Project: IEEE P802.15 Study Group for Wireless Personal Area Networks (WPANs)

Submission Title: PSSS proposal – Parallel reuse of 2.4 GHz PHY for the sub-1-GHz bands

Date Submitted: 17 November 2004

Source: Andreas Wolf, DWA Wireless GmbH and Hans van Leeuwen, STS-wireless

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Re: Analysis of PSSS for higher data rates for PHY for sub-1-GHz

Abstract: The proposed parallel reuse of the 2.4 GHz 802.15.4 modulation technology in PSSS offers highly attractive performance improvement, fulfills all key OEM requirements, and visibly increases market opportunities.

Purpose: Further analysis of PSSS as in accepted joint PHY proposal from September 2004

PSSS Proposal

Parallel reuse of 2.4 GHz PHY for the sub-1-GHz bands

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- PSSS Implementation aspects
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  - PSD
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- Status
- Summary

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  - PSSS PHY Tx operation
  - Selected Rx implementation options
  - Linearity

Key requirements for sub-1-GHz band PHY

- **Bitrate over 200 kBit/s**
  - Number of permitted transactions/hr is insuffcient in IEEE802.15.4-2003 868 Mhz
    - 1% duty cycle at 20 kbit/s translates into typically only 600-800 transactions/hr
    - With > 200 kbit/s sufficient number of transactions/hr for our targeted applications
    - Disadvantage of 1% duty cycle limit turns into *protection against interference*
  - Extension from 20/40 kbit/s extends total battery lifetime by 15-40%
- **Visibly improved multipath fading robustness over IEEE802.15.4-2003 2.4 GHz**
  - Improve coverage in "challenging" RF environments – Especially commercial, industrial
  - Achieve PER < 10^{-3} at channels with at least 1 μs delay spread (non-exponential channel models)
- **Support of current RF regulatory regimes *plus* enable the use of extended bands**
  - Support 2 MHz wide channels in the USA and other countries were they are permitted
  - Support of current 600 kHz band available at 1% duty cycle in Europe today
  - Allow use of extended European bands and bands in other countries once they become available
    - Allow addition of additional 600 kHz channels as per current ETSI / ECC report (4/6 channels?)
    - Do not expect US-like wide, unrestricted bands or all egulatory domains
  - Support of more flexible channel selection method to flexibly add support for more countries
- **Backward compatibility to IEEE802.15.4-2003 (915/868 MHz)**
  - Interoperability when switched to 15.4-2003 mode
  - No fully transparent backward compatibility as in 802.11b vs. 802.11 or 802.11g vs. 802.11b
- **Low cost and low power consumption (!)**

Source: Danfoss IEEE 15-04-327-01-004b; TG4b discussion in September 2004 meeting

PSSS variants reviewed in this presentation

	PSSS 234-600	PSSS 225-600	PSSS 210-600	PSSS 250-600 a/b	PSSS 250-2000
Bandwidth	600 kHz	600 kHz	600 kHz	600 kHz	2,000 kHz
Chiprate	500 cps	480 cps	450 cps	266.6 / 400 cps	800 kcps
Bitrate	234 kit/s	225 kbit/s	210 kbit/s	250 kbit/s	250 kbit/s
Spectral efficiency	15/32 bit/s/Hz	15/32 bit/s/Hz	15/32 bit/s/Hz	0.9375 / 0.625 bit/s/Hz (30/32; 20/32)	0.3125 bit/s/Hz (10/32)
Spreading	15x 32-chip seq.	15x 32-chip seq.	15x 32-chip seq.	10x 32/15x32-complex chip seq.	5 x 32 complex chip seq.
RF backward compatibility	Single BPSK / ASK radio	Single BPSK / ASK radio	Single BPSK / ASK radio	IQ radio	IQ radio
Comments	Original mode in joint proposal	Added upon TG4b request to have "more even" bitrate	Added upon chip manufacturer input to reduce complexity / costs	Added as variant based on I/Q modulator + low cost 250 kbit/s in 600 KHz	Added as variant to show that use of PSSS is also attractive in 2 MHz channels

Choice to be discussed in TG4b

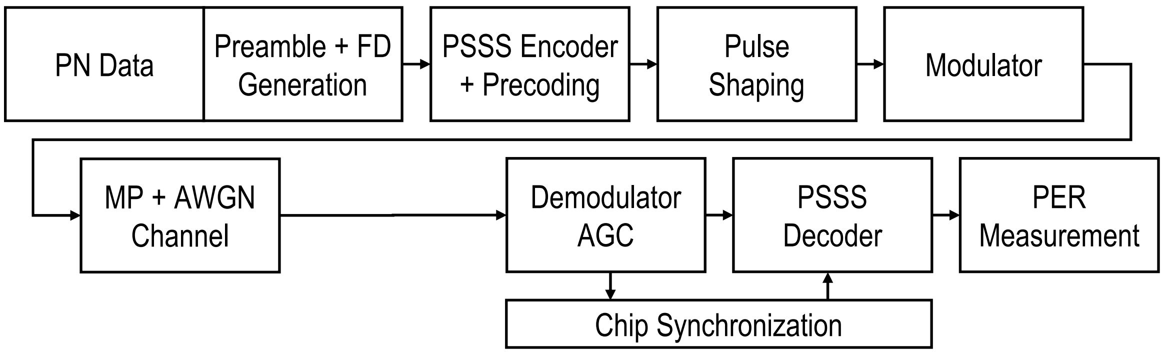
Note:
DWA fully supports the accepted joint proposal - variants are provided to provide a more comprehensive analysis

Challenges in comparison of PHY variants in TG4b PHY subcommittee

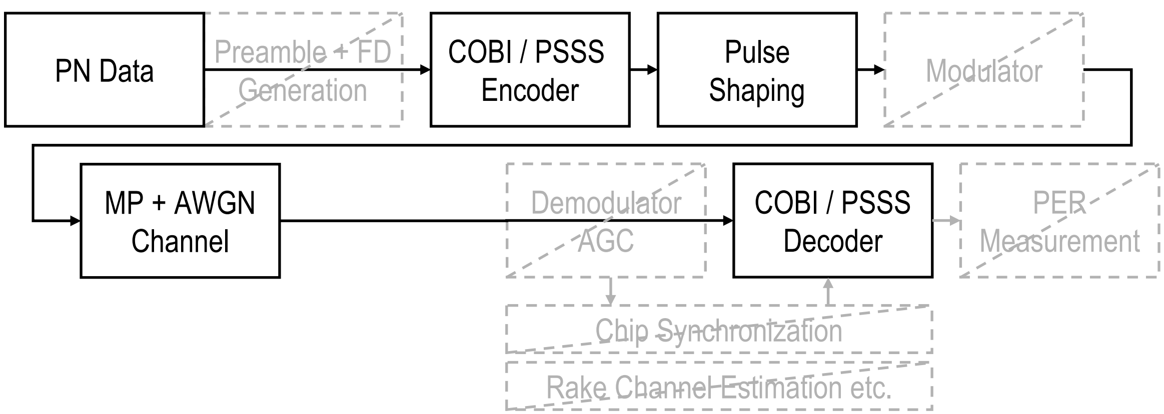
- Uneven level of analysis and scrutiny between PSSS and COBI
  - Despite major deviation from IEEE802.15.4-2003 2.4 Ghz design, many implementation challenges are not yet reviewed for COBI, e.g. synchronization, PSD, required linearity, Rake receiver
- Current COBI simulations discussed are not suitable to drive conclusions
  - Limited, incomplete simulation model – e.g. without preamble, synchronization
  - Critical parts of Rake receiver are not simulated (furthermore, experience is that even full Rake simulations deviate significantly from actual implementations – commonly accepted in scientific literature)
  - Switch from agreed comparison of PER to BER (focus on irrelevant BER values)
  - COBI8 variants shown *cannot* fulfill ETSI spectrum mask (Nyquist)
- Unclear PSSS simulations from IIR
  - Results from September 2004 and now are inconsistent
  - PSSS without precoding is shown with lower performance than with precoding
  - PSSS is shown with unnecessary Rake receivers driving irrelevant and misleading conclusions

Simulation models used

Simulation model used by DWA



Simulation model used by IIR in TG4b PHY discussions

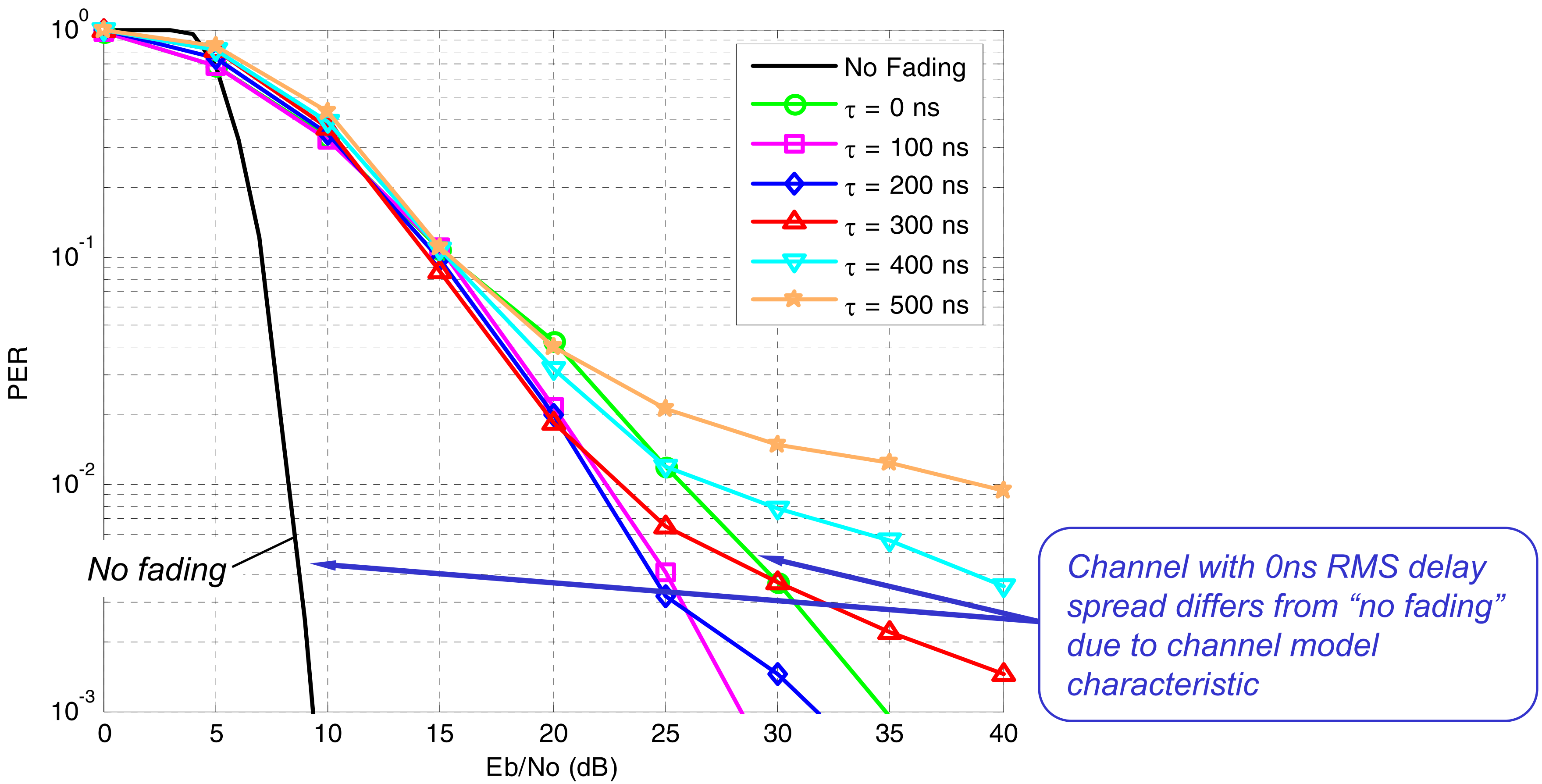


- Agreed simulation model used by DWA:

  Discrete exponential model
  - Sampled version of diffuse model (high sampling rate)
  - At least 1000 random channel realizations
  - **PER** calculated on complete PPDUs with preamble and FD

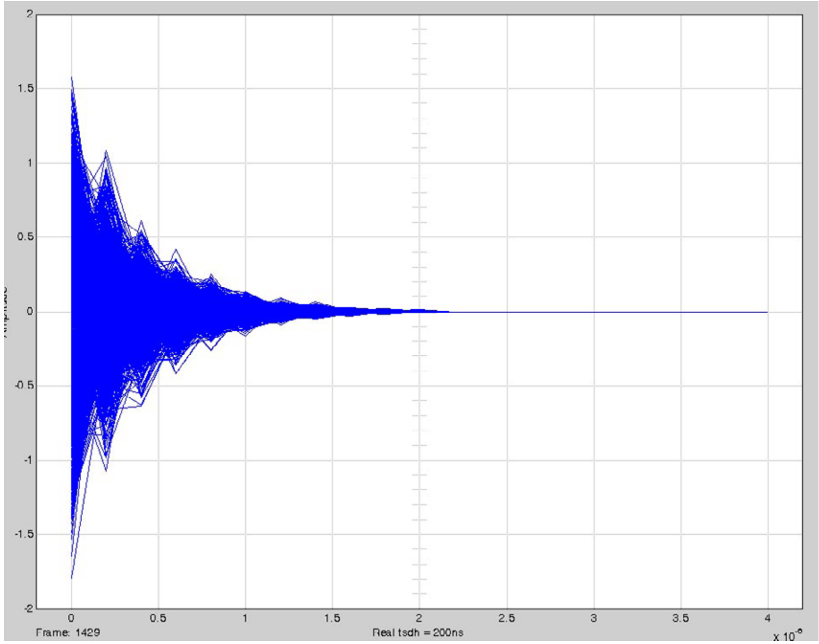
- Notes:
  - Results shown by IIR for COBI8 are based on model with PSD that violates ETSI
  - **BER** of only 10^{-3} / 10^{-4} shown is insufficient for target market – ***PER*** of 10^{-3} is typically used in IEEE802
  - COBI Rake receiver structure unclear
  - Preamble proposed by IIR for COBI16/8 is inappropriate for use with rake (i.e. too short for accurate channel estimation)
  - Is preamble proposed sufficient for other COBI modes?
  - **Rake receiver requires higher accuracy for AGC and linearity. Effects have to be investigated.**

