

Project: IEEE P802.15 Working Group for Wireless Personal Area Networks (WPANs)

Submission Title: Mobile Channel Characterization in Typical Subway Tunnels at 30 GHz

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Abstract: Based on extensive ray-tracing simulations, this document presents channel characterization for a receiver moving at the speed of 100 m/s in tunnels with 2 GHz bandwidth between 31.5 GHz and 33.5 GHz. The main channel parameters, such as path loss, Rician K-factor, delay spread, Doppler spread, coherence time, decorrelation distance, XPD and CPR are analyzed for three antenna setups.

Purpose: The Channel characteristics can be helpful for link-level simulation of mobile communications in tunnel environments at 30 GHz

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Outline

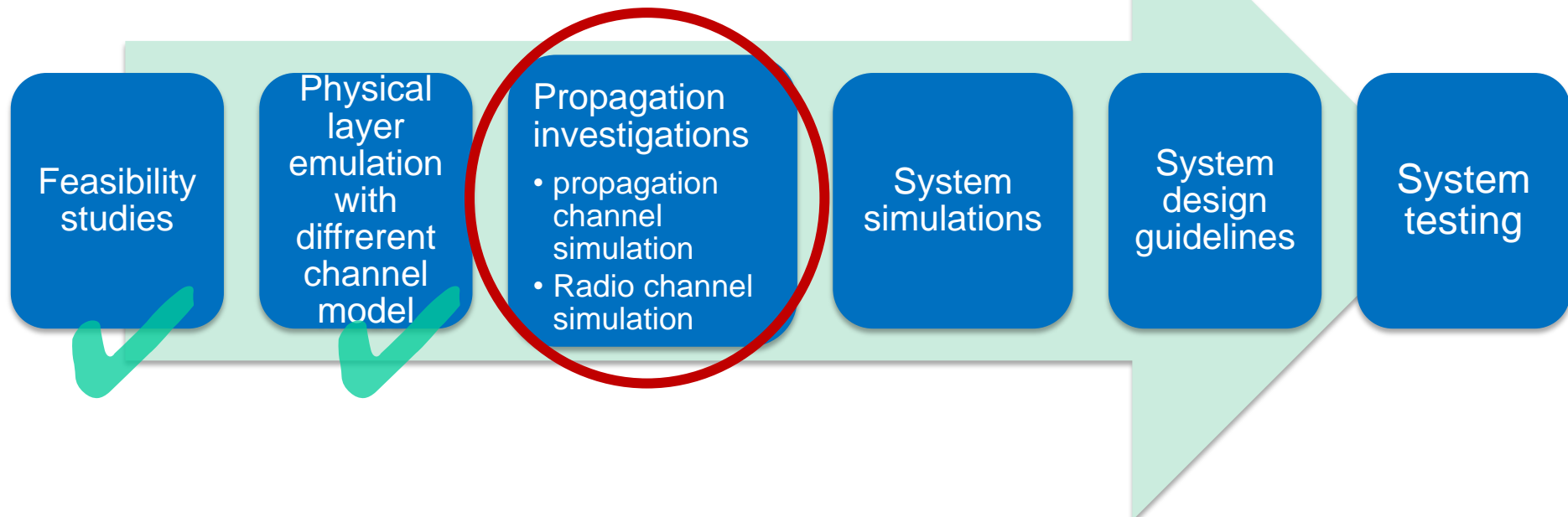
- **Motivation of channel characterization for HRRC**
- **Deterministic channel modeling approach**
 - **Ray-tracing simulator**
 - **Frequency domain simulation**
- **Channel simulation and characteristics**
- **Conclusion and future work**

Outline

- **Motivation of channel characterization for HRRC**
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1. Motivation of channel characterization for HRRC

Gbps data rate with high performance should be provided to the user groups inside of the fast-moving vehicles.



Outline

- Introduction of BJTU and RCS
- Motivation of channel characterization for HRRC
- **Deterministic channel modeling approach**
 - Ray-tracing simulator
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2. Deterministic channel modeling approach

Ray-tracing based channel modeling

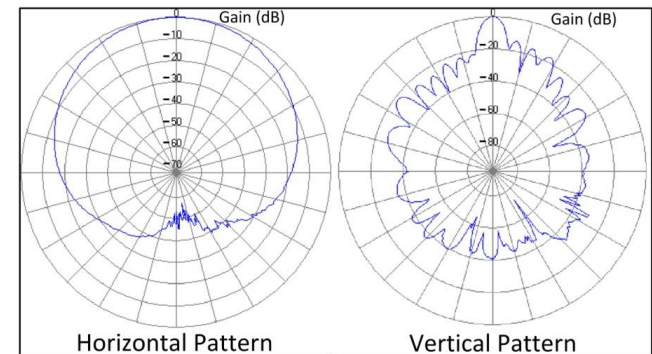
Deterministic modeling approach towards real scenarios

3D ray optical channel model:
Transmission, reflection, scattering

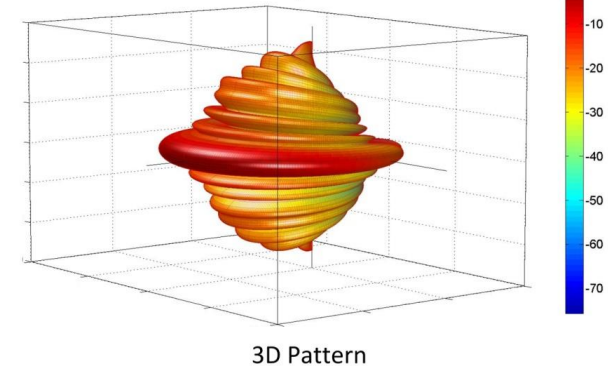
Antenna modeling

Ray optical method

$$h(\tau, t) = \sum_{k=1}^{N(t)} a_k(t) \cdot e^{j(2\pi f \tau_k(t) + \varphi_k(t))} \cdot \delta(\tau - \tau_k(t))$$

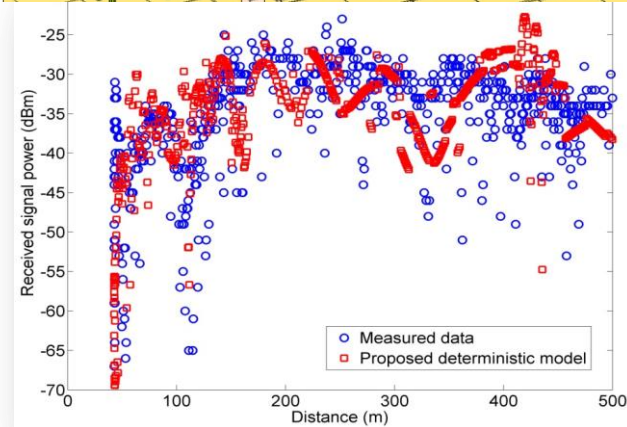
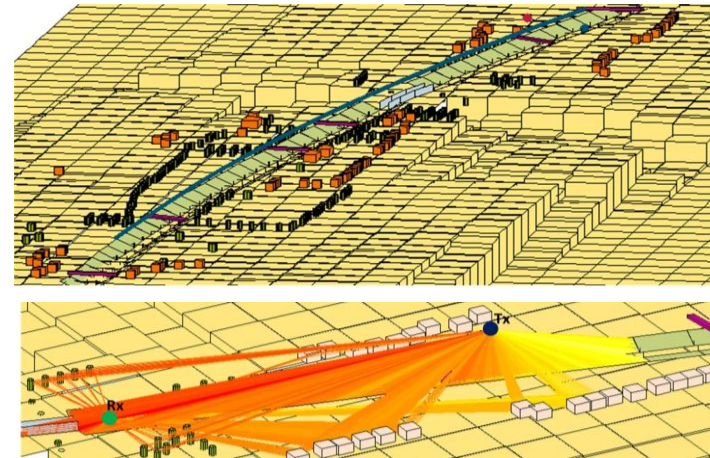
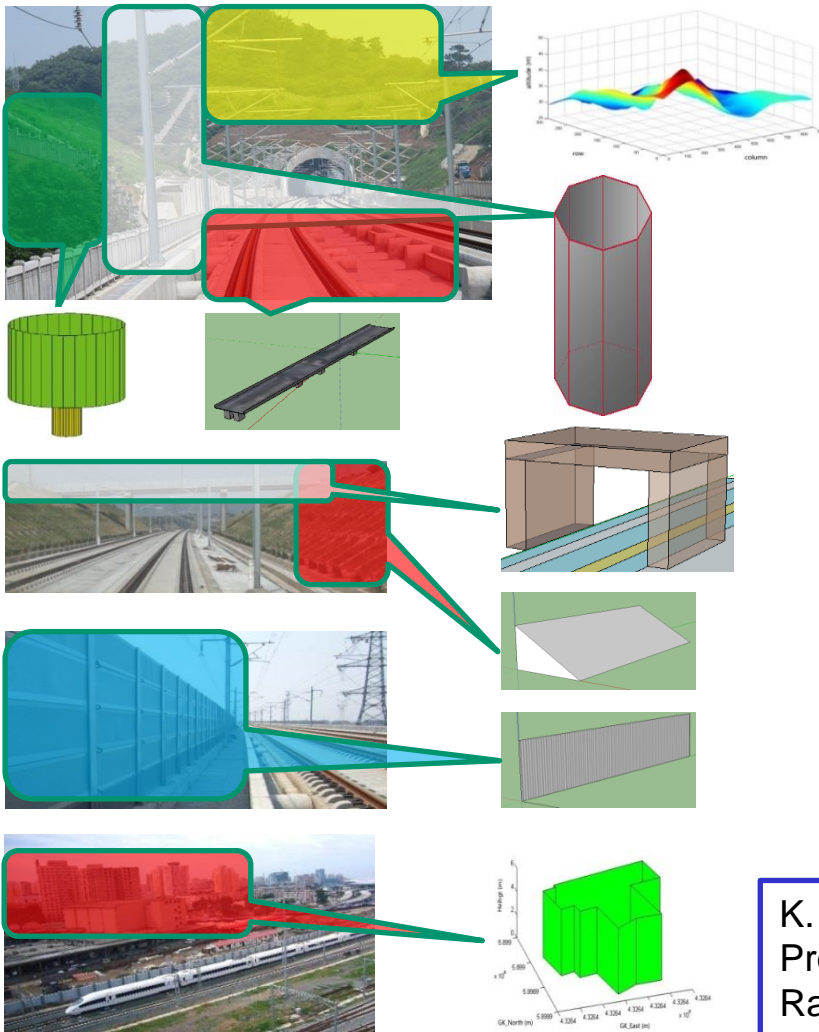


Interpolation



J. Nuckelt, et al., "Deterministic and stochastic channel models implemented in a physical layer simulator for Car-to-X communications," 2010 Advances in Radio Science, 2010.

