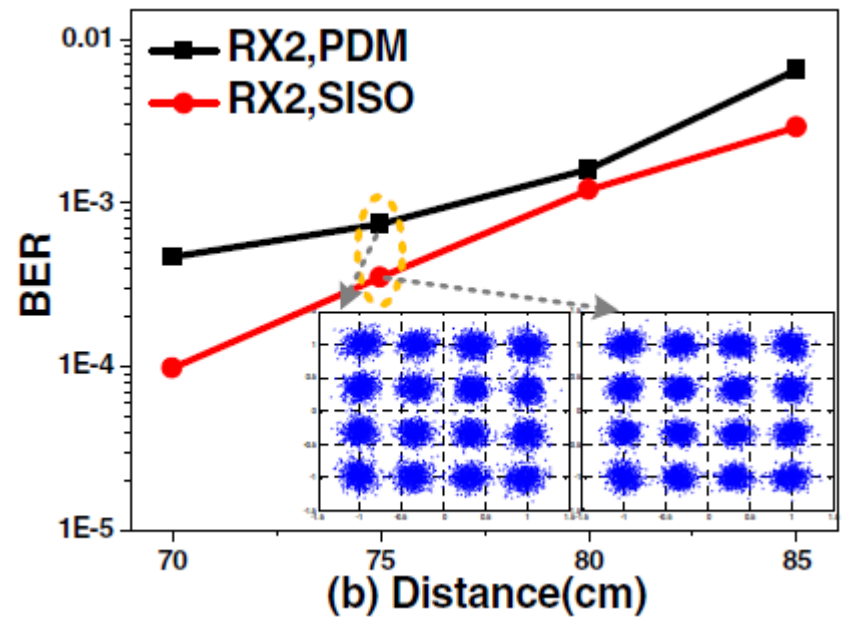
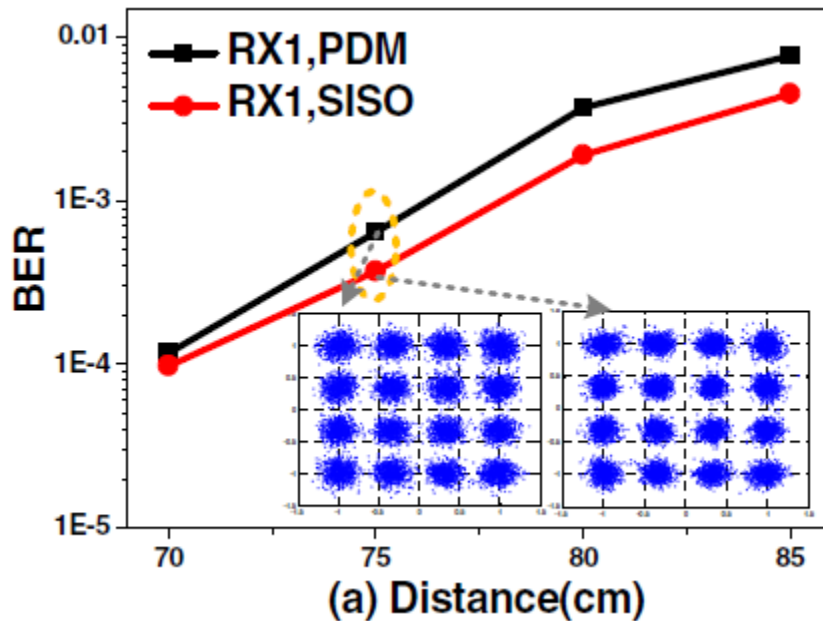
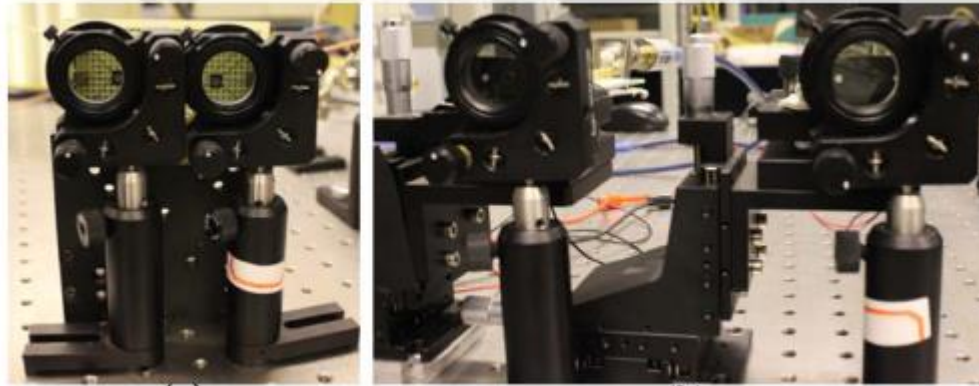


As the lights emitting from incoherent LEDs are natural lights, they can be decomposed as two orthogonal bases of x polarization and y polarization by polarizers.