Earlier results of basic model also used by IIR

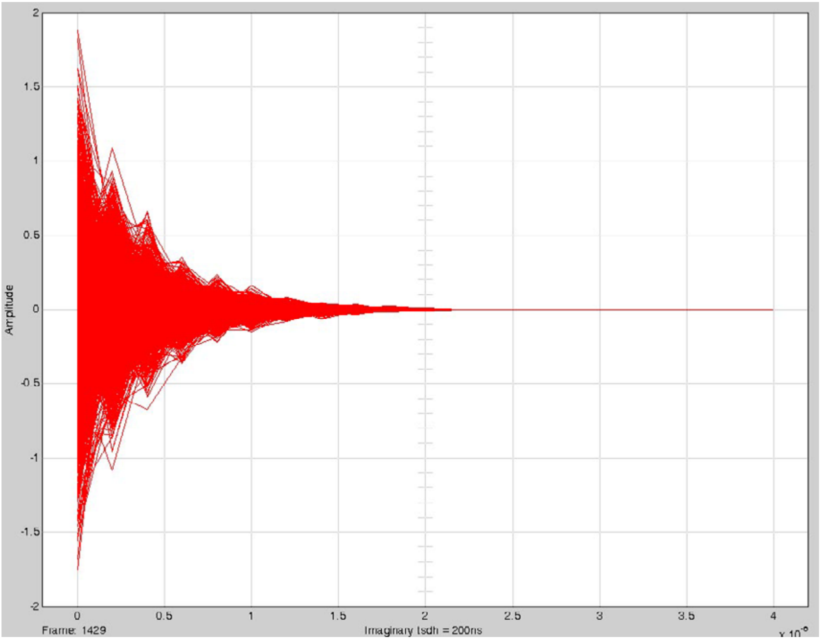


Source Halfrate 2.4 GHz: IEEE 15-04-337-00-004b, Motorola, slide 6

Channel Reponse – Simulation of 1429 Frames used by DWA



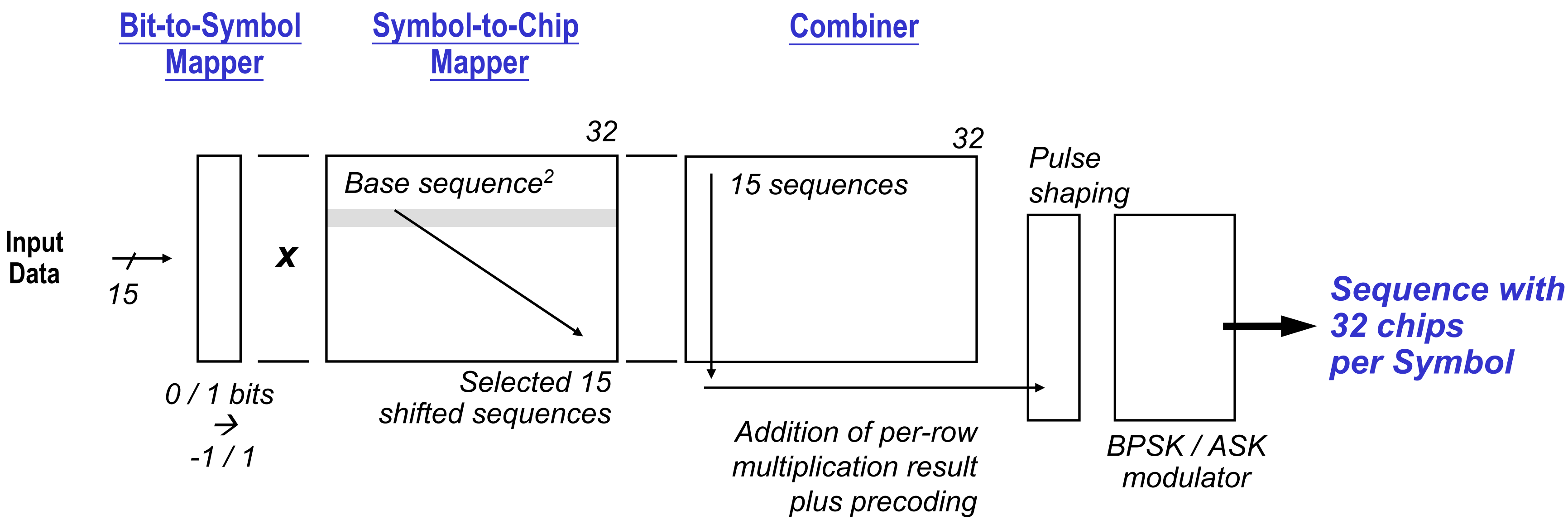
Real Part



Imaginary Part

Note:
Actual channels in industrial and commercial environments are having significantly higher probability for non-exponential amplitude/time than assumed in the agreed and used model

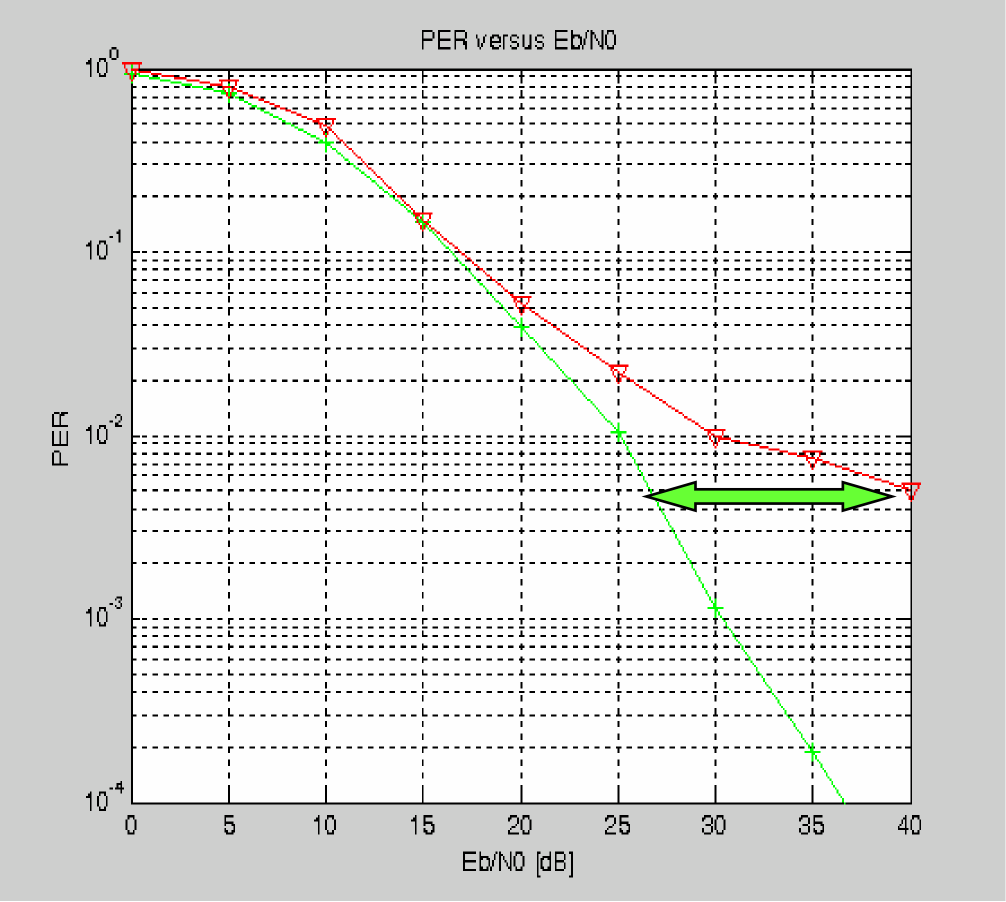
PSSS – BPSK/ASK variant[1] (15/32 bit/s/Hz) simulated



...addition of multiple parallel sequences instead of selection of single sequence

1: PSSS 225-600 + PSSS 210-600
2: Use of single base sequence simplifies implementation in Rx

PER Performance PSSS BPSK/ASK variant – Discrete Exponential Channel, 370ns RMS Delay Spread

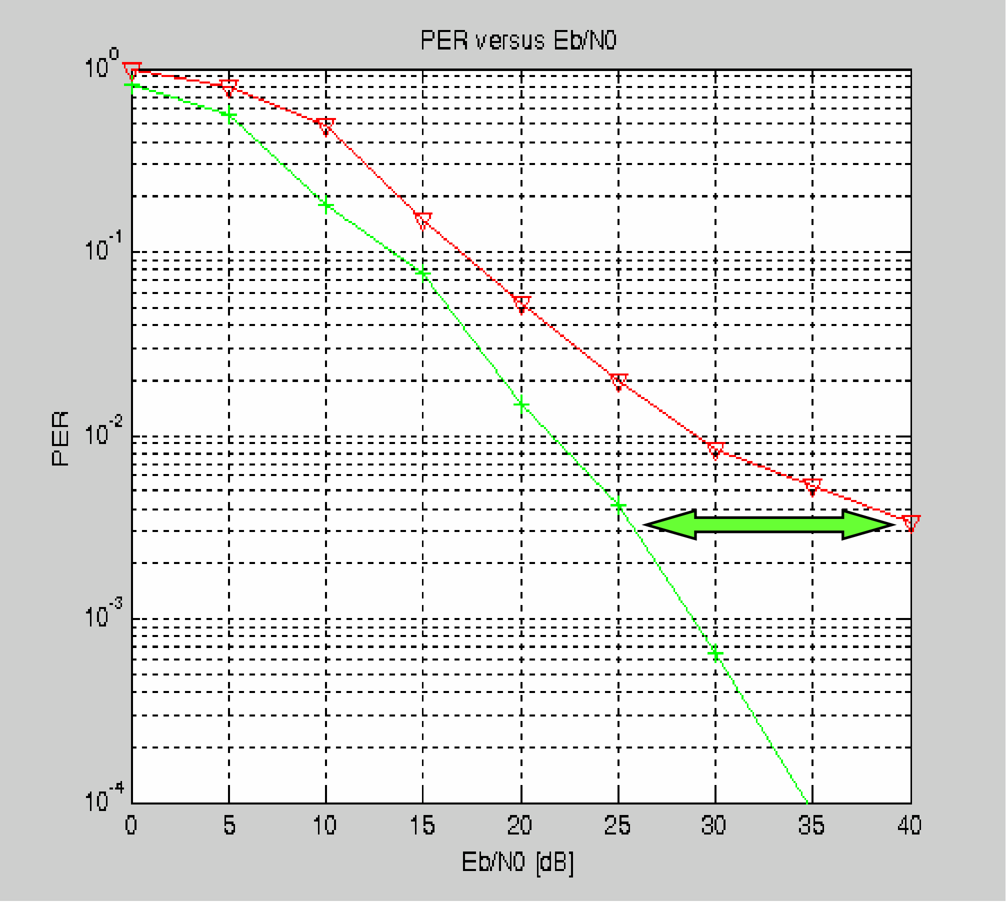


- Over 12 dB performance benefit in relevant PER range
  - Even higher benefit in environments with higher MP fading challenges
- COBI8 performance is estimated to be 4...7dB weaker than even COBI16
  - Little if any performance benefit over 868MHz FSK chips

PSSS fulfills performance requirements without adding complexity, cost, and power consumption for rake receivers

– PSSS 206 kbit/s – COBI16+1 235 kbit/s *> 10000 Channel, no Rake receivers*

PER Performance PSSS BPSK/ASK variant – Discrete Exponential Channel, 250ns RMS Delay Spread



- Over 14 dB performance benefit in relevant PER range
  - Even higher benefit in environments with higher MP fading challenges
- COBI8 performance is estimated to be 4...7dB weaker than even COBI16
  - Little if any performance benefit over 868MHz FSK chips

PSSS fulfills performance requirements without adding complexity, cost, and power consumption for rake receivers