2. Deterministic channel modeling approach



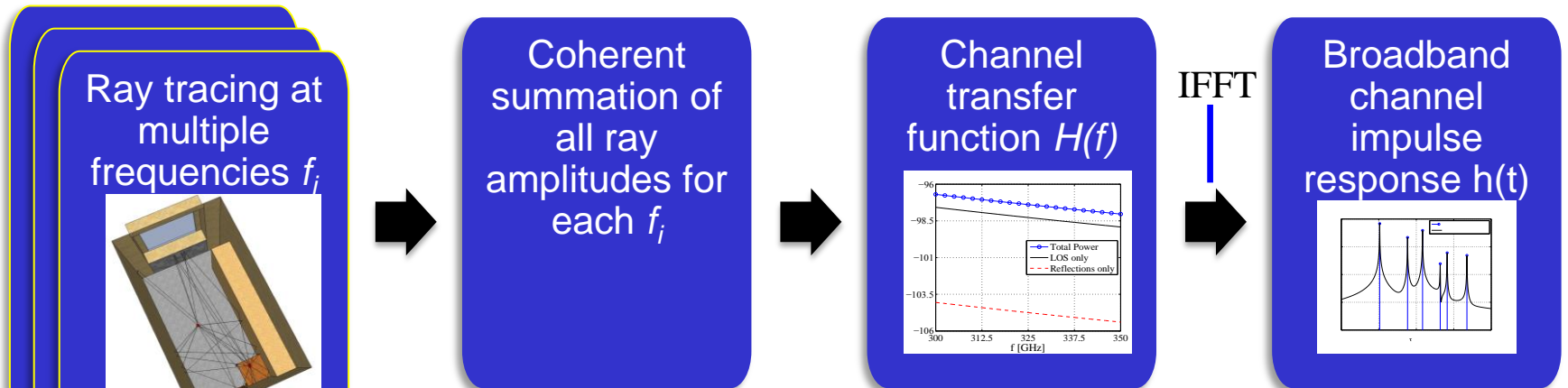
K. Guan, Z.D. Zhong, B. Ai, and T. Kuerner, "Deterministic Propagation Modeling for the Realistic High-Speed Railway Environment," IEEE 77th VTC2013-Spring, Dresden, Germany, pp. 1-5, June 2013.

2. Deterministic channel modeling approach

UWB channel: Frequency domain simulation

UWB → Significant frequency dispersion → Single frequency ray tracing insufficient

Idea: Channel simulation in frequency domain



$$H(f_i) = \sum_{j=1}^{N_{\text{Rays}}} E_j(f_i)$$

Ray amplitudes from ray tracing

$$H(f) = \sum_{i=1}^{N_f} H(f_i)$$

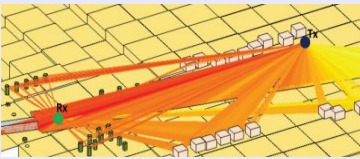
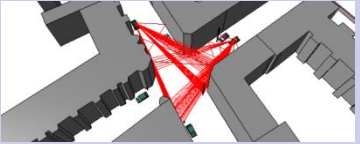
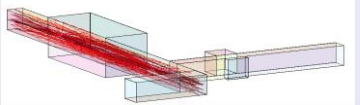
$$N_f \cdot \left(\frac{1}{B}\right) \geq \tau_{\text{max}}$$

$$N_f \geq \tau_{\text{max}} \cdot B$$

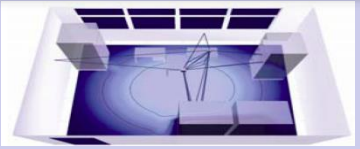
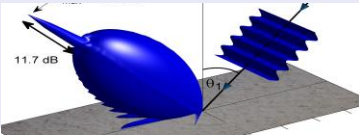
S. Priebe and T. Kürner, "Stochastic Modeling of THz Indoor Radio Channels," IEEE Trans. Wireless Commun., vol.12, no.9, pp. 4445–4455, Sept. 2013.

2. Deterministic channel modeling approach

The ray-tracing simulator has been verified by extensive measurements:

| Frequency | System | Scenario | reference |
|-----------|--------|---|--|
| 930 MHz | GSM-R | High-speed railway  | K. Guan, et al., "Deterministic Propagation Modeling for the Realistic High-Speed Railway Environment," IEEE VTC2013-Spring, Dresden, Germany, pp. 1-5, June 2013. |
| 5.9 GHz | DSRC | Urban  | T. Abbas, et al., "Simulation and Measurement Based Vehicle-to-Vehicle Channel Characterization: Accuracy and Constraint Analysis," IEEE Trans. on Ant. and Prop., 2015. |
| 15 GHz | 5G D2D | Corridor  | Q. Wang, et al., "Ray-Based Analysis of Small-Scale Fading for Indoor Corridor Scenarios at 15 GHz," APEMC 2015 |

30 GHz, HRRC, Typical subway tunnel

| | | | |
|-----------------|--------------|---|---|
| 60 GHz | WLAN | Indoor  | M. Jacob, et al., "Diffraction in MM and Sub-MM Wave Indoor Propagation Channels," IEEE Transactions on Microwave Theory and Techniques, vol.60, no.3, pp.833-844, Mar. 2012. |
| 275 GHz-325 GHz | WLAN WPAN | Indoor  | S. Priebe et al., "Stochastic Modeling of THz Indoor Radio Channels," IEEE Trans. Wireless Commun., vol.12, no.9, pp. 4445–4455, 2013. |

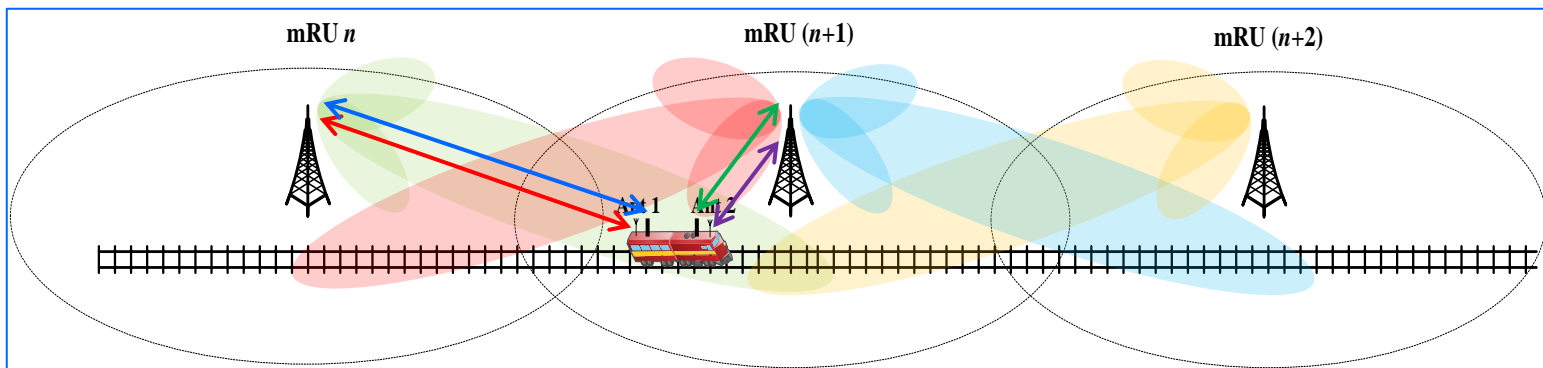
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3. Channel simulation and characteristics

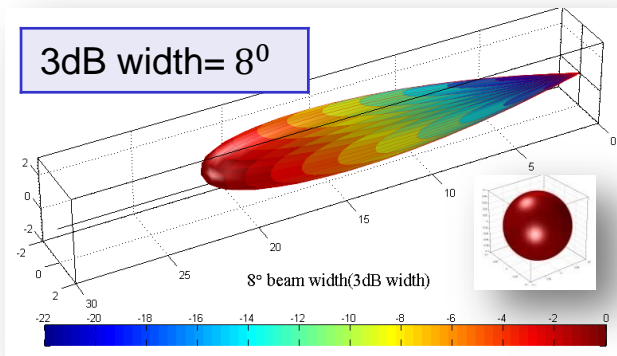
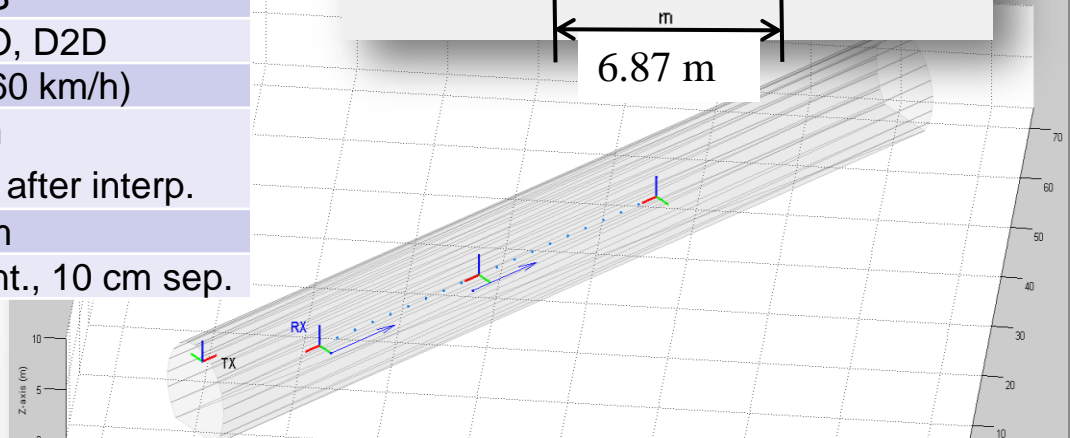
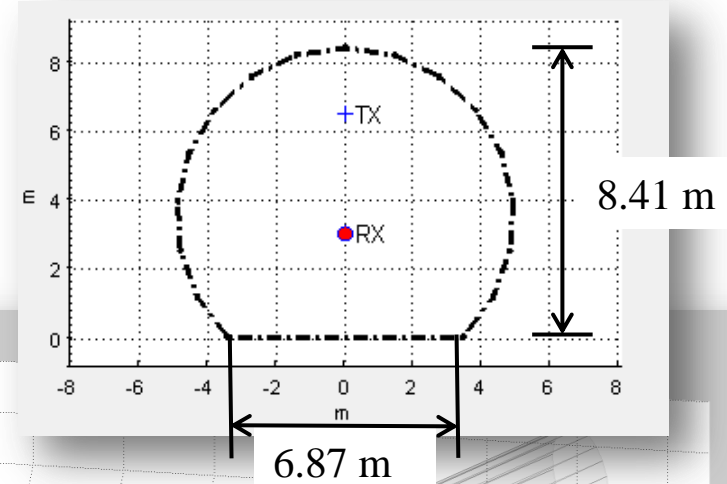
| System: Mobile Hotspot Network (MHN) | |
|--------------------------------------|---|
| Cell Coverage | 1 km (mRU interval) |
| Frequency | 31.5~33.5 GHz |
| Bandwidth | 125 MHz x 4 |
| Length of Train | 200 m |
| Antenna Setup | Head ant. + Tail ant. |
| MIMO Configuration | 1x1, 1x2, 2x1, 2x2 |
| Target Maximum Speed | 400 ~ 500 km/h |
| Antenna Type | Patch array antenna (directional ant.) w. 8° beam width (3 dB width) |
| Antenna Separation | Around 10 cm |