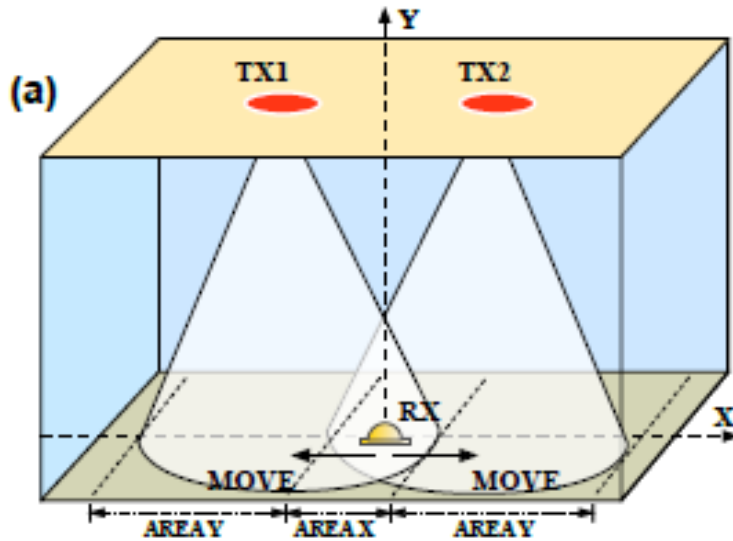
The received optical intensity can be expressed as

$$\begin{pmatrix} Y_1 \\ Y_2 \end{pmatrix} = H \begin{pmatrix} I_1 \\ I_2 \end{pmatrix} + N = \frac{1}{2} \begin{pmatrix} \cos^2 \alpha_{11} & \cos^2 \alpha_{12} \\ \cos^2 \alpha_{21} & \cos^2 \alpha_{22} \end{pmatrix} \begin{pmatrix} I_1 \\ I_2 \end{pmatrix} + N$$

Polarization division multiplexing (PDM) transmission



Channel coding for MIMO: Self-adaptive space-time block coding (STBC) for MISO



STBC technique can easily provide the diversity at receiver for MISO VLC transmission

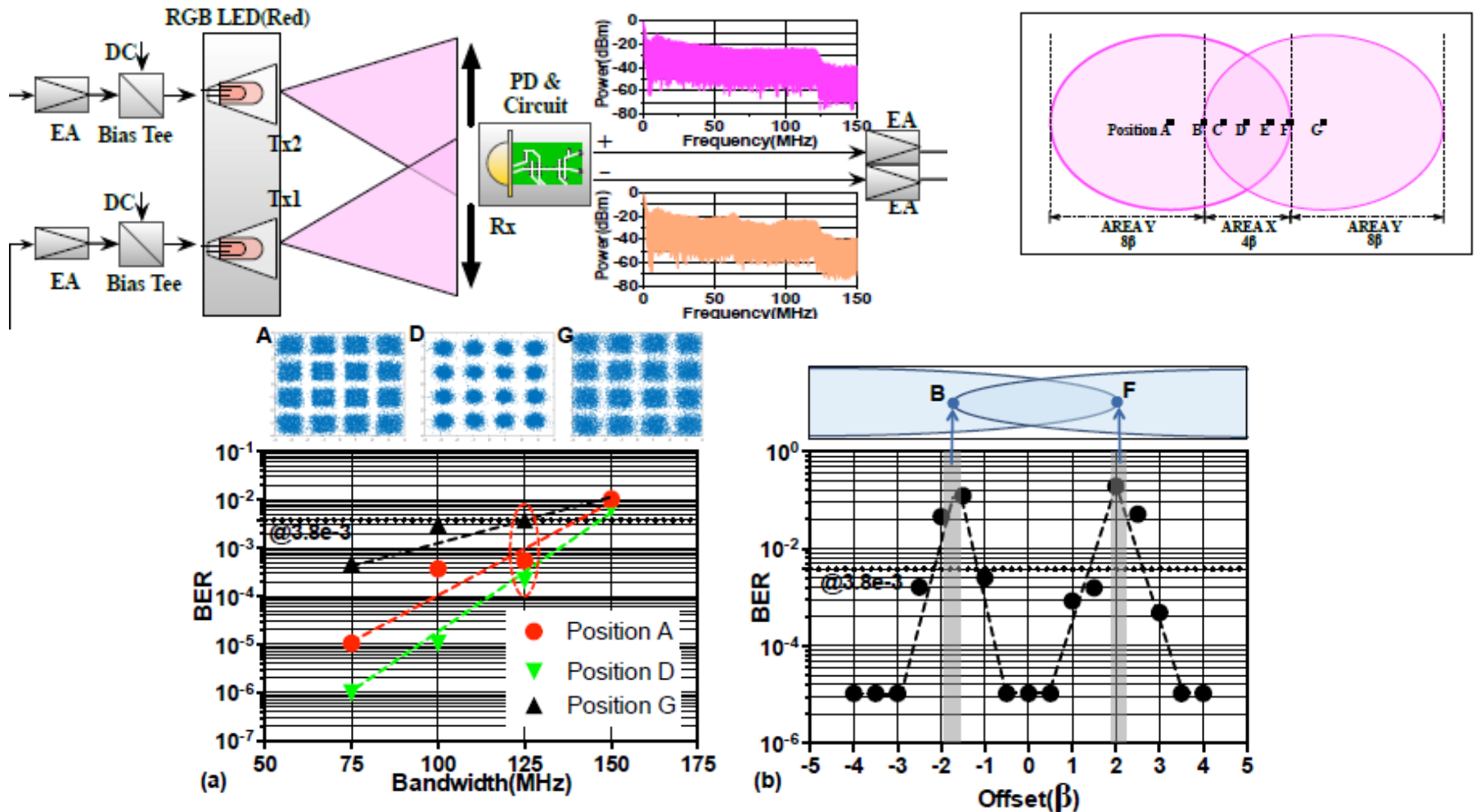
Alamouti's STBC scheme is employed to encode the data with two transmit LEDs and one receiver. Space-time encoding as:

$$\begin{bmatrix} c_1 & c_2 \\ -c_2^* & c_1^* \end{bmatrix}$$

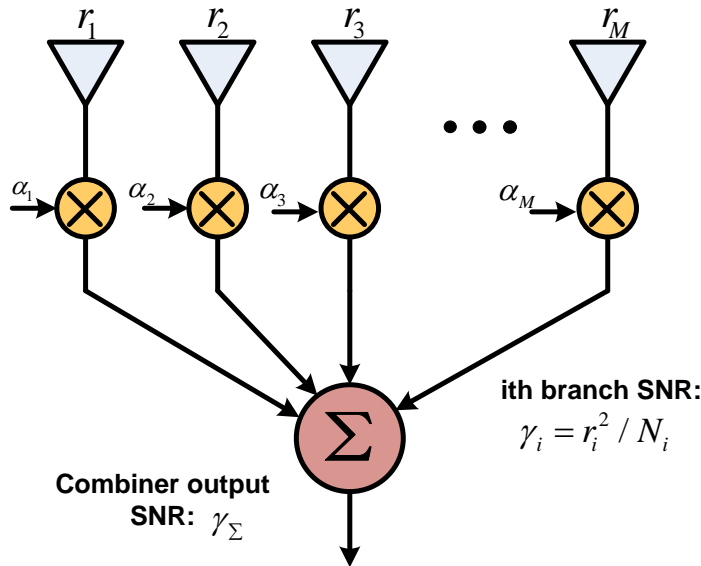
the decoded signal can be obtained as

$$r = \begin{bmatrix} r_1 \\ r_2^* \end{bmatrix} = \begin{bmatrix} h_1 & h_2 \\ h_2^* & -h_1^* \end{bmatrix} \begin{bmatrix} c_1 \\ c_2 \end{bmatrix} + \begin{bmatrix} \eta_1 \\ \eta_2^* \end{bmatrix} = Hc + n \quad \longrightarrow \quad \tilde{r} = \tilde{H}^H r = \tilde{H}^H Hc + \tilde{H}^H n$$

Channel coding for MIMO: Self-adaptive space-time block coding (STBC) for MISO



Channel coding for MIMO: Maximal ratio combining (MRC) based receiver diversity



➤ In receiver diversity, the outputs of multiple receivers are combined which is a weighted sum of the different branches

$$\gamma_\Sigma = \frac{r^2}{N_{tot}} = \frac{\left[\sum_{i=1}^M \alpha_i r_i \right]^2}{\sum_{i=1}^M \alpha_i^2 N_i}$$

- The goal of MRC is to find the weight to maximize the output SNR
- According to the Schwarz inequality: the maximum SNR of the combiner output is the sum of SNRs in each branch:

$$\gamma_\Sigma = \sum_{i=1}^M r_i^2 / N_i = \sum_{i=1}^M \gamma_i$$

Conclusion

In this contribution, we propose several general technique considerations for high-speed VLC modulation formats and MIMO transmission.

□ Modulation formats:

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