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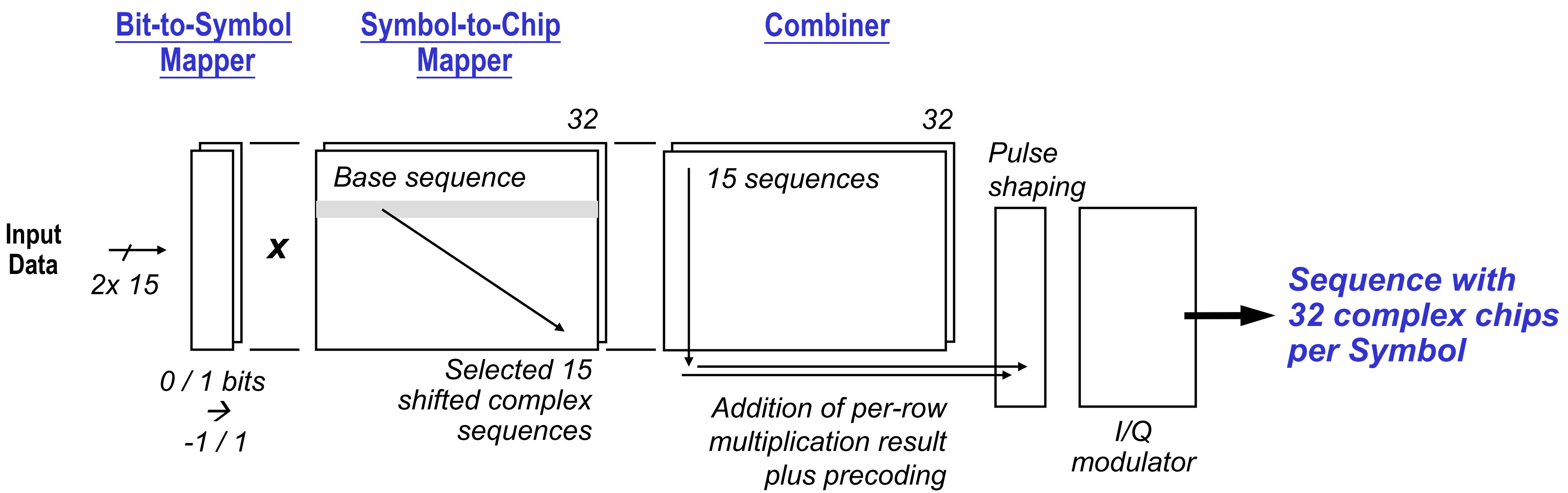
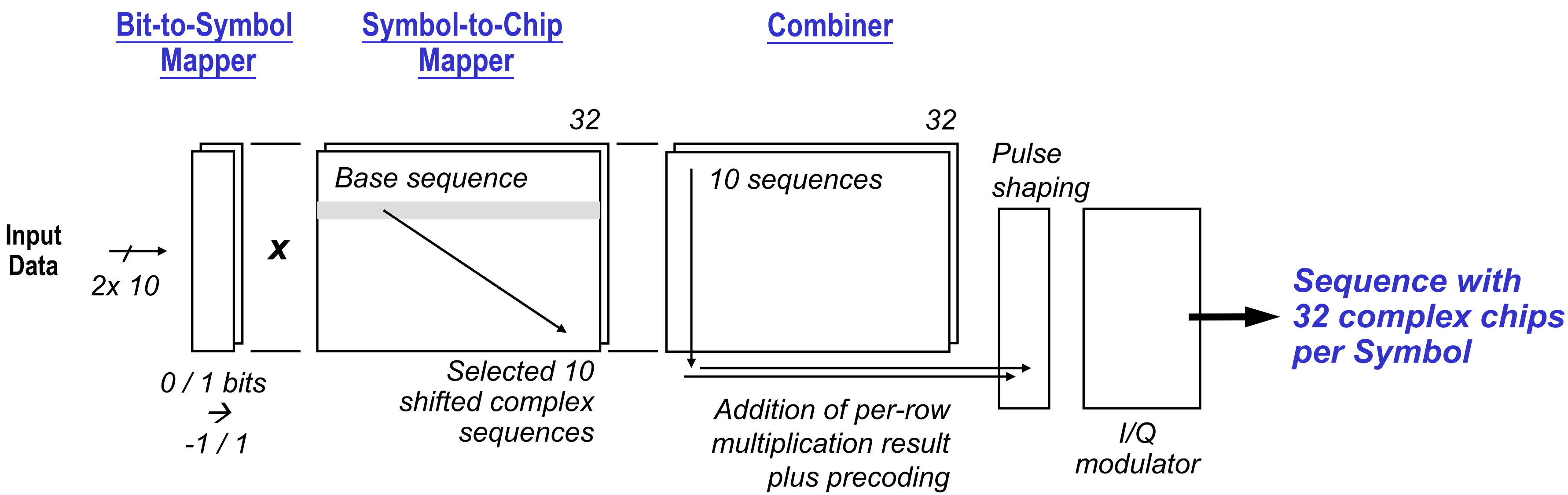
PSSS – 250 kbit/s I/Q variant 1 (IQ1) simulated[1]



... simplest pulse shaping enabling very low cost implementation

1: PSSS 250-600a

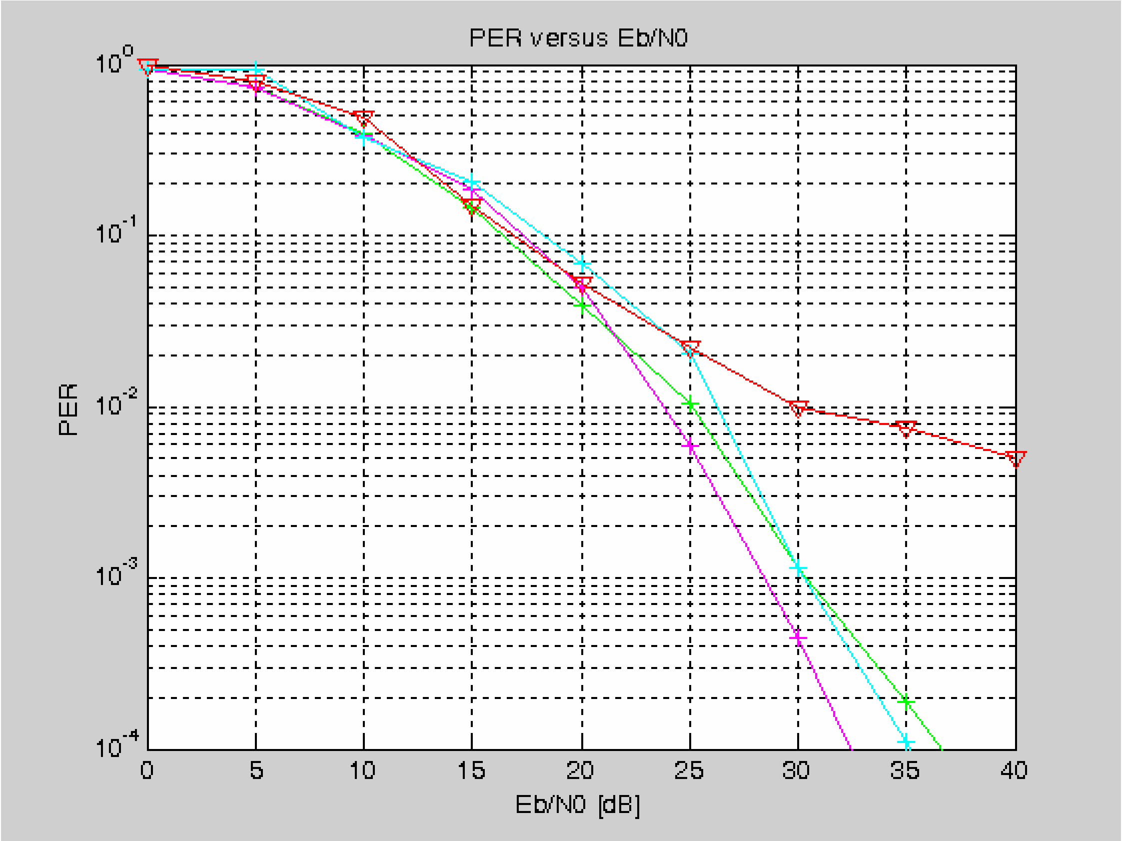
PSSS – 250 kbit/s I/Q variant 2 (IQ2) simulated[1]



... enables reuse of chip designs with I/Q modulator / demodulator

1: PSSS 250-600b

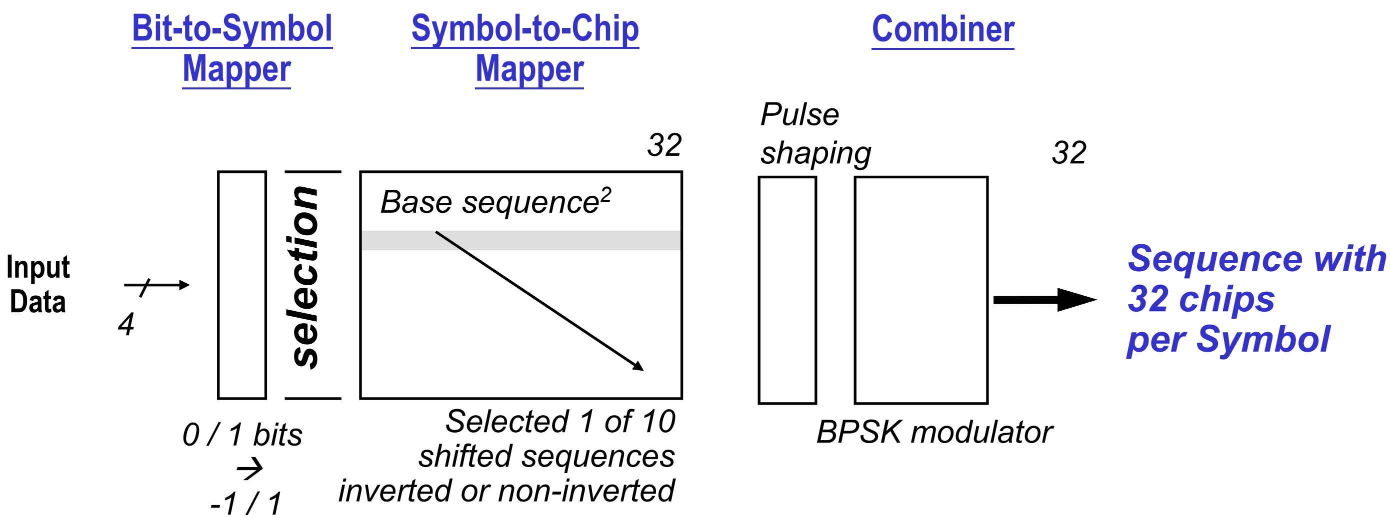
PER Performance PSSS IQ variants – Discrete Exponential Channel, 370ns RMS Delay Spread



Similar and even higher benefit over COBI16

- PSSS 225 kbit/s
- COBI16+1 coherent, 235 kbit/s
- PSSS IQ1 (250-600a)
- PSSS IQ2 (250-600b)

PSSS – BPSK variant[1] (4/32 bit/s/Hz) simulated 900 MHz



1: PSSS 250-2000
2: Use of single base sequence simplifies implementation in Rx

Bit to Symbol and Symbol to Chip Mapping PSSS 250-2000

Data	Sequence number
0000	0
0001	1
0010	2
0011	3
0100	4
0101	5
0110	6
0111	7
1000	8
1001	9
1010	0
1011	1
1100	2
1101	3
1110	4
1111	5

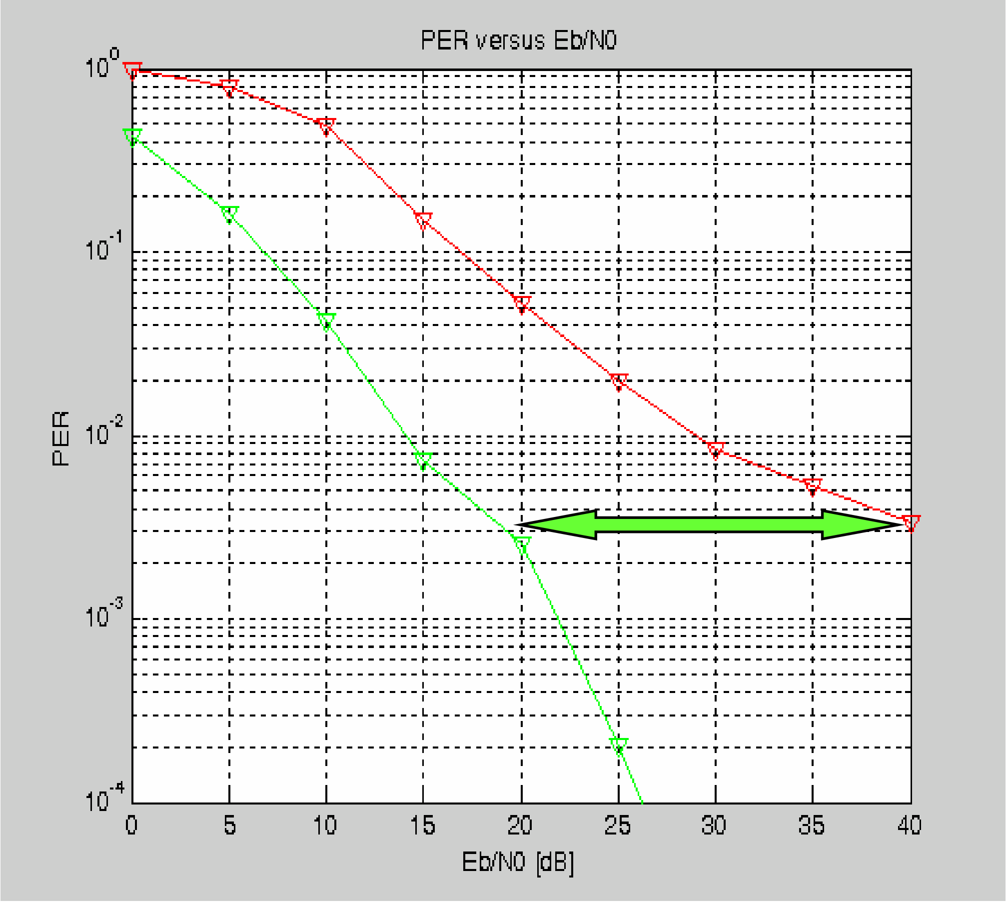
bit to symbol

Sequence number	Chip number 0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	24	25	26	27	28	29	30	31
0	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1
1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1
2	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1
3	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1
4	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1
5	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1
6	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1
7	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1
8	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1
9	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1

symbol to chip

underlined means to use the sequence inverted

PER Performance PSSS 250-2000 BPSK variant – Discrete Exponential Channel, 250ns RMS Delay Spread