Dae-Soon Cho, et al., "Performance of downlink control channels for Mobile Hotspot Network system," in 2013 International Conference on ICT Convergence (ICTC), pp.909-912, 14-16 Oct. 2013



3. Channel simulation and characteristics

| Simulation Setup | |
|--------------------|---|
| Environment | Subway tunnel |
| Tunnel Type | Arched subway tunnel |
| Frequency | 31.5~33.5 GHz |
| Bandwidth | 125 MHz (Time resolution: 8 ns) |
| Antenna Height | 6.5 m for mRU, 3 m for mTE |
| Transmission Power | 20 dBm |
| Cable Loss | 6 dB |
| Antenna Type | O2O, D2O, D2D |
| Speed | 100 m/s (360 km/h) |
| Sampling Interval | 1 m |
| Distance | 2 mm ($\approx 0.2\lambda$) after interp. |
| Distance | 1 km |
| Antenna Setup | 2 Head + 2 Tail ant., 10 cm sep. |



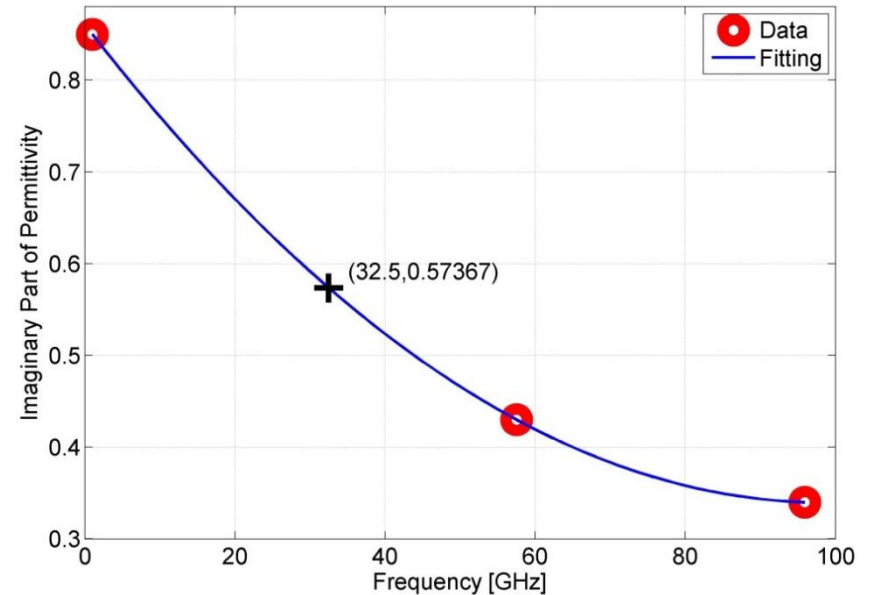
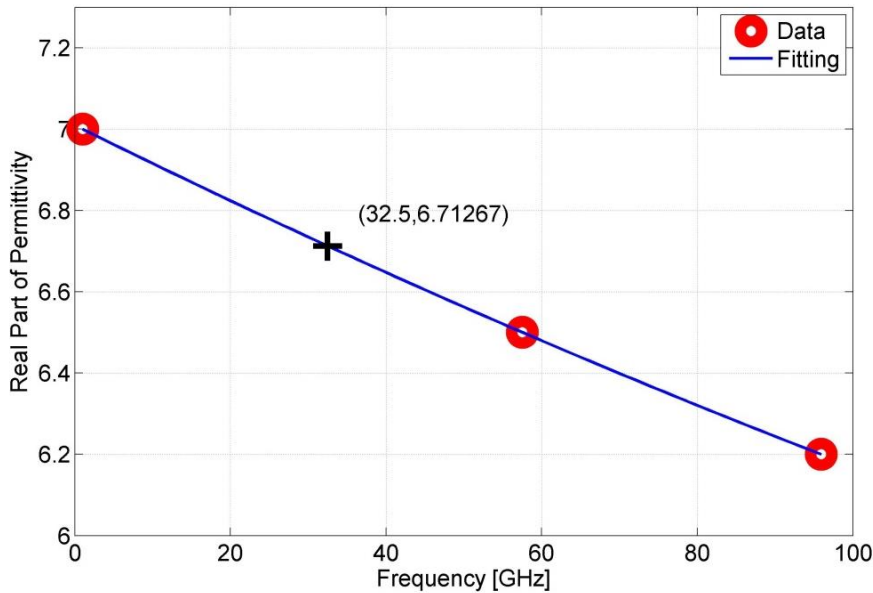
- K. Guan, et al., “Measurements and Analysis of Large-Scale Fading Characteristics in Curved Subway Tunnels at 920 MHz, 2400 MHz, and 5705 MHz,” to appear, IEEE Transactions on Intelligent Transportation Systems, 2015.
- M. Schack, Integrated Simulation of Communication Applications in Vehicular Environments, Ph.D. Dissertation

Electromagnetic property

| Material | |
|---|----------|
| Tunnel walls | Concrete |
| Real part of relative permittivity | 6.71267 |
| Imaginary part of relative permittivity | 0.67367 |

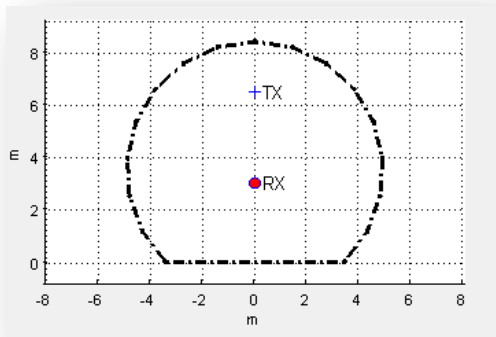
$$\epsilon = \epsilon' - j\epsilon''$$

$$\text{Loss tangent} = \arctan \frac{\epsilon''}{\epsilon'}$$



Recommendation ITU-R P.1238-7-EM Property

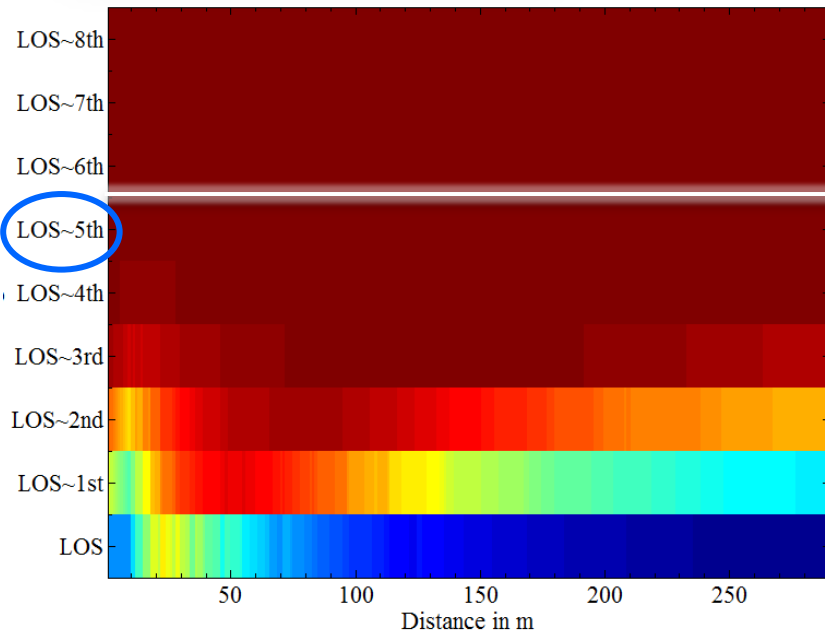
Geometry of tunnel and orders of reflections



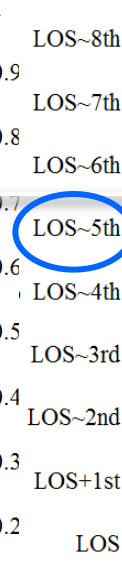
Tunnel – 17 faces
Direct ray + reflected rays
Order of reflections – 5

K. Guan, et al., “Measurements and Analysis of Large-Scale Fading Characteristics in Curved Subway Tunnels at 920 MHz, 2400 MHz, and 5705 MHz,” to appear, IEEE Transactions on Intelligent Transportation Systems, 2015.