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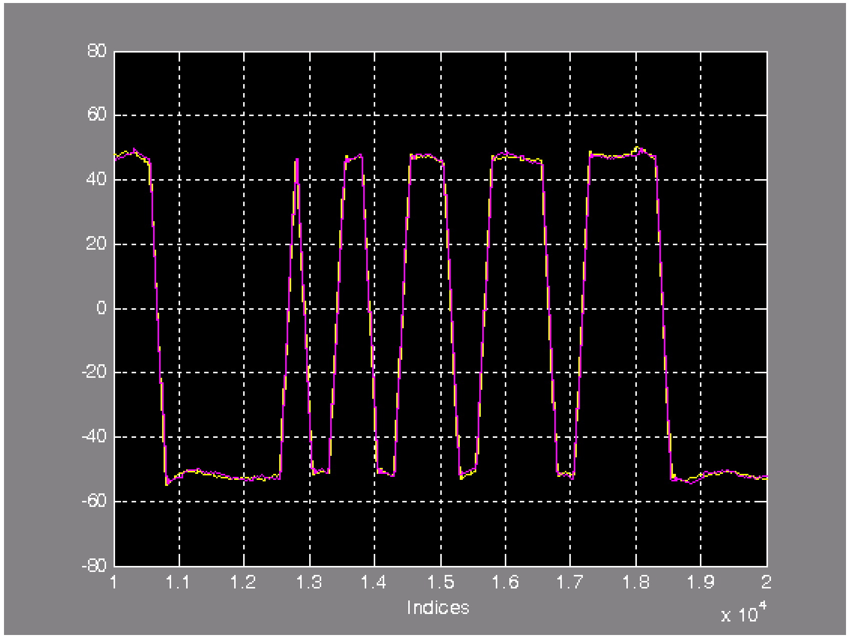
Crystal quality – Tolerated frequency offset

- Performance against frequency offset –
  Original target in TG4: Up to ±40ppm
  - Assumptions for chip clock:
    - PDU length 127 Byte = 8*127 bit = 1016 bit
    - 15 bit per PSSS Symbol (32 chip)
    - → 68 PSSS Symbols with 2176 chips (Chip duration Tc= 2µs)
  - Results
    - 40ppm for 2176 chips = 0.087 chip error for the whole PDU
    - For one PSSS Symbol with 32 chips
      the error is about 40ppm*32 chip = 0,00128 chip

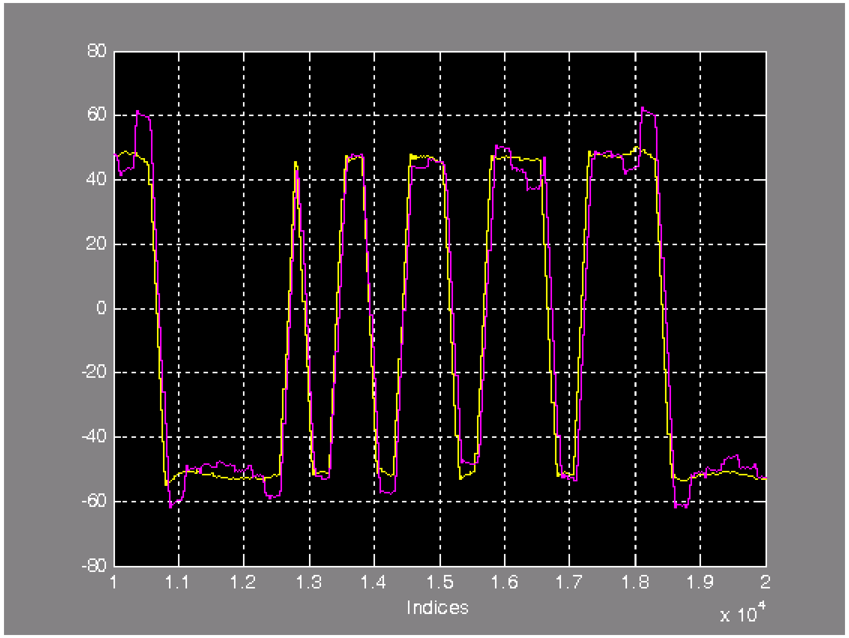
➡ No influence to PSSS Performance by ±40ppm and worse crystal

Crystal quality – Tolerated frequency offset – Measurements from PSSS prototype

0.1% Chip Clock Error



1% Chip Clock Error

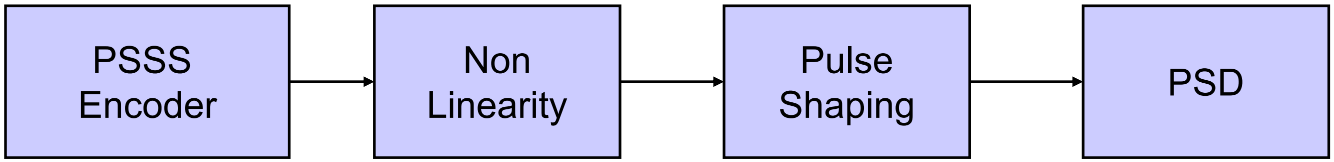


Yellow: 0% chip clock error reference signal
Pink: 0.1% and 1% chip clock error

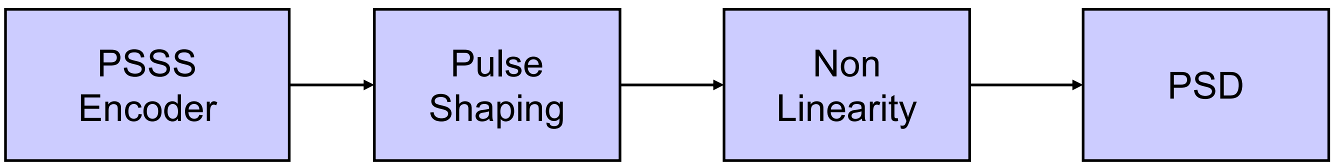
Calculation of crystal quality tolerance confirmed with prototype

Simulation models used for pulse shaping

Passband pulse shaping model



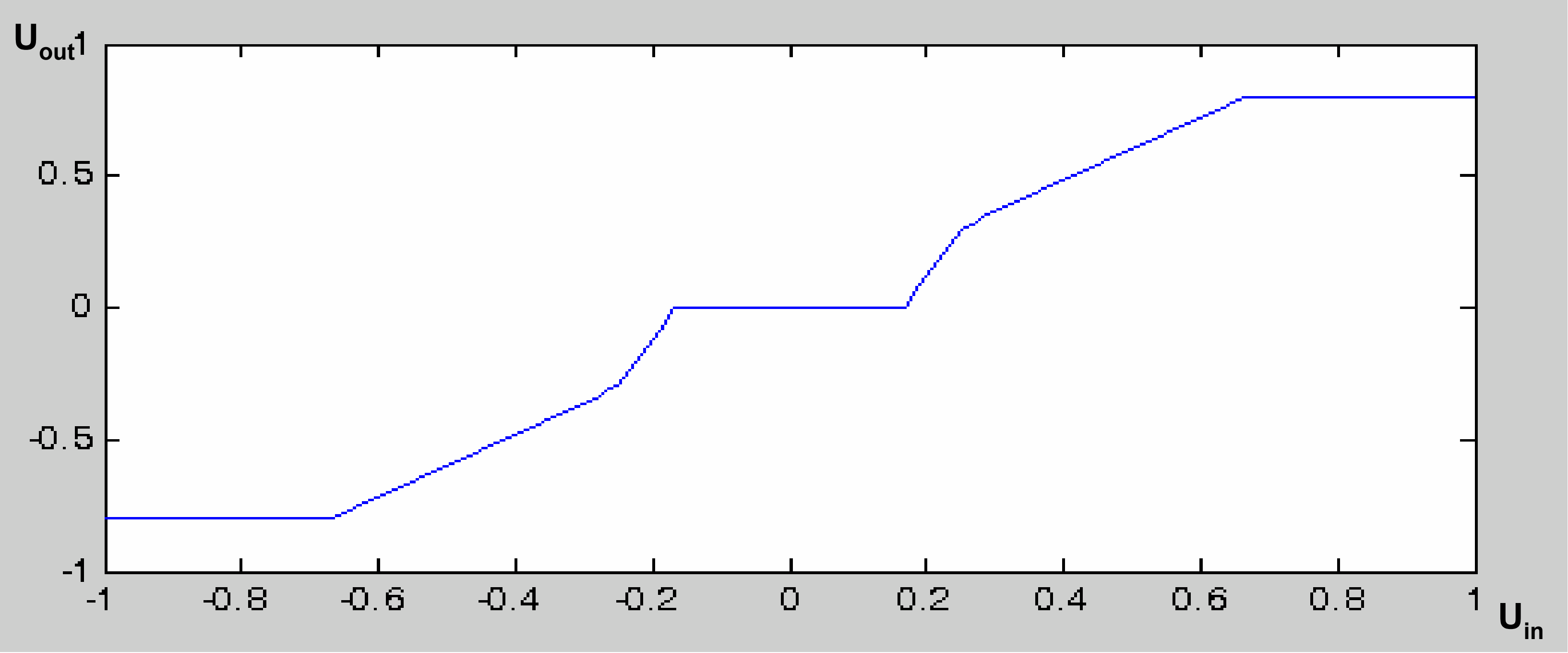
Baseband pulse shaping model



Notes:

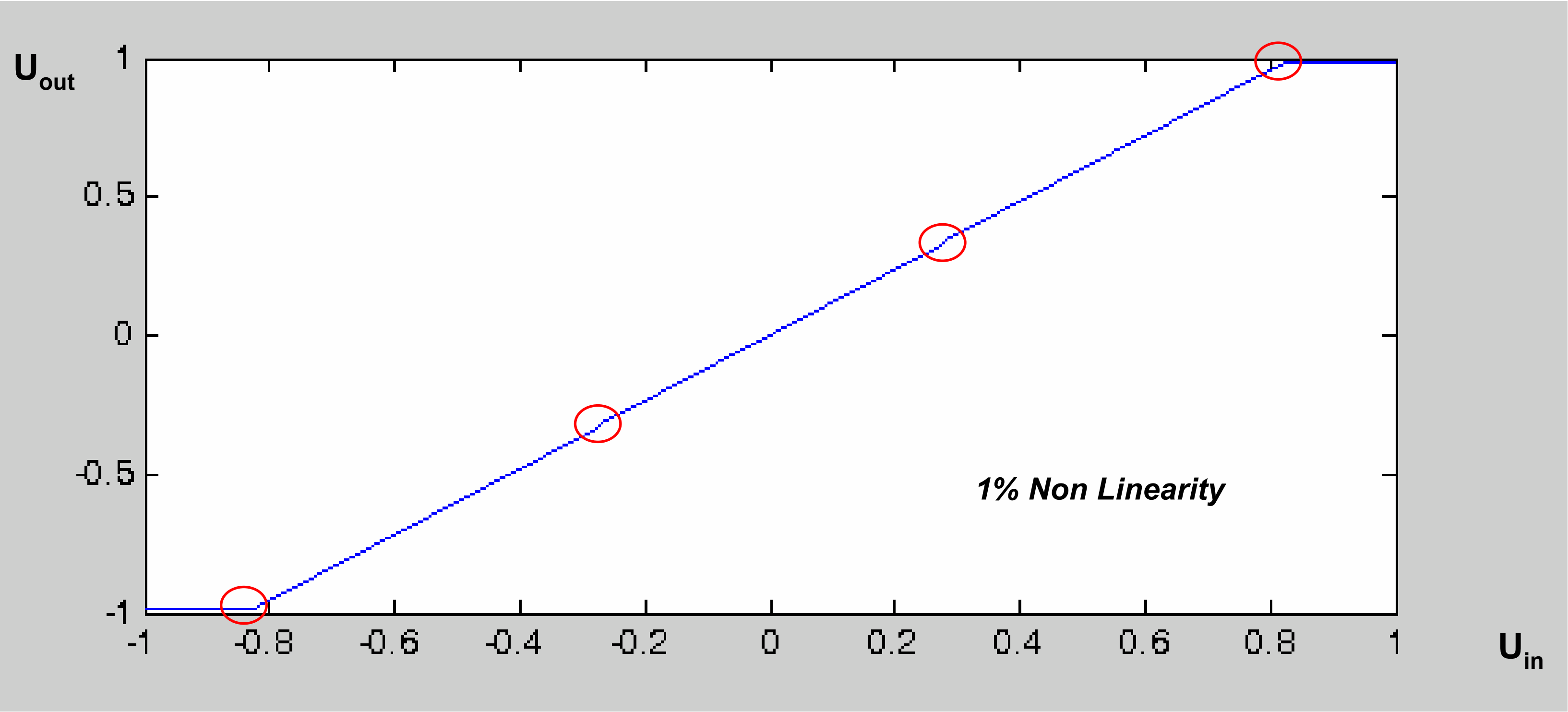
- **Pulse shaping per draft specification text provided submitted by DWA**
- **Details of models conformant to ETSI recommendations**
- **Actual bandwidth for PSD 16 kHz simulation**
- **PSSS: Square root raised cosine filter r=0.1**
  - **Theoretical limit r=0.2**
- **ETSI power limits are absolute +14 dBm inband, -36 dBm outband**
  - **For simulation assumed to send with max. power +14 dBm**
  - **Therefore simulation results contain relative PSD levels**
    - **+14 dBm -> 0 dB**
    - **-36 dBm -> -50 dB**