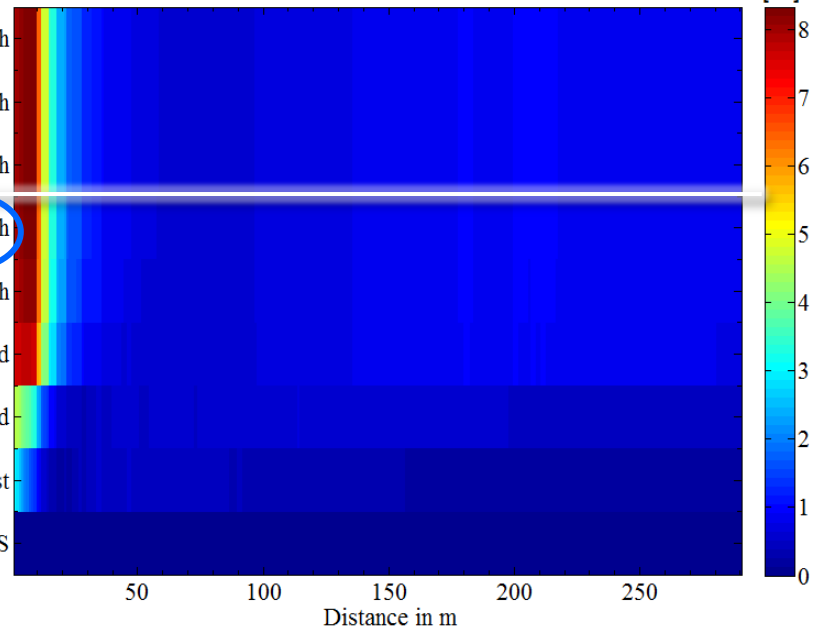
Received power



Rate



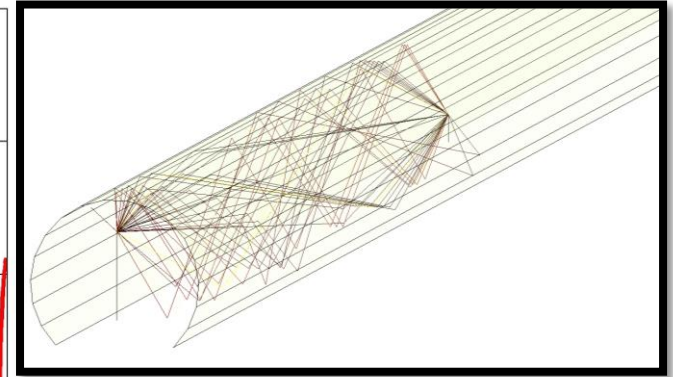
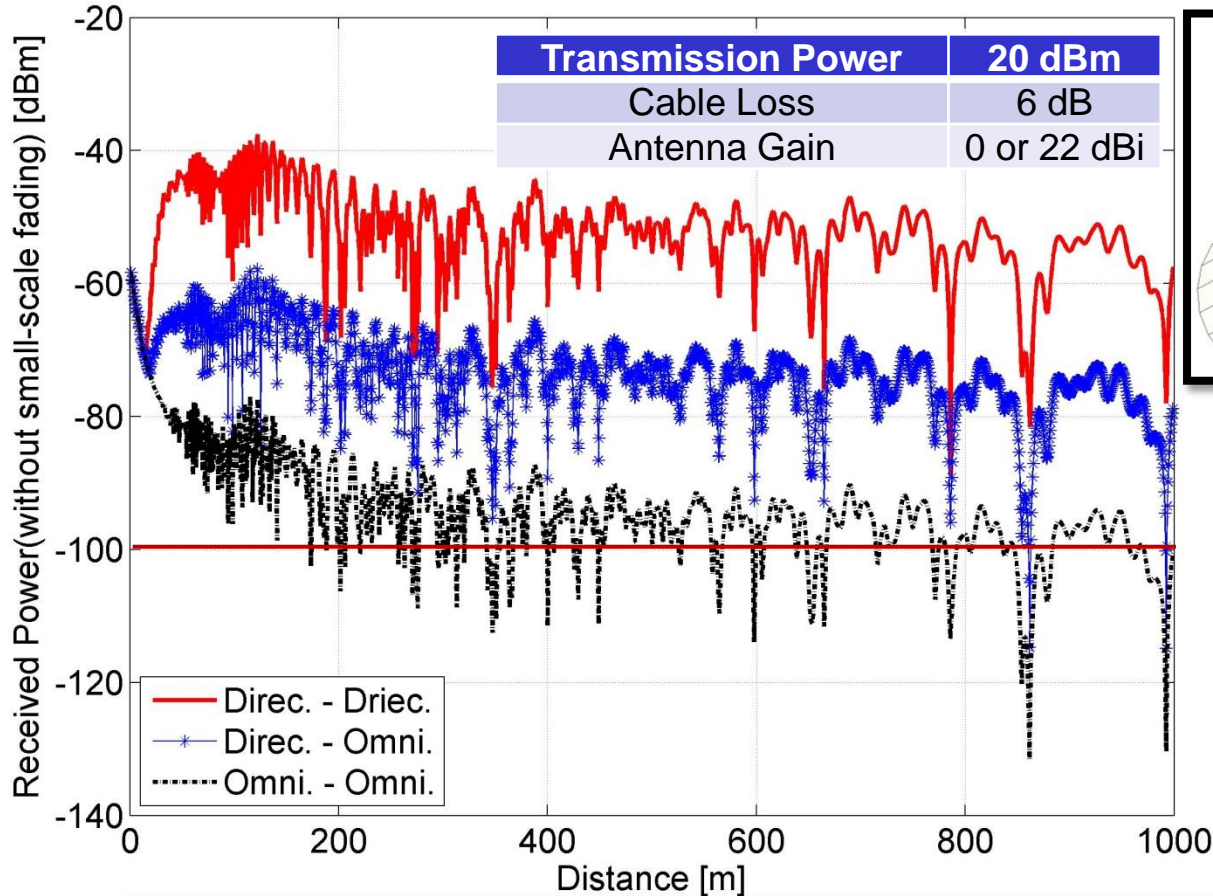
RMS delay spread



Parameters of channel characteristics

| Parameters of channel characteristics | |
|---------------------------------------|---|
| Loss and fading | Received power |
| | Rician K-factor |
| Power delay profile | RMS delay spread |
| Doppler spectrum | RMS Doppler spread |
| | Mean Doppler shift |
| Time-varying property | Coherence time |
| Correlation of shadow fading | Decorrelation distance |
| | Cross-correlation coefficient |
| Polarization | Cross polarization discrimination (XPD) |
| | Co-polarization power ratio (CPR) |

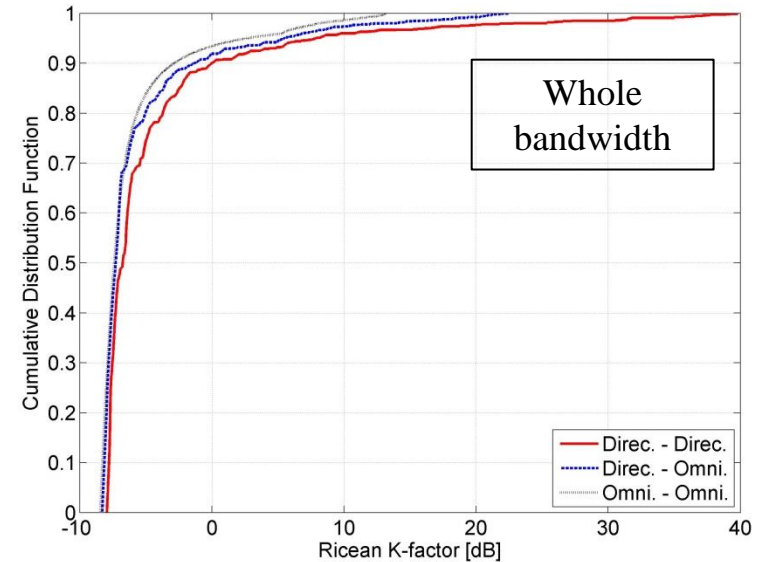
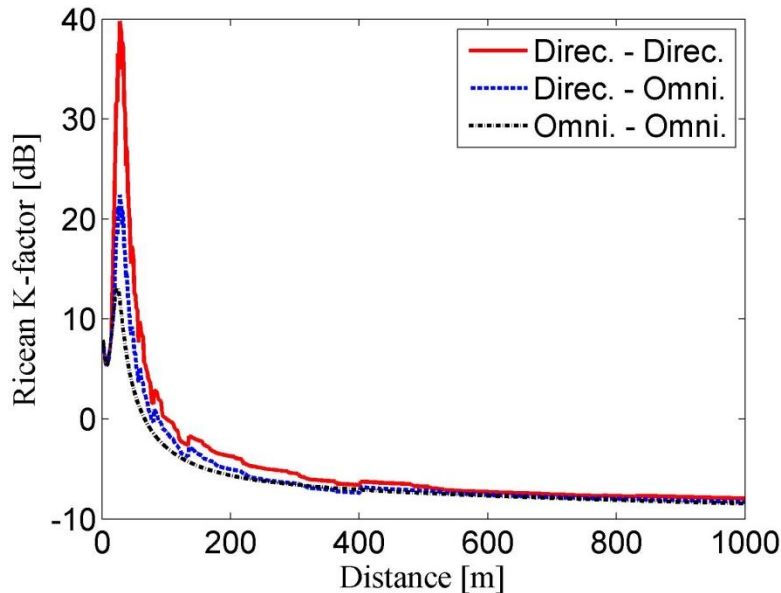
Received power (without small scale fading)



| Coverage range [m] | -100 dBm |
|--------------------|----------|
| Direc. – Direc. | 1000 + |
| Direc. – Omni. | 1000 |
| Omni. – Omni. | < 300 |

- Enough coverage can be realized by usage of directional antenna
- At least one directional antenna should be used for enough link length

Rician K-factor



- Rician K-factor is defined as the power ratio of LOS components to NLOS components.
- $K=0$ -- Rayleigh fading; $K=\infty$ -- no fading.

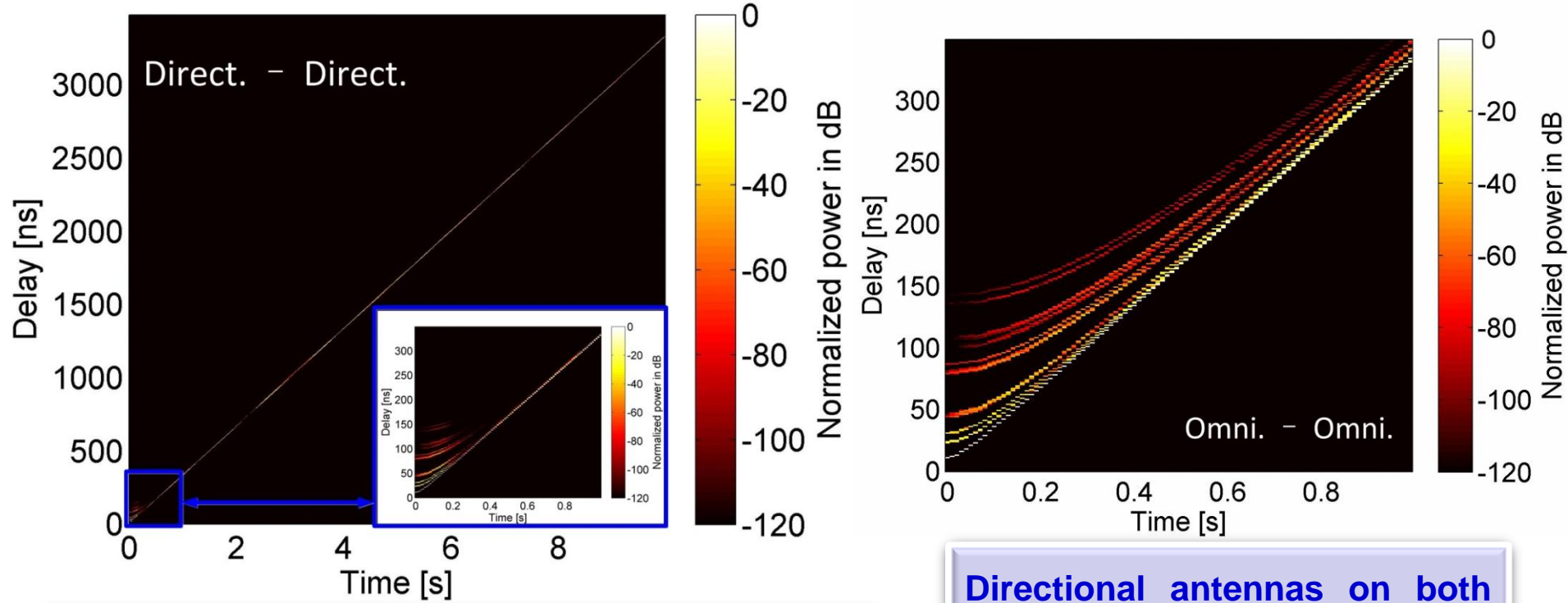
| Rician K-factor [dB] | 50% | 90% |
|----------------------|--------|--------|
| Direc. – Direc. | -6.750 | -0.006 |
| Direc. – Omni. | -7.280 | -1.368 |
| Omni. – Omni. | -7.400 | -2.844 |

In the first 50 m, channel fading is weak, but after 50 m, channel goes through a serious fading.