Non Linear Transfer Function – Passband pulse shaping



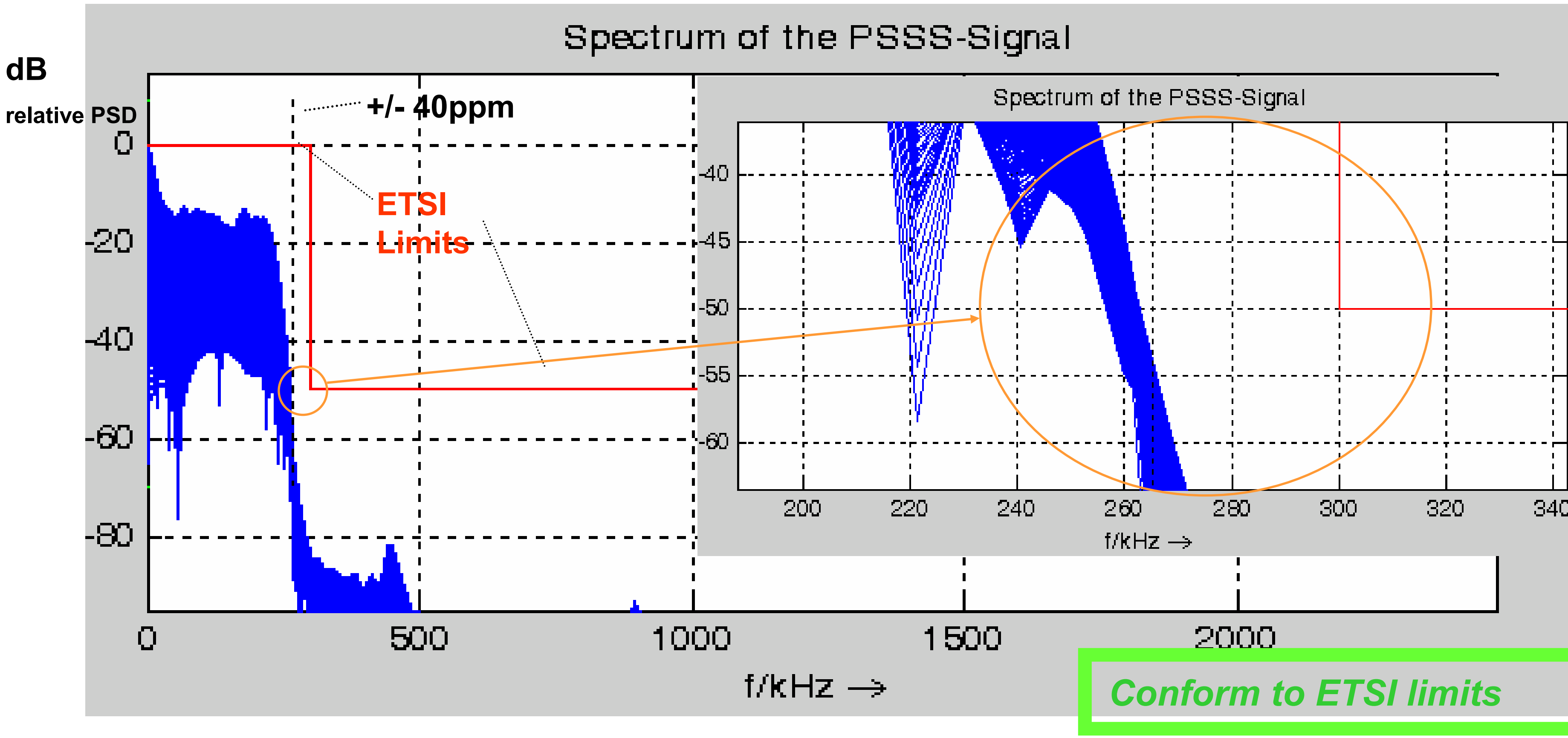
Used transfer function for simulating PSD for non linearity

Non Linear Transfer Function – Baseband pulse shaping



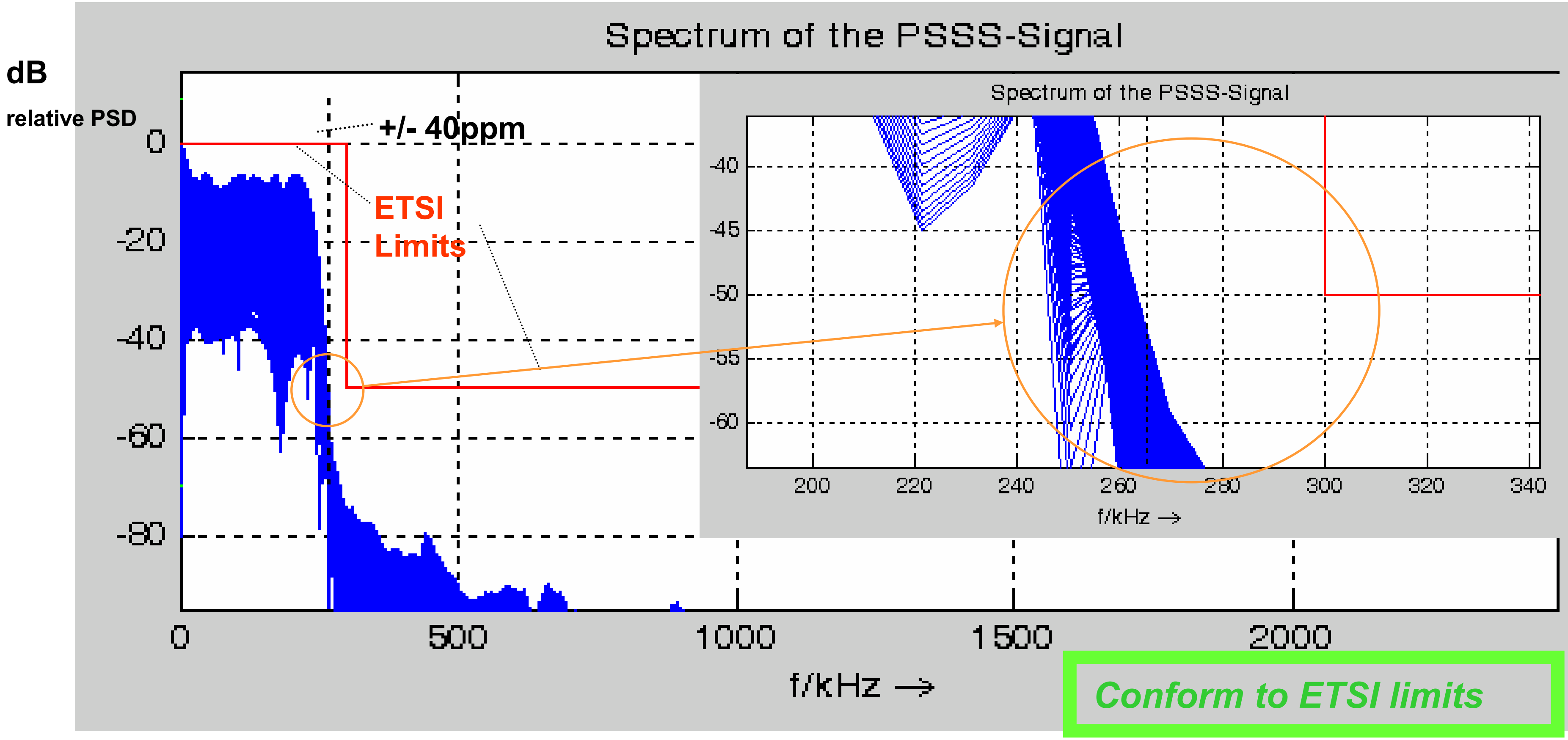
Used transfer function for simulating PSD for non linearity

PSD PSSS Signal – Passband pulse shaping, linear, no precoding



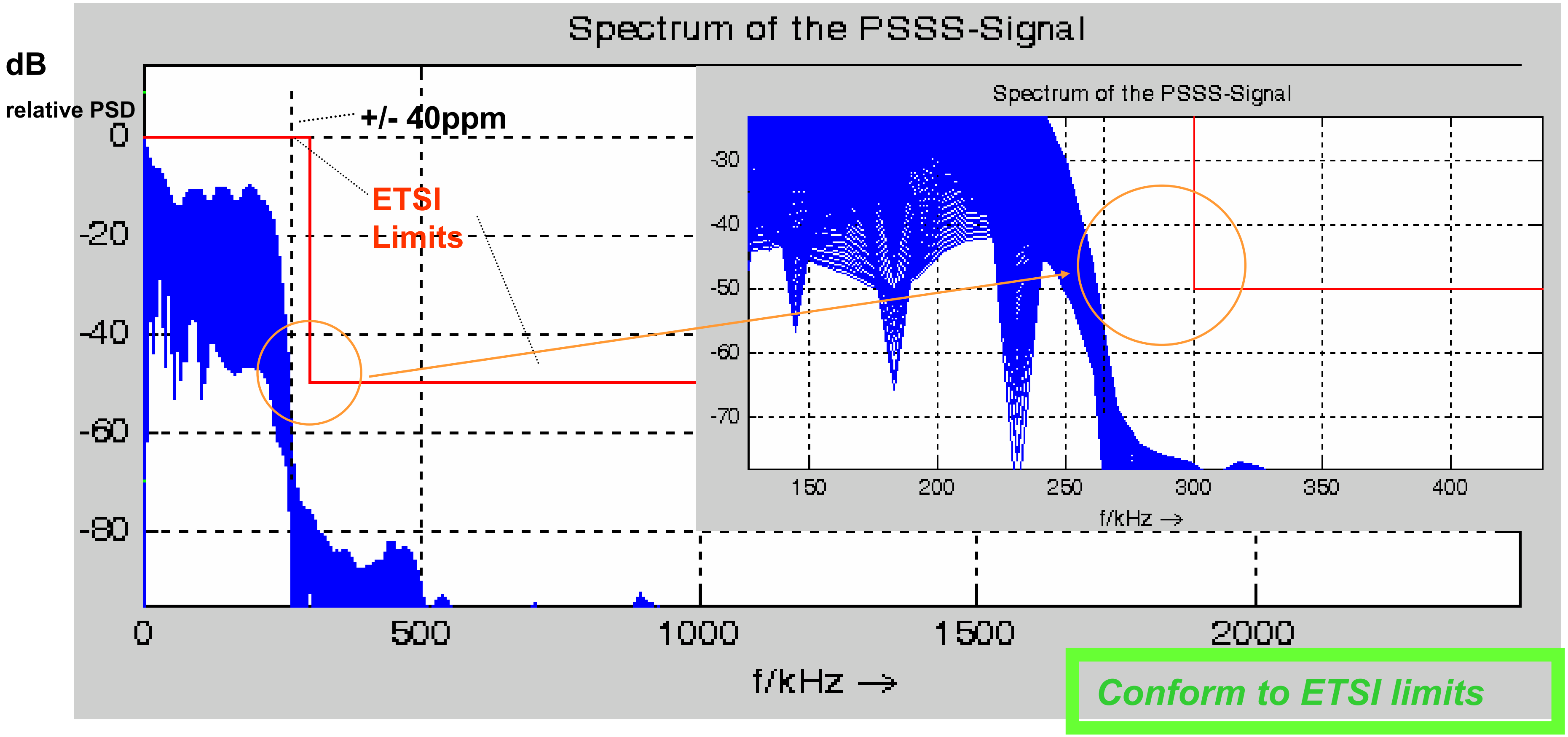
Simulations of the relative PSD in dB for the PSSS signal at 450 kchips/s, 210 kbit/s, +/- 40ppm.

PSD PSSS Signal – Passband pulse shaping, linear, precoding



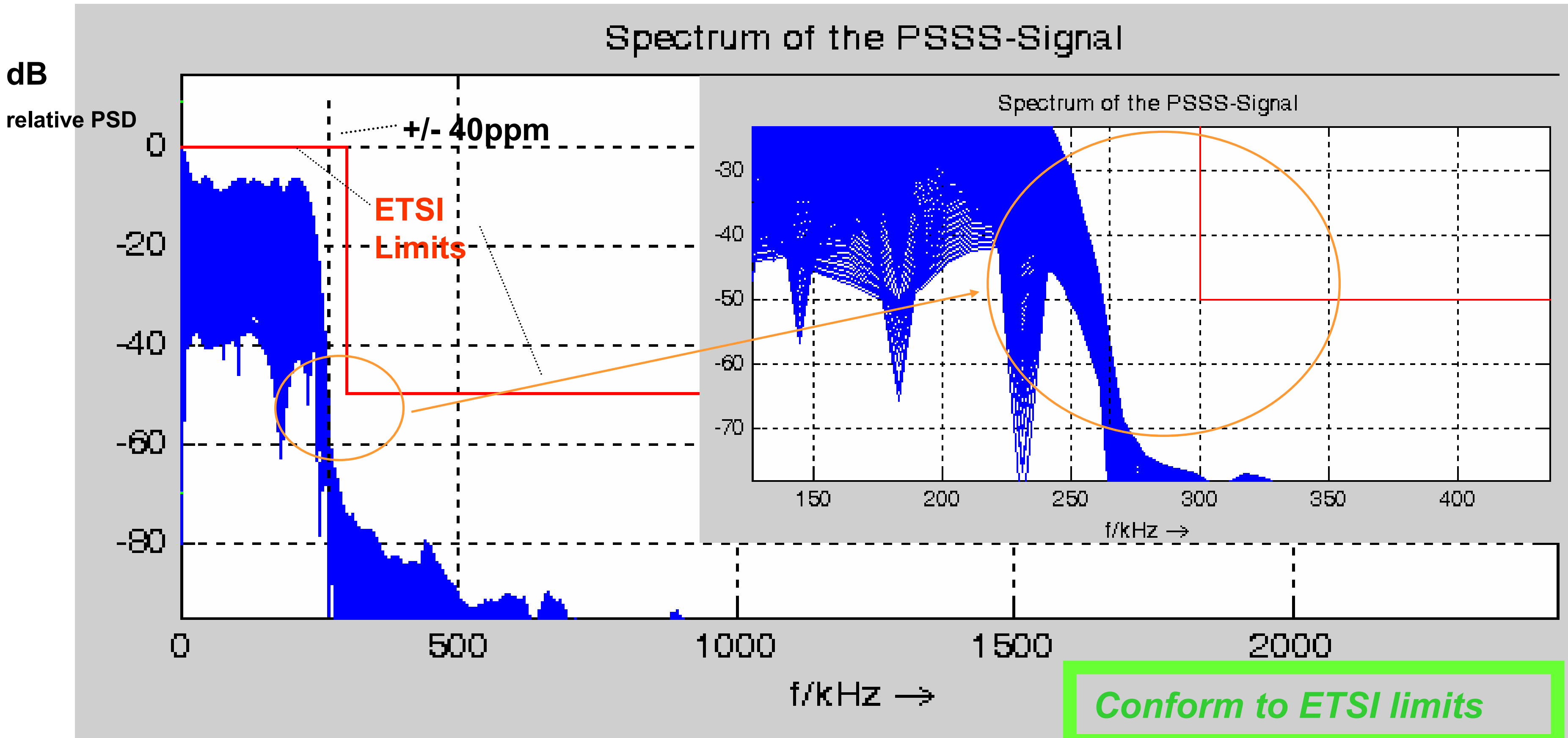
Simulations of the relative PSD in dB for the PSSS signal at 450 kchips/s, 210 kbit/s, +/- 40ppm.

PSD PSSS Signal – Passband pulse shaping, non linear, no precoding



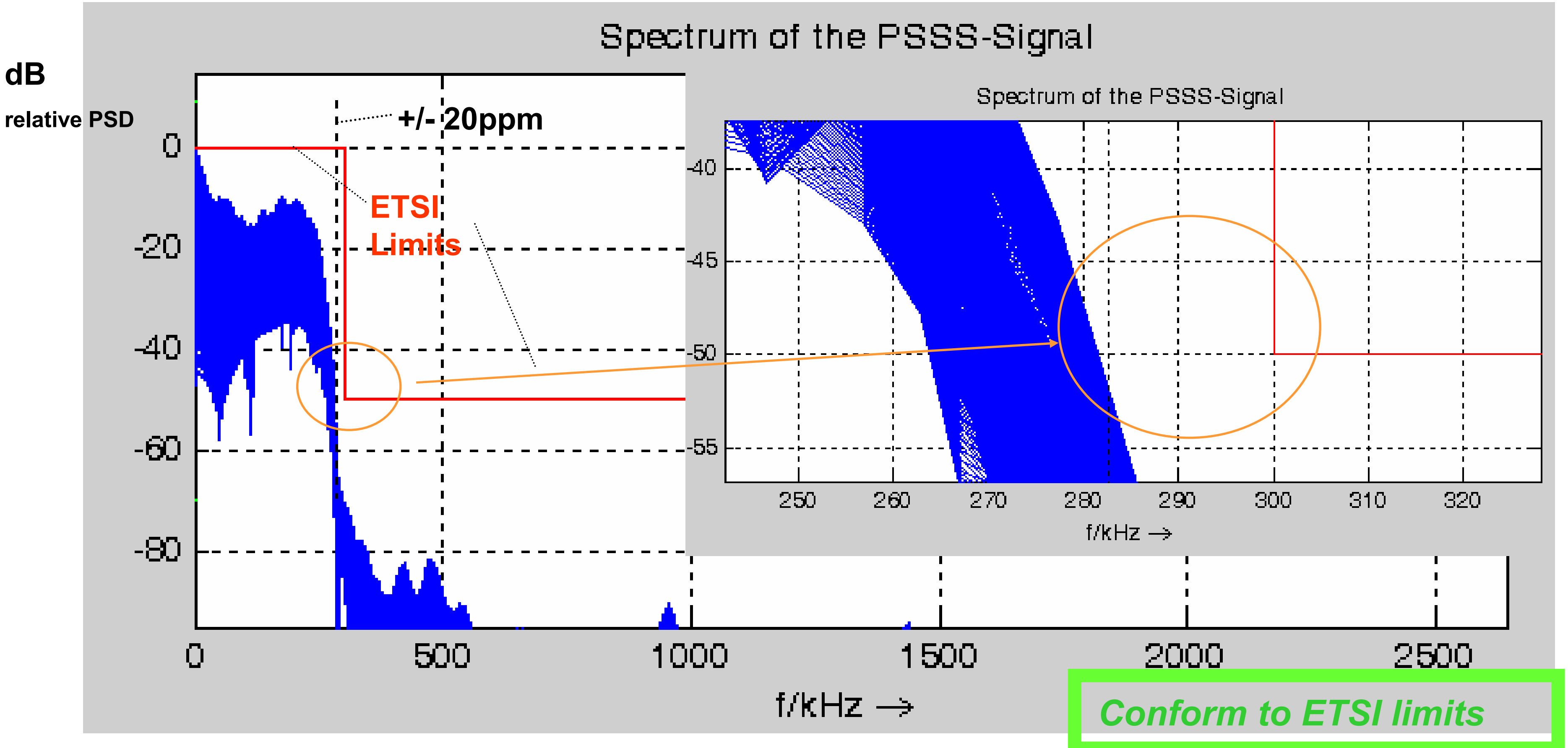
Simulations of the relative PSD in dB for the PSSS signal at 450 kchips/s, 210 kbit/s, +/- 40ppm.

PSD PSSS Signal – Passband pulse shaping, non linear, precoding



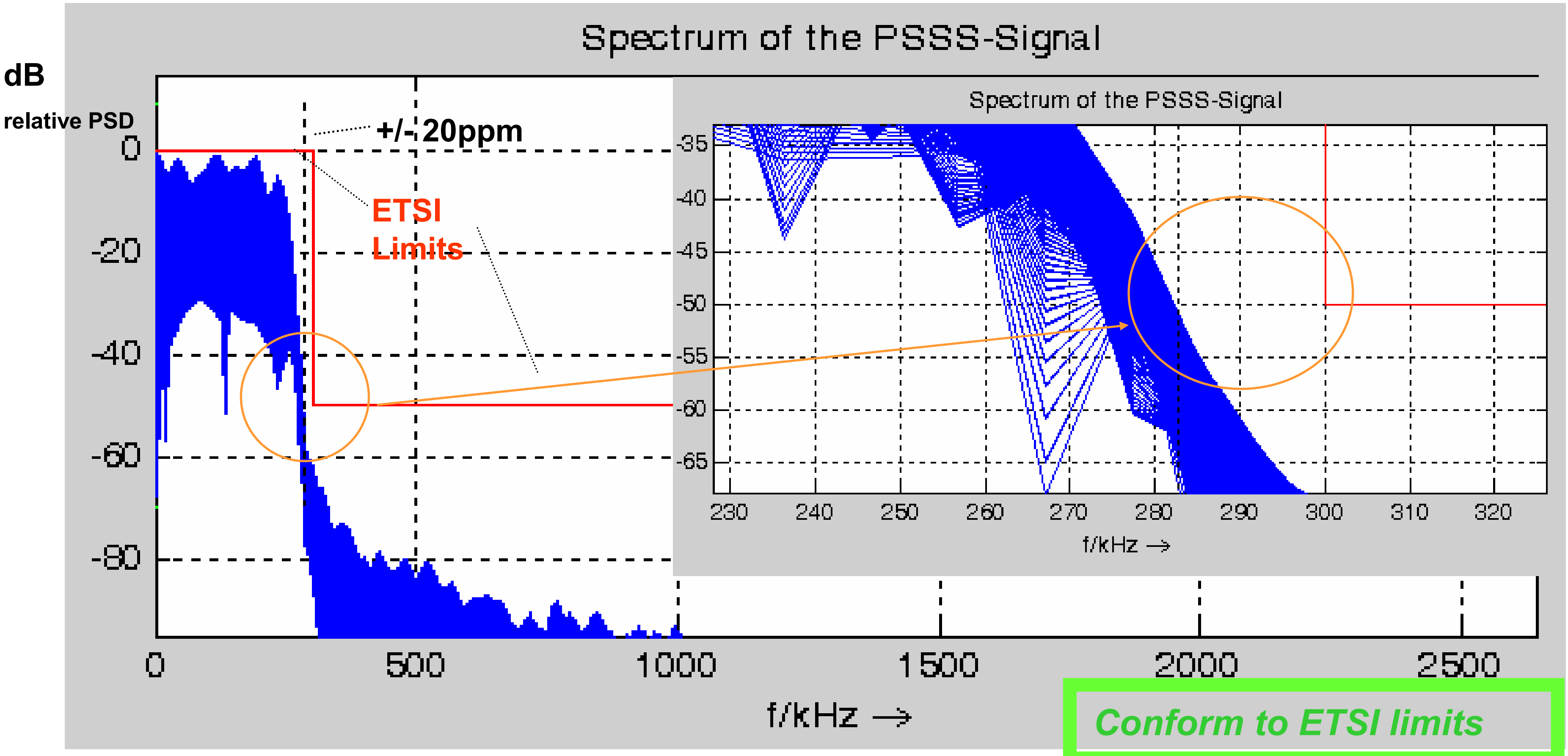
Simulations of the relative PSD in dB for the PSSS signal at 450 kchips/s, 210 kbit/s, +/- 40ppm.

PSD PSSS Signal – Passband pulse shaping, linear, no precoding



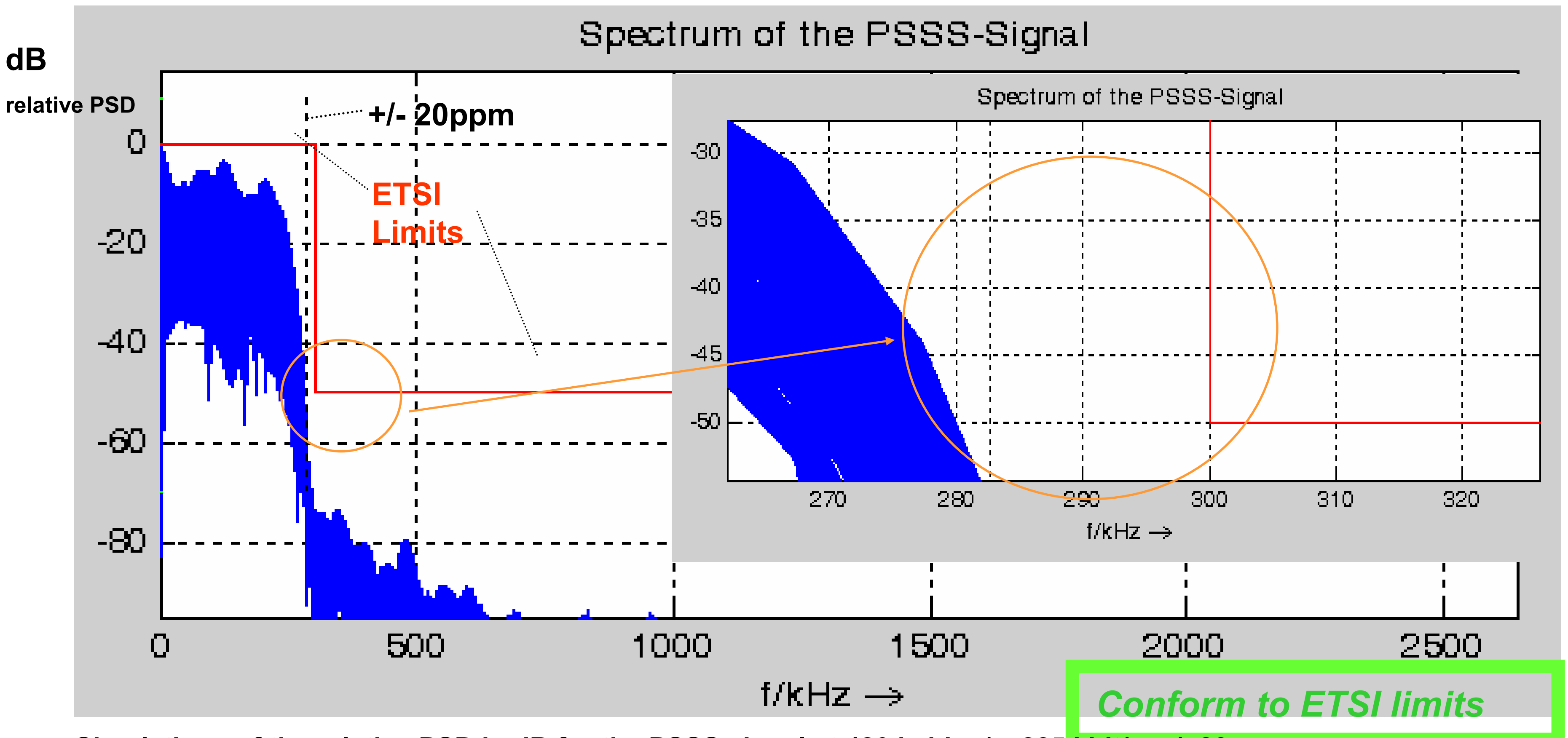
Simulations of the relative PSD in dB for the PSSS signal at 480 kchips/s, 225 kbit/s, +/- 20ppm. Conditions: linear, no precoding