Andrea Goldsmith. 2005. Wireless Communications. Cambridge University Press, New York, NY, USA.

Power delay profile

Normalized Power Delay Profile (32.5GHz)



Very short delay spread due to LOS condition, high frequency and limited space

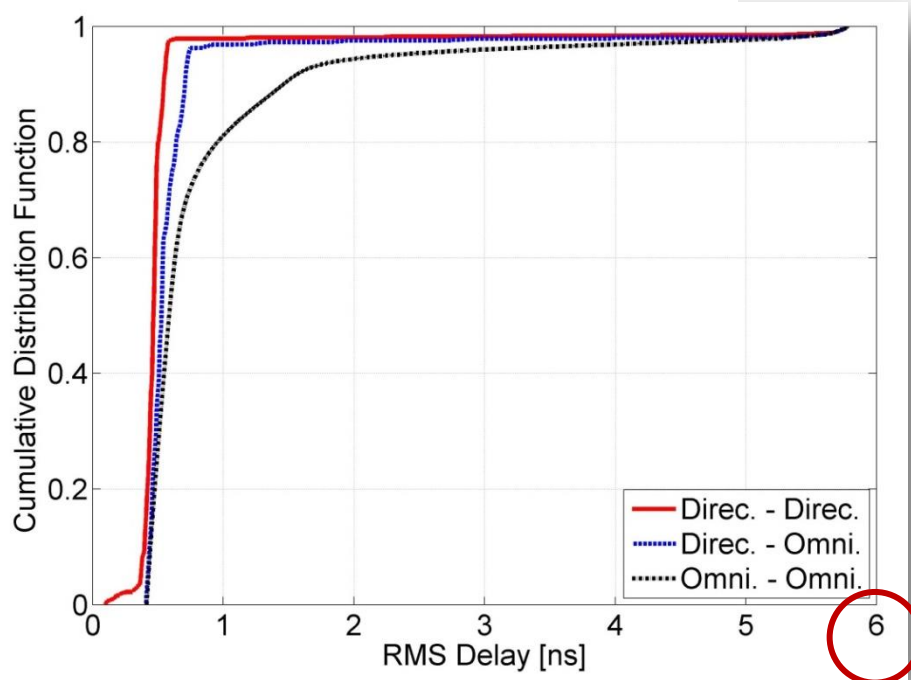
Directional antennas on both Tx and Rx sides reduce the delay spread

Root-mean-square (RMS) delay spread

$$\tau_{\text{rms}} = \sqrt{\frac{\sum_{r=1}^N \tau_r^2 X(r)}{\sum_{r=1}^N X(r)} - \left(\frac{\sum_{r=1}^N \tau_r X(r)}{\sum_{r=1}^N X(r)} \right)^2}$$

$X(r)$ ($r=1,2,\dots,N$) be the r -th sample of the power delay profile (PDP) at a certain time

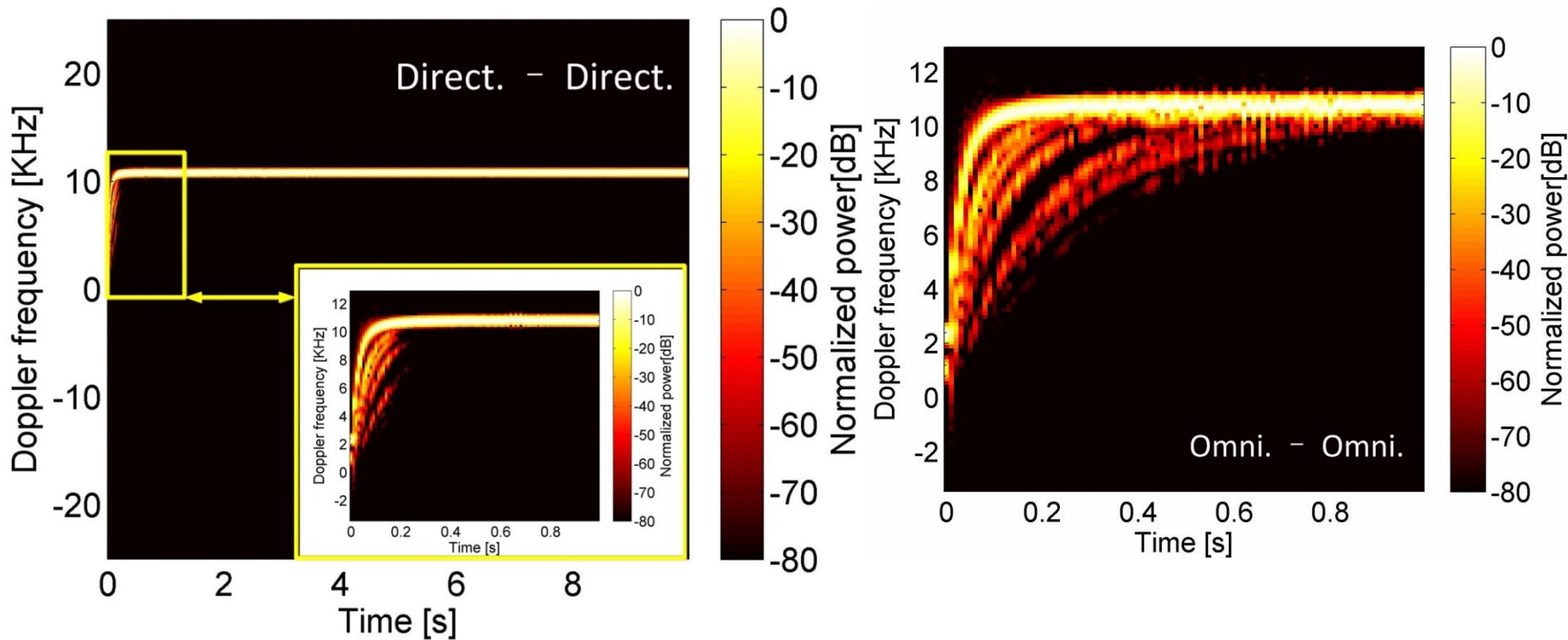
τ_r denote the excess delay of the r -th sample of PDP



| RMS delay spread [ns] | 50% | 90% |
|-----------------------|-------|-------|
| Direc. – Direc. | 0.467 | 0.545 |
| Direc. – Omni. | 0.528 | 0.708 |
| Omni. – Omni. | 0.586 | 1.463 |

So short time delay spread will not introduce Inter-symbol Interference (ISI)

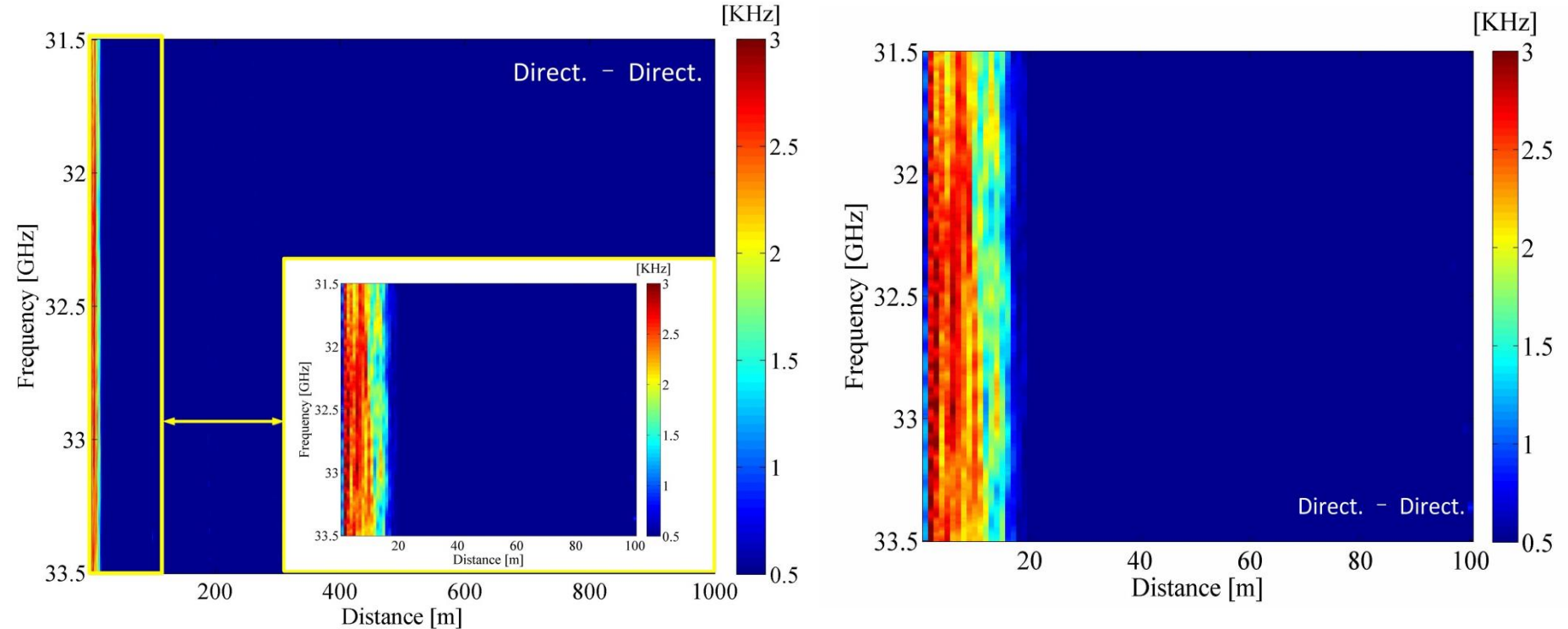
Doppler spectrum at 32.5 GHz



- Maximum Doppler shift is around 12 kHz
- Doppler spread mainly happens in the first 100 m

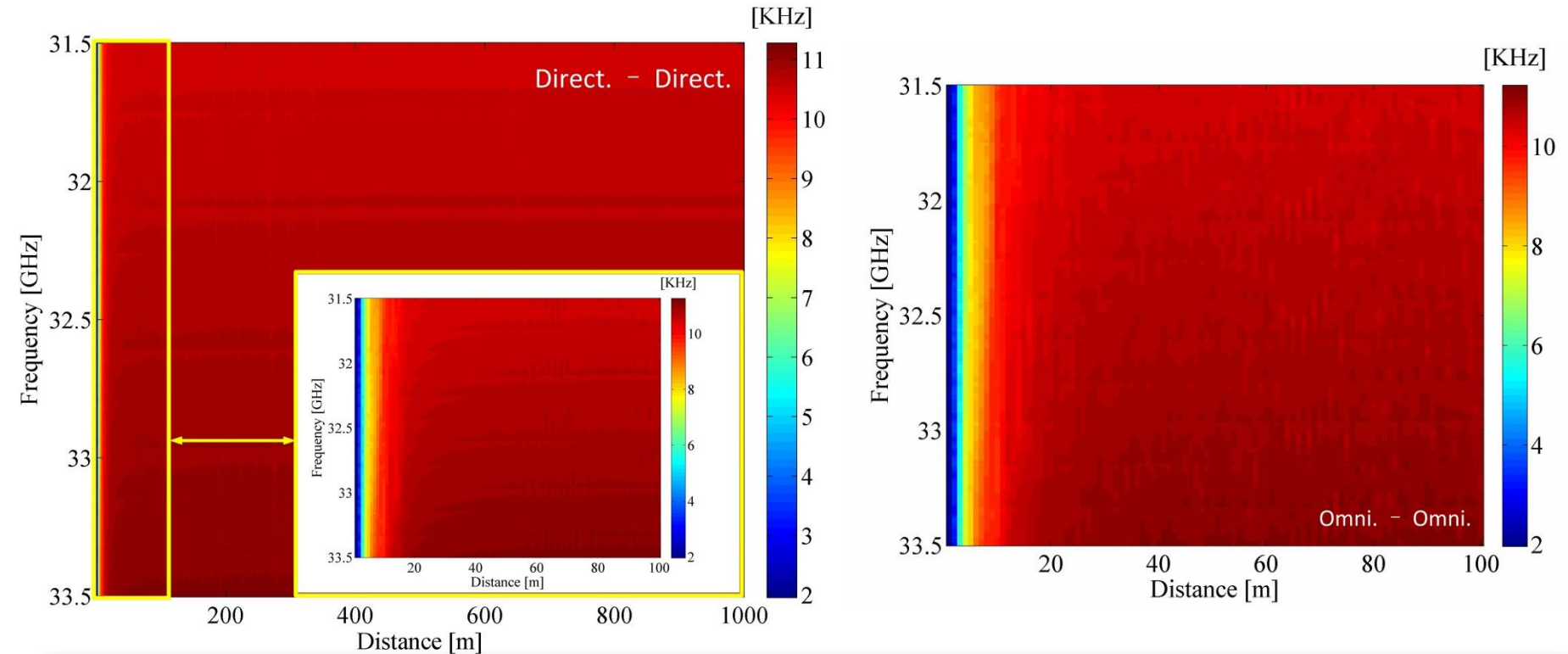
Directional antennas reduce the RMS Doppler spread

RMS Doppler Spread for the whole 2-GHz bandwidth



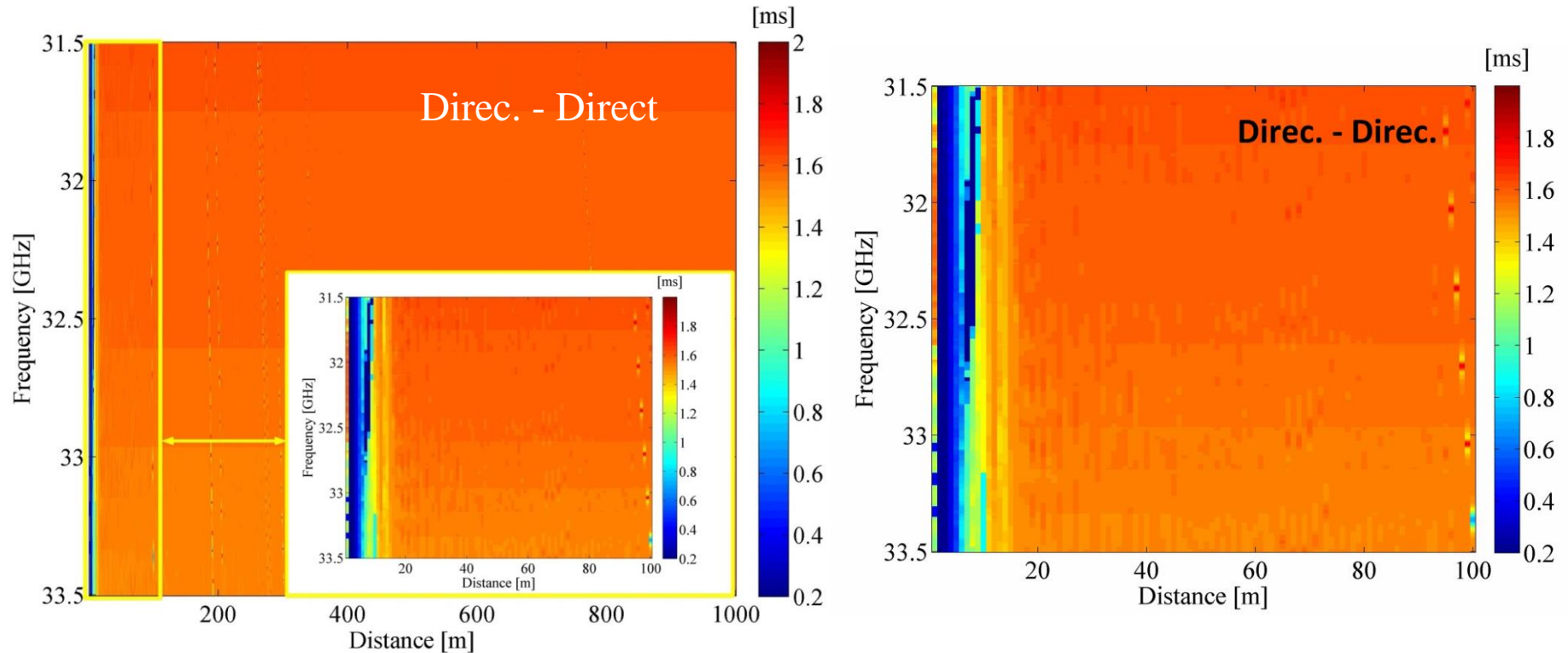
- In the first 20 m, the RMS Doppler spread is up to 3 kHz
- Directional antennas reduce the distance of the signal suffering Doppler spread

Mean Doppler Shift at 32.5 GHz



- Mean Doppler shift increases very fast up to 12 kHz within the first 10 m
- Directional antennas do NOT influence Doppler shift as it comes from the direct path at short distance, and at long distance the reflections are close to direct paths.