PSD PSSS Signal – Passband pulse shaping, linear, precoding



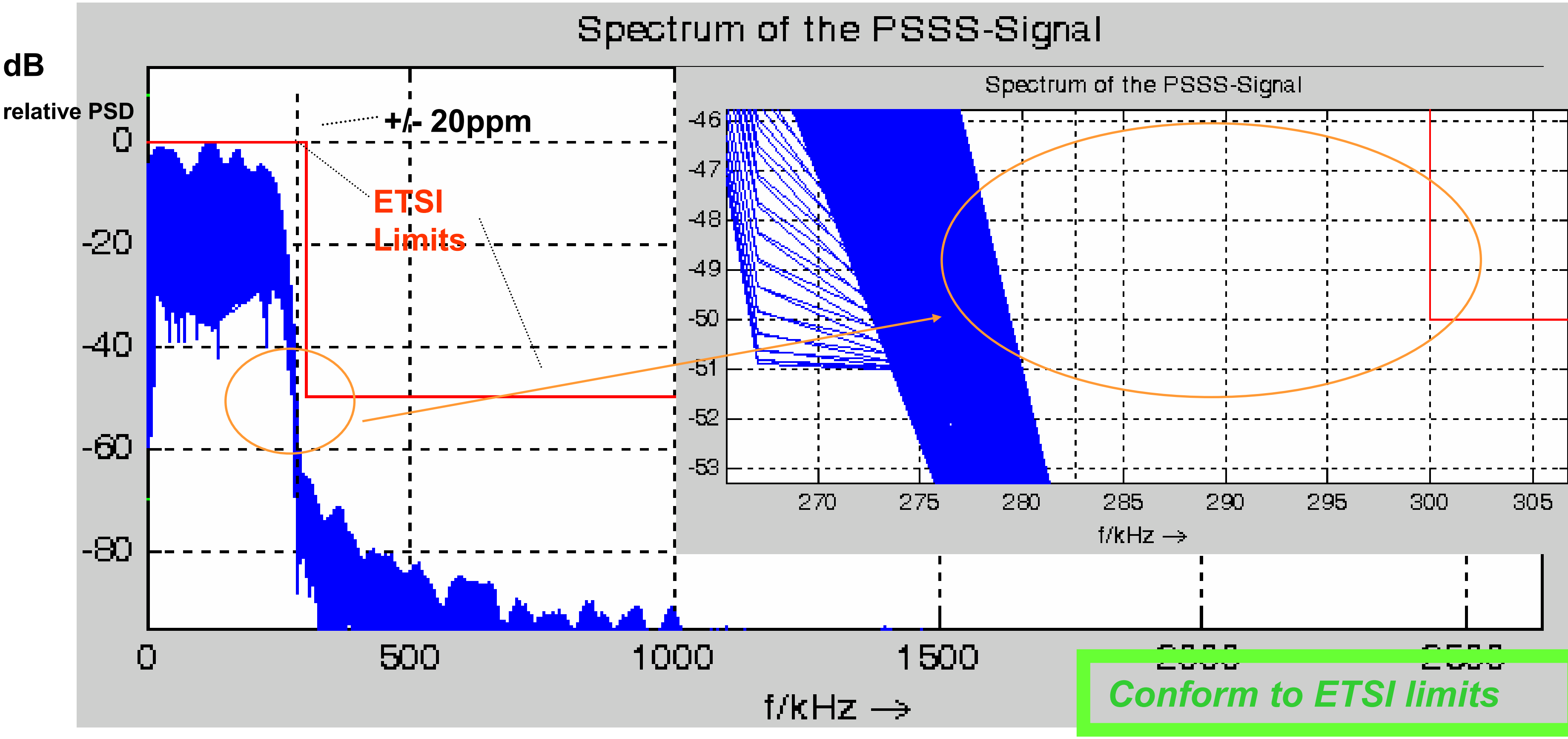
Simulations of the relative PSD in dB for the PSSS signal at 480 kchips/s, 225 kbit/s, +/- 20ppm.
Conditions: linear, precoding

PSD PSSS Signal – Passband pulse shaping, non linear, no precoding



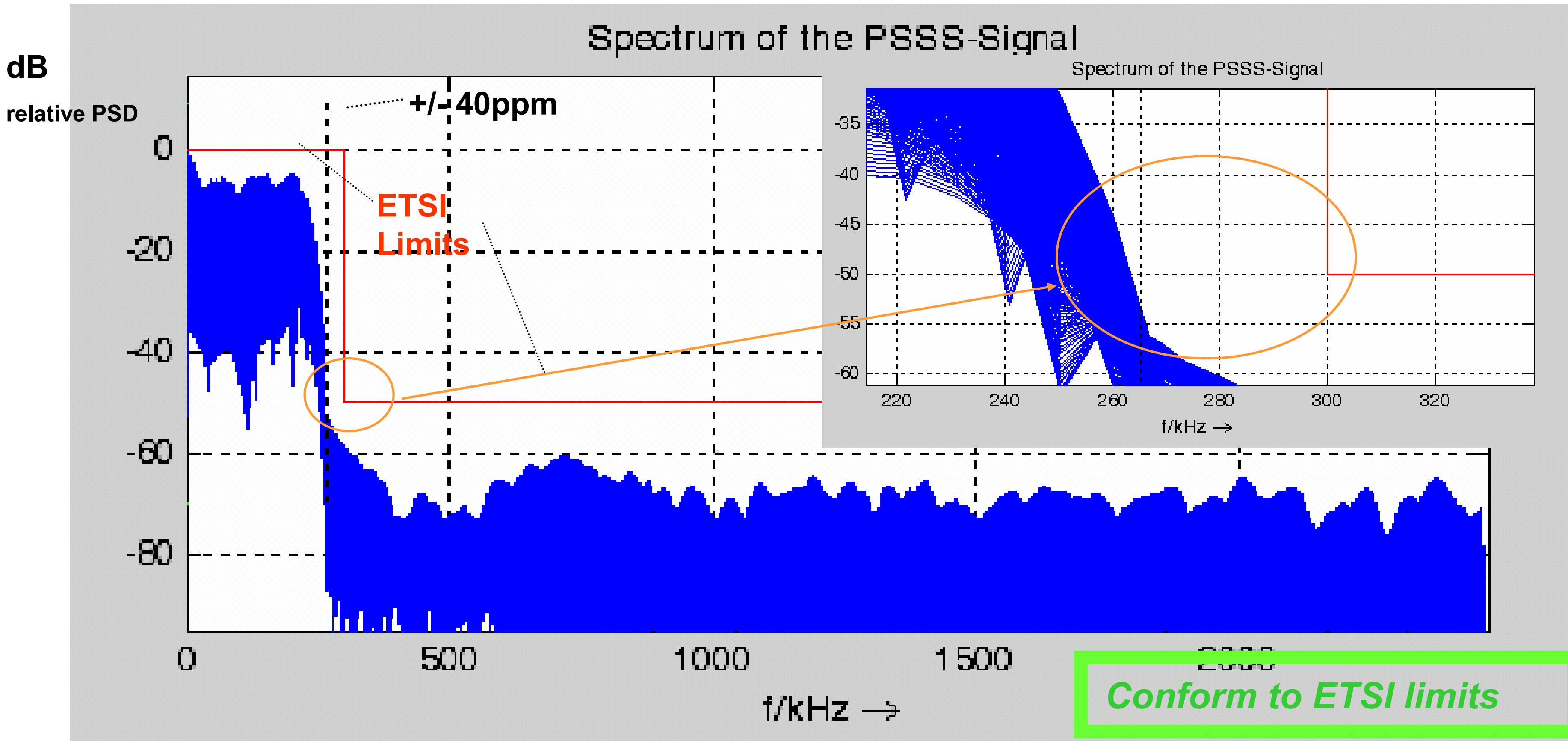
Simulations of the relative PSD in dB for the PSSS signal at 480 kchips/s, 225 kbit/s, +/- 20ppm.
Conditions: non linear, no precoding

PSD PSSS Signal – Passband pulse shaping, non linear, precoding



Simulations of the relative PSD in dB for the PSSS signal at 480 kchips/s, 225 kbit/s, +/- 20ppm.
Conditions: non linear, precoding

PSD PSSS Signal –
Baseband pulse shaping, non linear, precoding

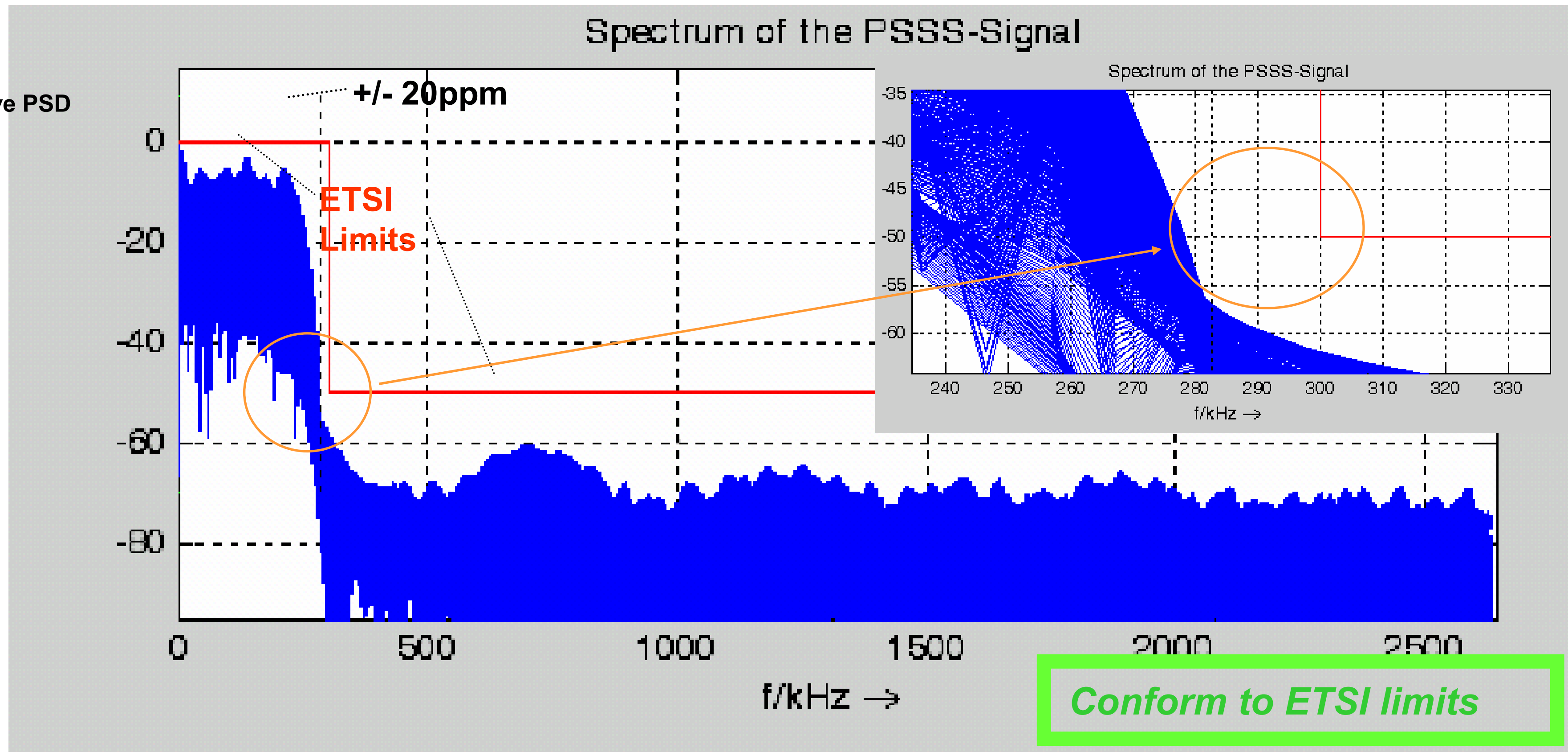


Simulations of the relative PSD in dB for the PSSS signal at 450 kchip/s 210 kbit/s, +/- 40ppm

PSD PSSS Signal – Baseband pulse shaping, non linear, precoding

dB

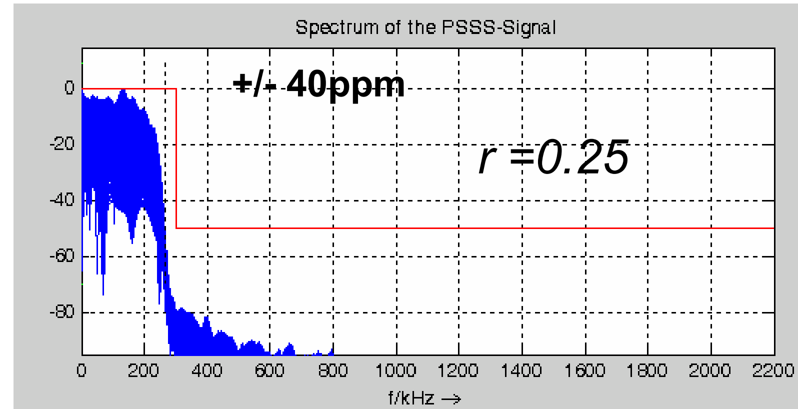
relative PSD

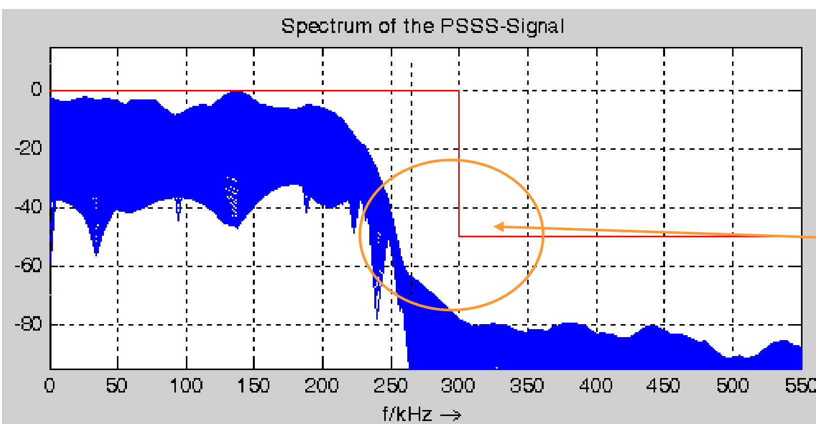


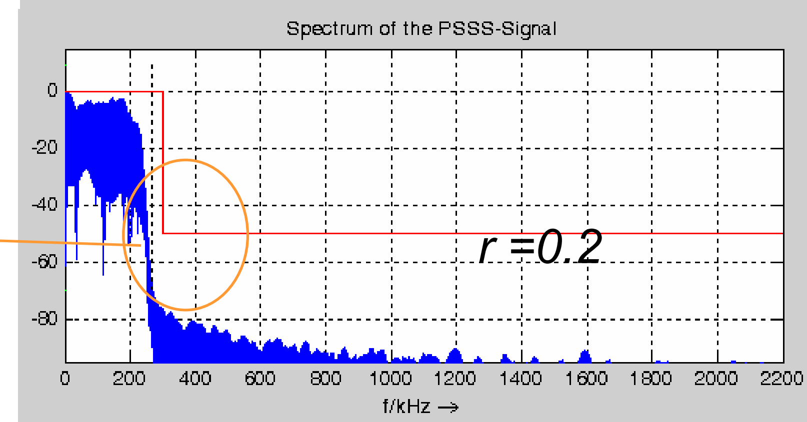
Simulations of the relative PSD in dB for the PSSS signal at 480 kchip/s, 225 kbit/s, +/-20 ppm