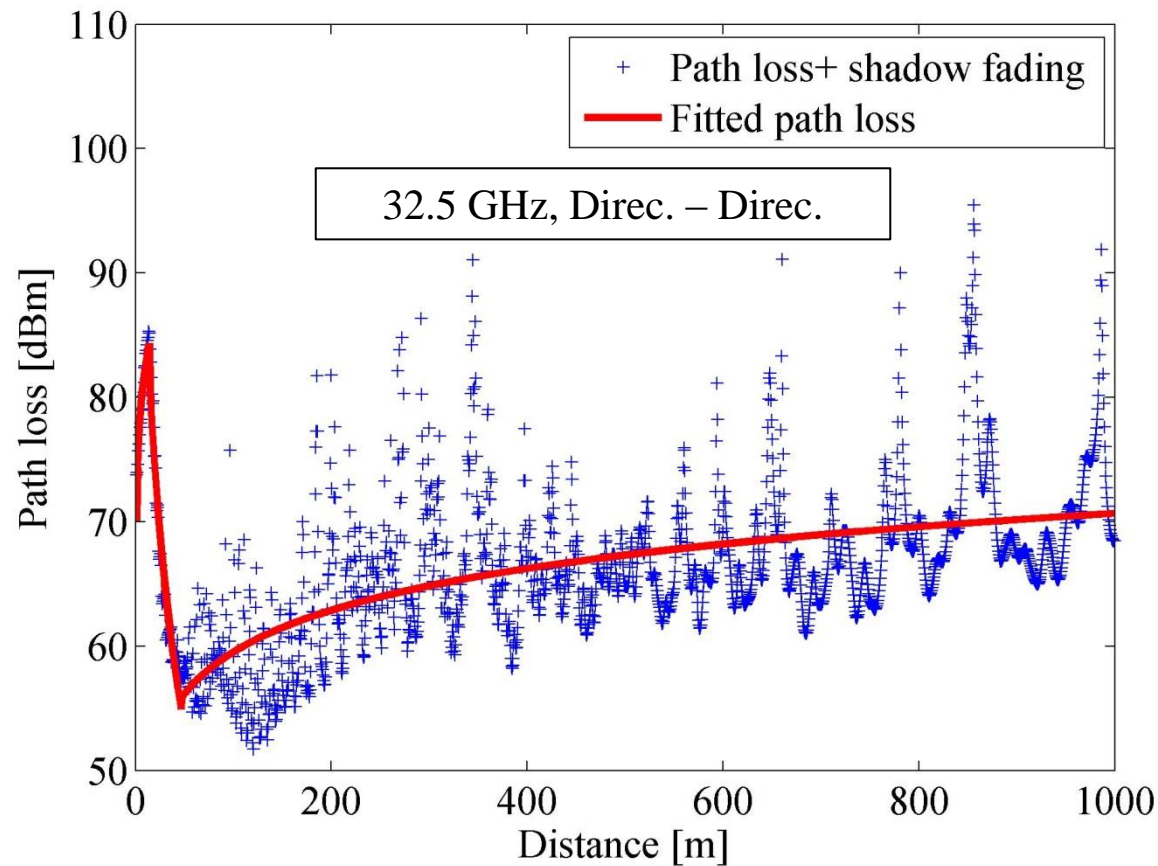
Coherence time



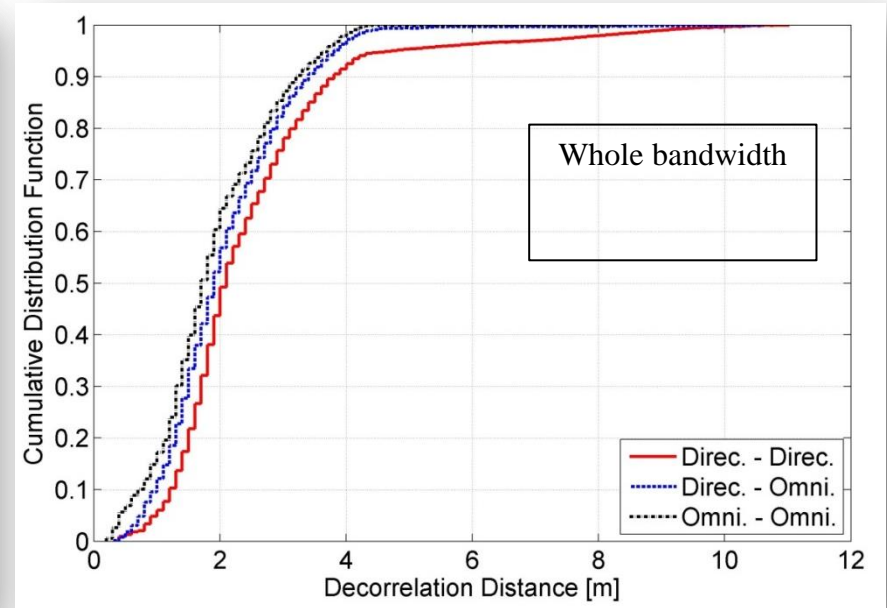
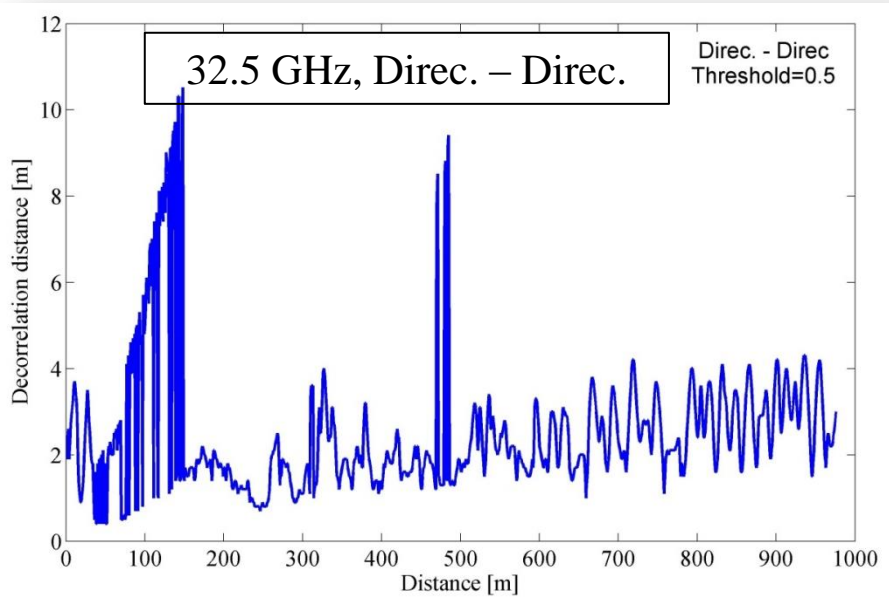
- Coherence time increases very fast up to around 1.5 ms within the first 10 m
- Directional antennas do NOT influence coherence time at the first short distance, but they can constrain the fluctuation of coherence time versus distance.
- Very short coherence time requires channel estimation on slot level

Correlation of shadow fading

$$L(d) = \bar{L}(d_0) + 10n \lg\left(\frac{d}{d_0}\right) + X_\sigma$$



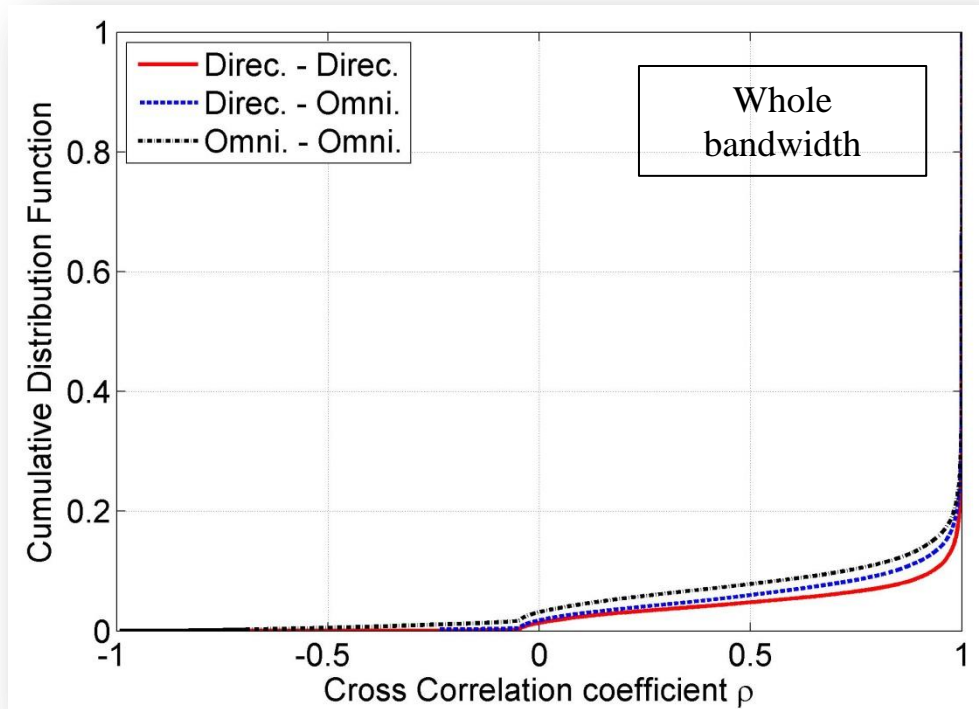
Decorrelation distance with threshold 0.5



| Decorrelation distance [m] | 50% | 90% |
|----------------------------|-------|-------|
| Direc. – Direc. | 2.104 | 3.807 |
| Direc. – Omni. | 1.904 | 3.406 |
| Omni. – Omni. | 1.703 | 3.206 |

- Directional antennas increase the decorrelation distance
- Mean decorrelation distance is around 2 m, MIMO antennas should separated further to get diversity gain

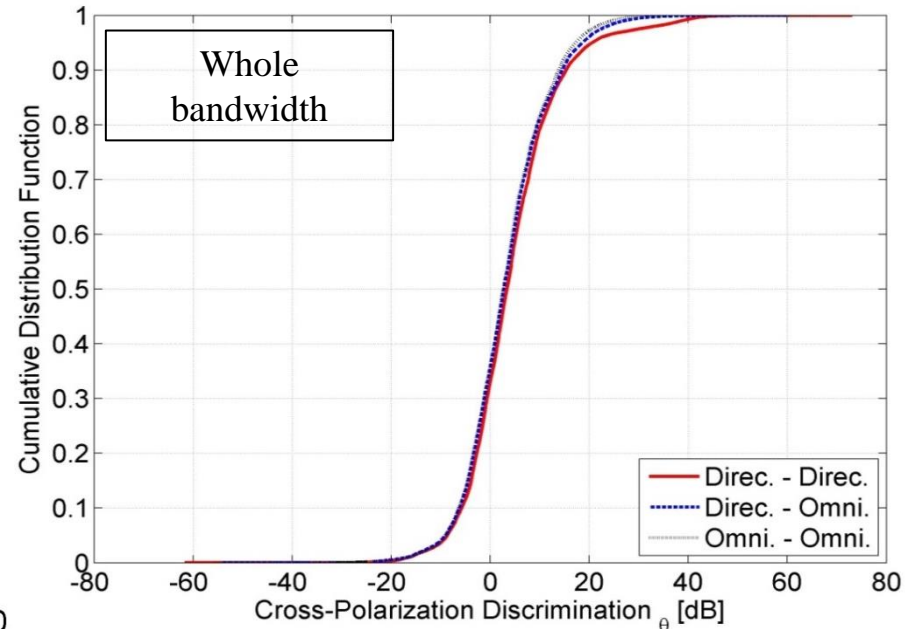
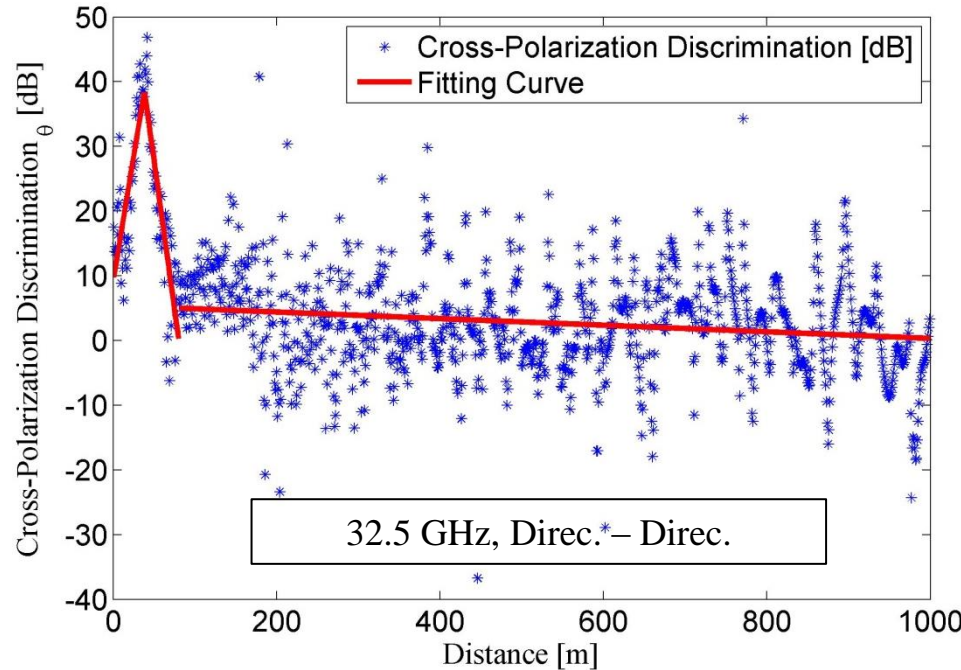
Cross correlation coefficient between two adjacent links



| Cross correlation coefficient | >0.5 | >0.9 |
|-------------------------------|-------|-------|
| Direc. – Direc. | 99.5% | 99.1% |
| Direc. – Omni. | 99.4% | 98.8% |
| Omni. – Omni. | 99.3% | 98.6% |

- **Cross correlation coefficient between two adjacent links with 10-cm separation is larger than 0.9 in 99% cases.**
- **Diversity gain can be expected by enlarging the separation between MIMO antennas**

Polarization – XPD_VH



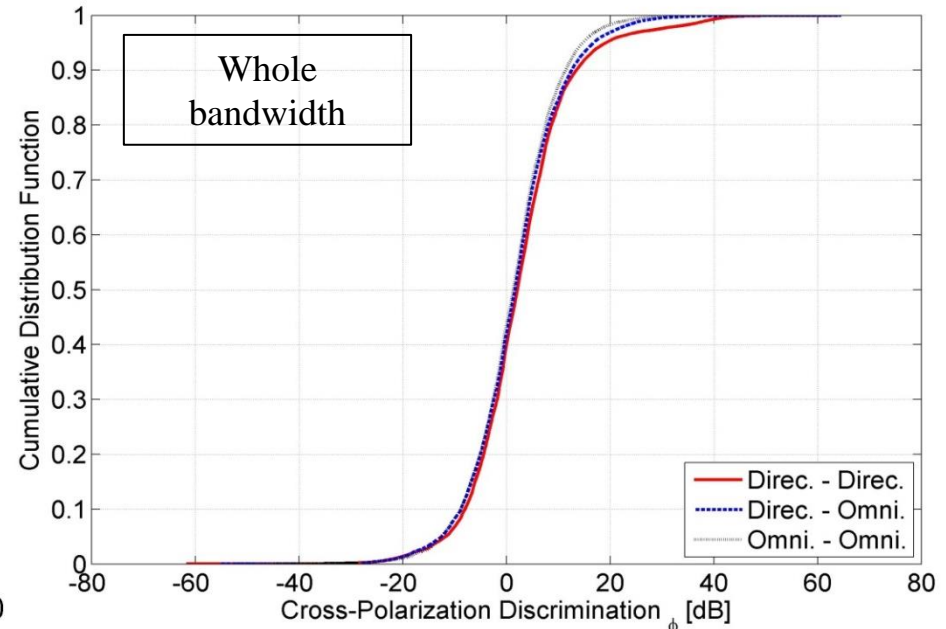
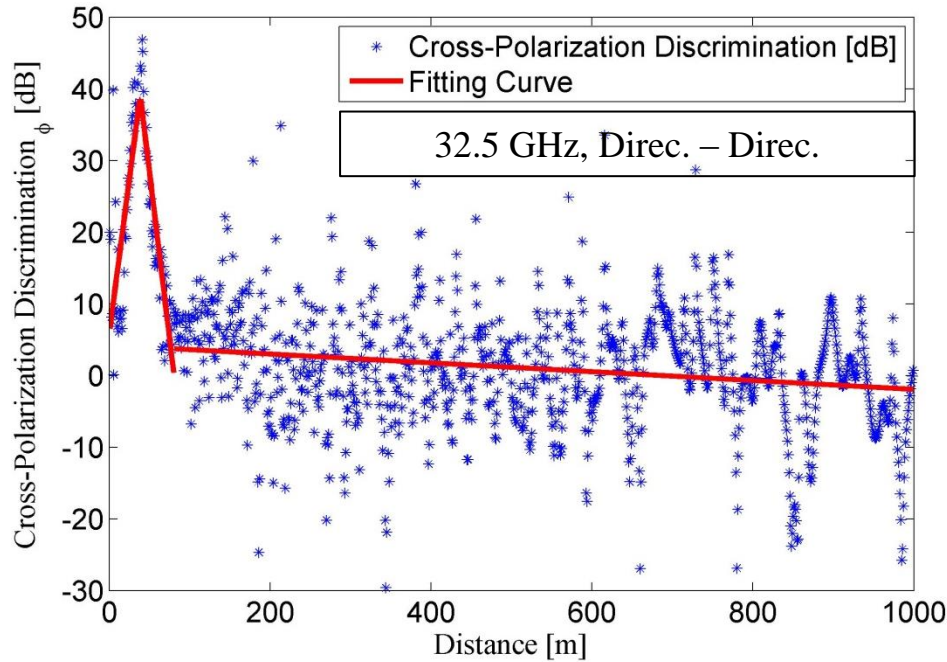
$$XPD_{\theta} = 20 \log_{10} \left(\frac{E_{Co}}{E_{Cross}} \right)$$

XPD_{θ} refers to the field received in θ co-polarization relative to the field transmitted in θ and received in ϕ polarization

| XPD $_{\theta}$ [dB] | 50% |
|----------------------|-------|
| Direc. – Direc. | 3.343 |
| Direc. – Omni. | 2.695 |
| Omni. – Omni. | 2.644 |

1/3 energy of the vertically polarized wave is depolarized

Polarization – XPD_HV



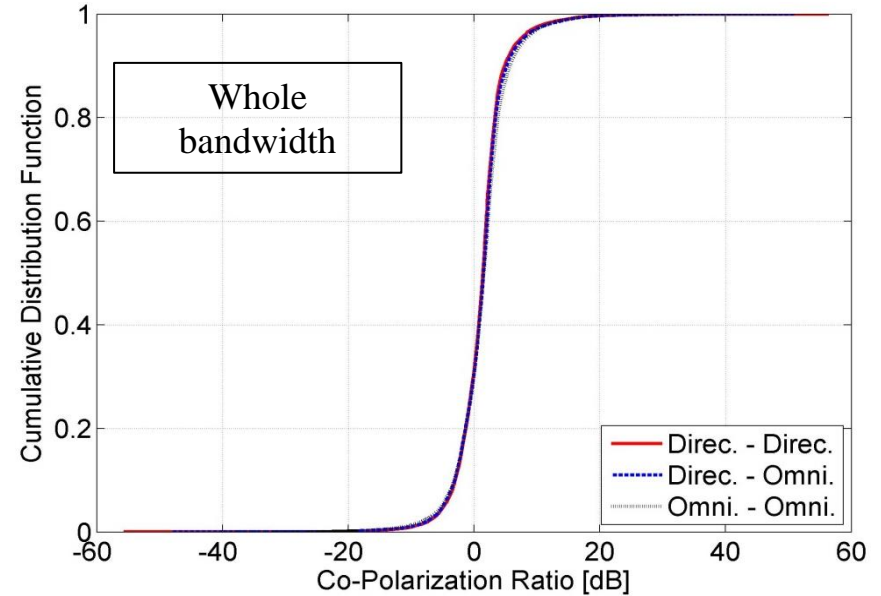
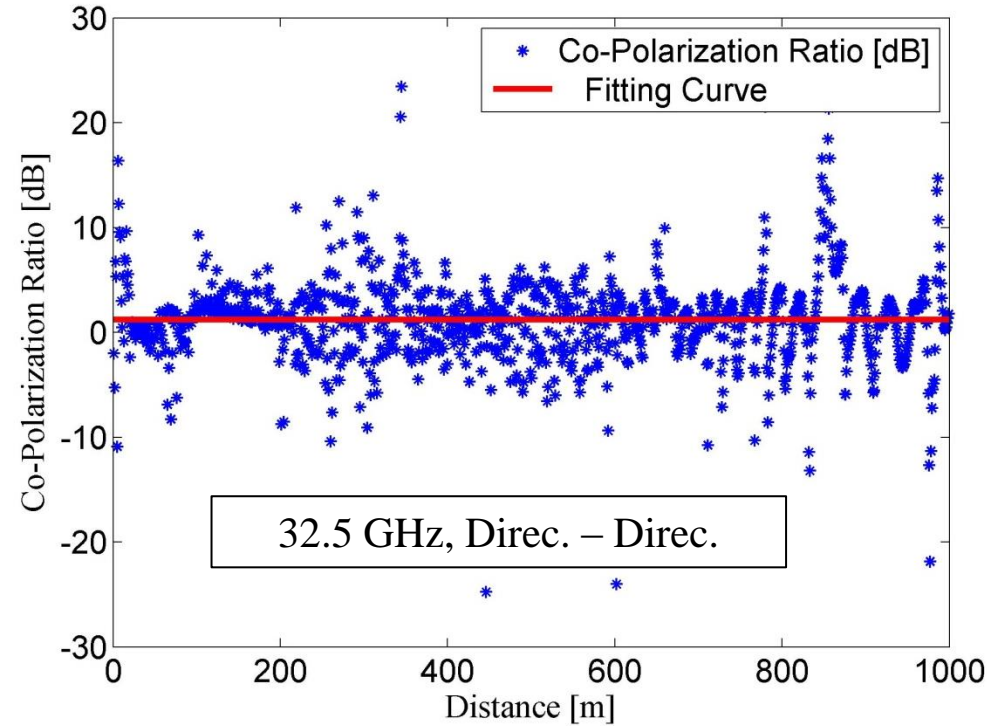
$$XPD_\phi = 20 \log_{10} \left(\frac{E_{Co}}{E_{Cross}} \right)$$

XPD_ϕ refers to the field received in ϕ co-polarization relative to the field transmitted in ϕ and received in θ polarization

| XPD_ϕ [dB] | 50% |
|-----------------|-------|
| Direc. – Direc. | 2.097 |
| Direc. – Omni. | 1.639 |
| Omni. – Omni. | 1.315 |

Half energy of the horizontally polarized wave is depolarized

Polarization – CPR



| Co-polarization Ratio [dB] | 50% |
|----------------------------|-------|
| Direc. – Direc. | 1.386 |
| Direc. – Omni. | 1.499 |
| Omni. – Omni. | 1.561 |

Vertical polarization is slightly better than horizontal polarization

The CPR, which describes the power ratio between the co-polarized V-channels (h_{VV}) and H-channels (h_{HH}), is defined:

$$CPR = \frac{E\{|h_{VV}|^2\}}{E\{|h_{HH}|^2\}}$$

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 - Frequency domain simulation
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- **Conclusion and future work**

Conclusion

Conclusion on channel characteristics

| | |
|-------------------------------|---|
| Received power | 1-km long coverage is promising by using dual-directional antenna setup |
| Rician K-factor | Very small and the channel suffers Rayleigh fading |
| RMS delay spread | Very short, no bother |
| RMS Doppler spread | Around 3 kHz in the first 20 m, reduced by directional antennas |
| Mean Doppler shift | Increase up to 12 kHz within the first 10 m |
| Coherence time | Much shorter than frame duration |
| Decorrelation distance | Around 2 m, much larger than 10 cm |
| Cross-correlation coefficient | Links separated only 10 cm are highly correlated |
| XPD | Around half energy is depolarized |
| CPR | Vertical polarization is slightly better |

Future work

- **Future work:**
 - More railway scenarios
 - More communication system setups
 - Stochastic modeling and channel realization



Rail traffic – an efficient and green transport model, playing a more and more significant role for human development!