PSSS IQ1 Mode

#	Code #	Spectral Efficiency	Data Rate kbs	Chiprate
1	1	0,625	250	400
2	4			
3	7			
4	10			
5	13			
6	16			
7	19			
8	22			
9	25			
10	28			





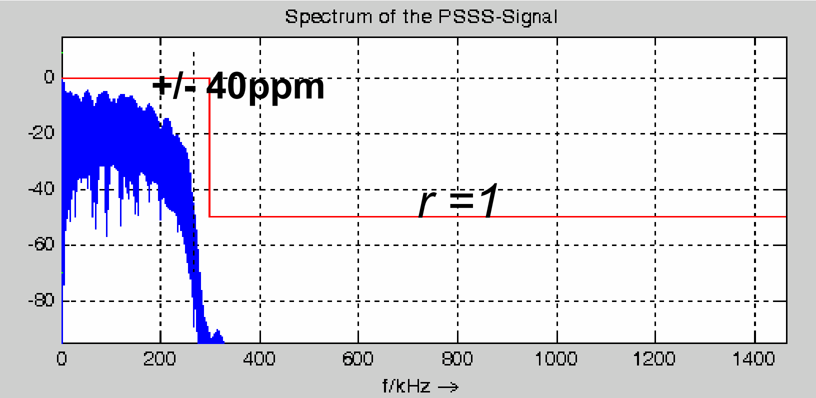


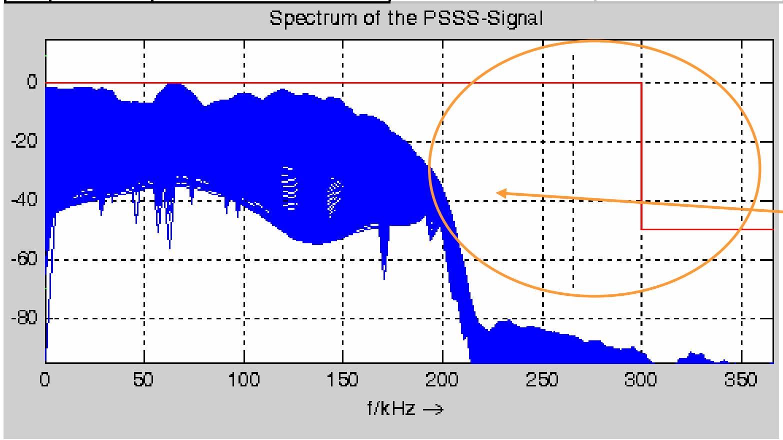
Simulations of the relative PSD in dB for the PSSS signal at 400 kchip/s 250 kbit/s.
Conditions: linear, precoding, +/-40 ppm, r = 0.25 roll on off

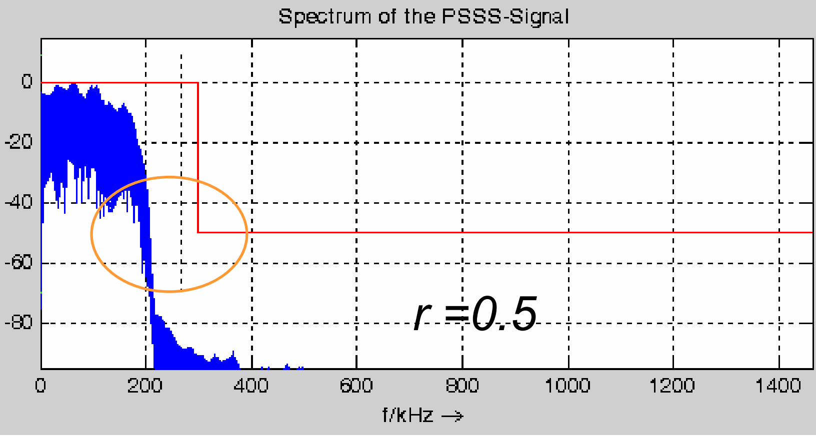
Conform to ETSI limits

PSSS IQ 2 Mode

#	Code #	Spectral Efficiency	Data Rate kbs	Chiprate
1	1	0,9375	250	266,6666667
2	3			
3	5			
4	7			
5	9			
6	11			
7	13			
8	15			
9	17			
10	19			
11	21			
12	23			
13	25			
14	27			
15	29			



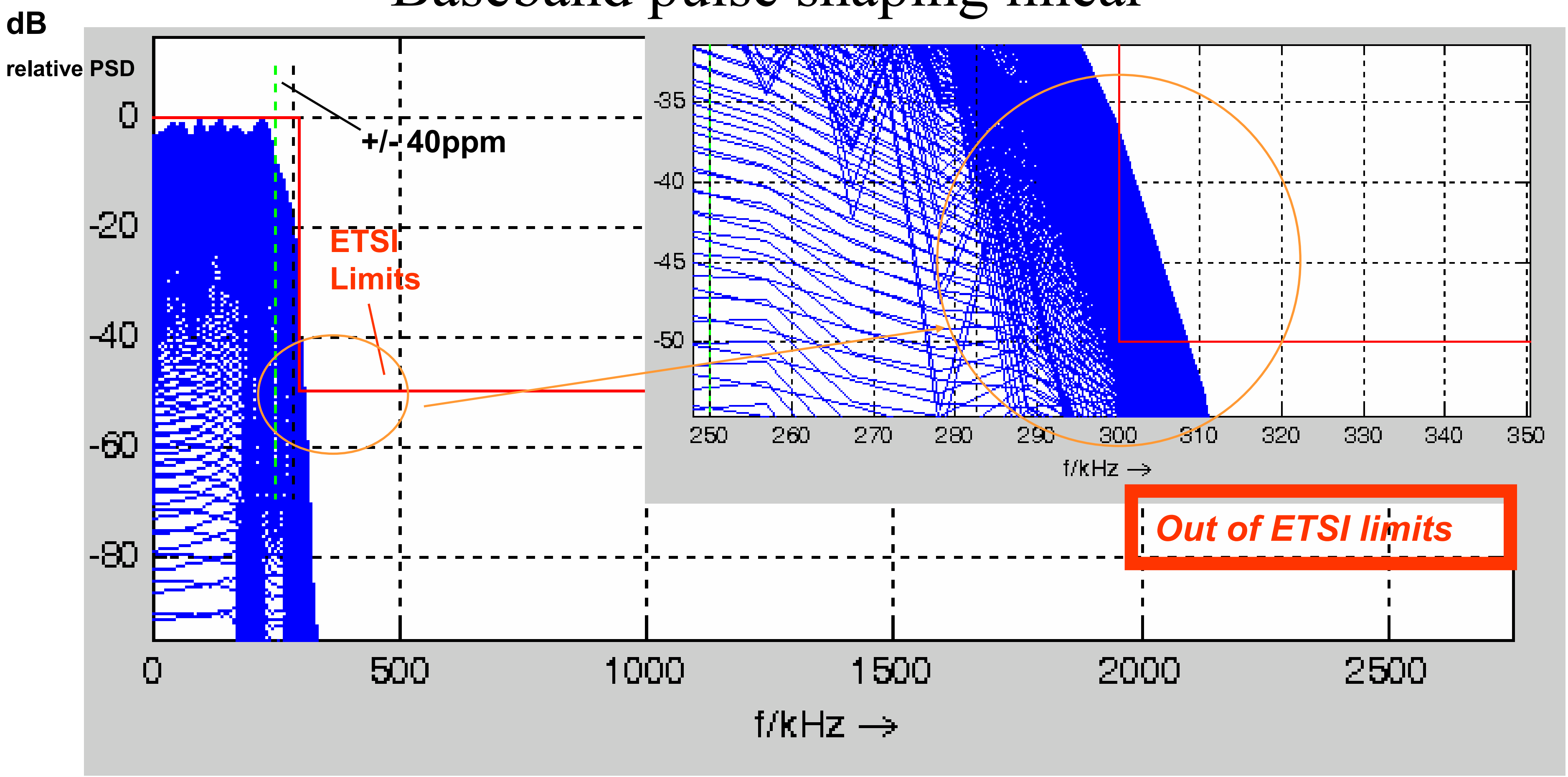




Simulations of the relative PSD in dB for the PSSS signal at 266 kchip/s 250 kbit/s.
Conditions: linear, precoding, +/-40 ppm, r = 1 roll on off

Conform to ETSI limits

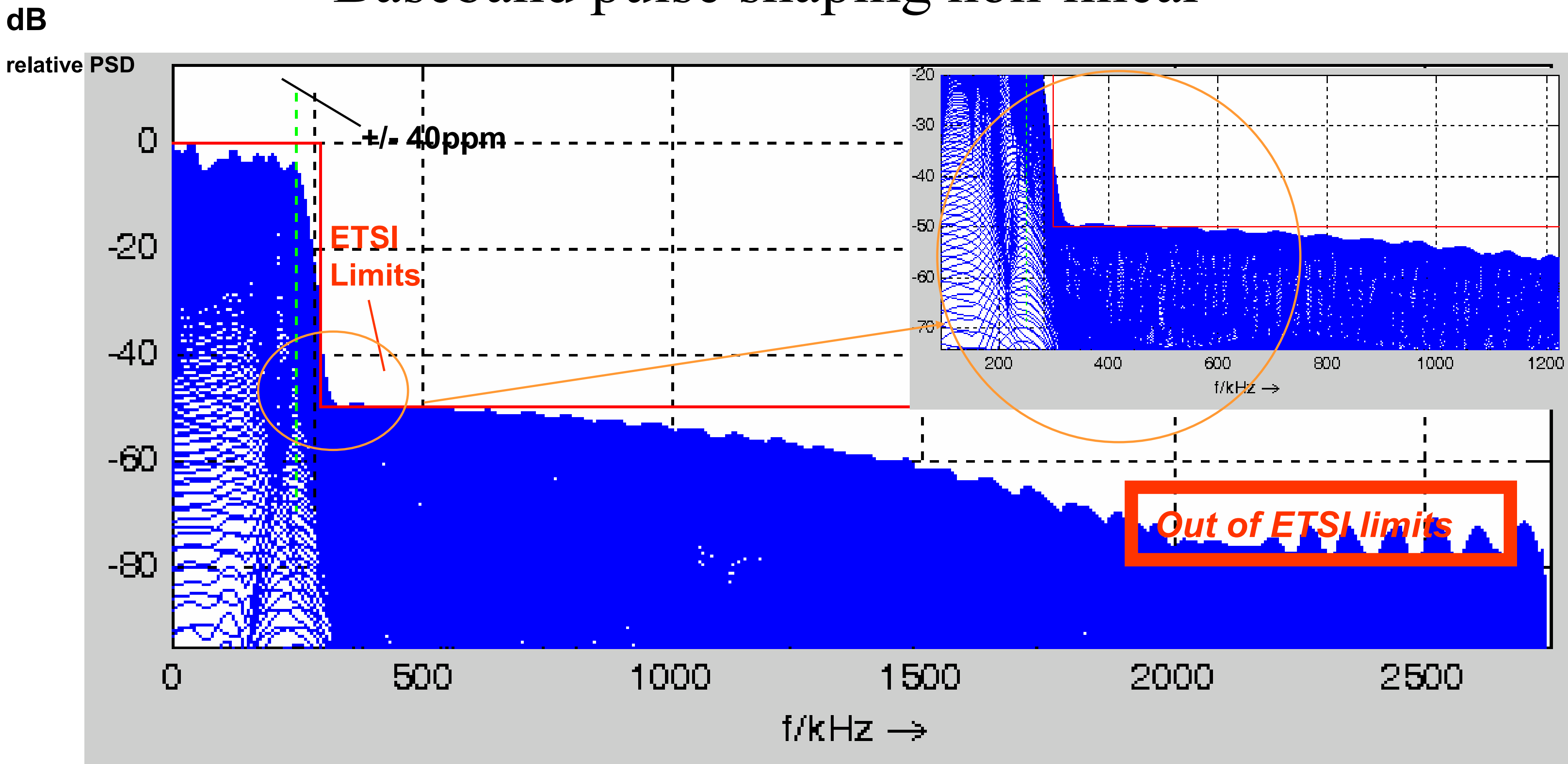
PSD for COBI8 in 600 KHz channel
Baseband pulse shaping linear



Simulations of the relative PSD in dB for the Cobi at 500 kchip/s, 250 kbit/s, r = 0.2, +/-40 ppm.

Reference for COBI 8: IEEE 802.15-04-0586-05-004b , slide 5

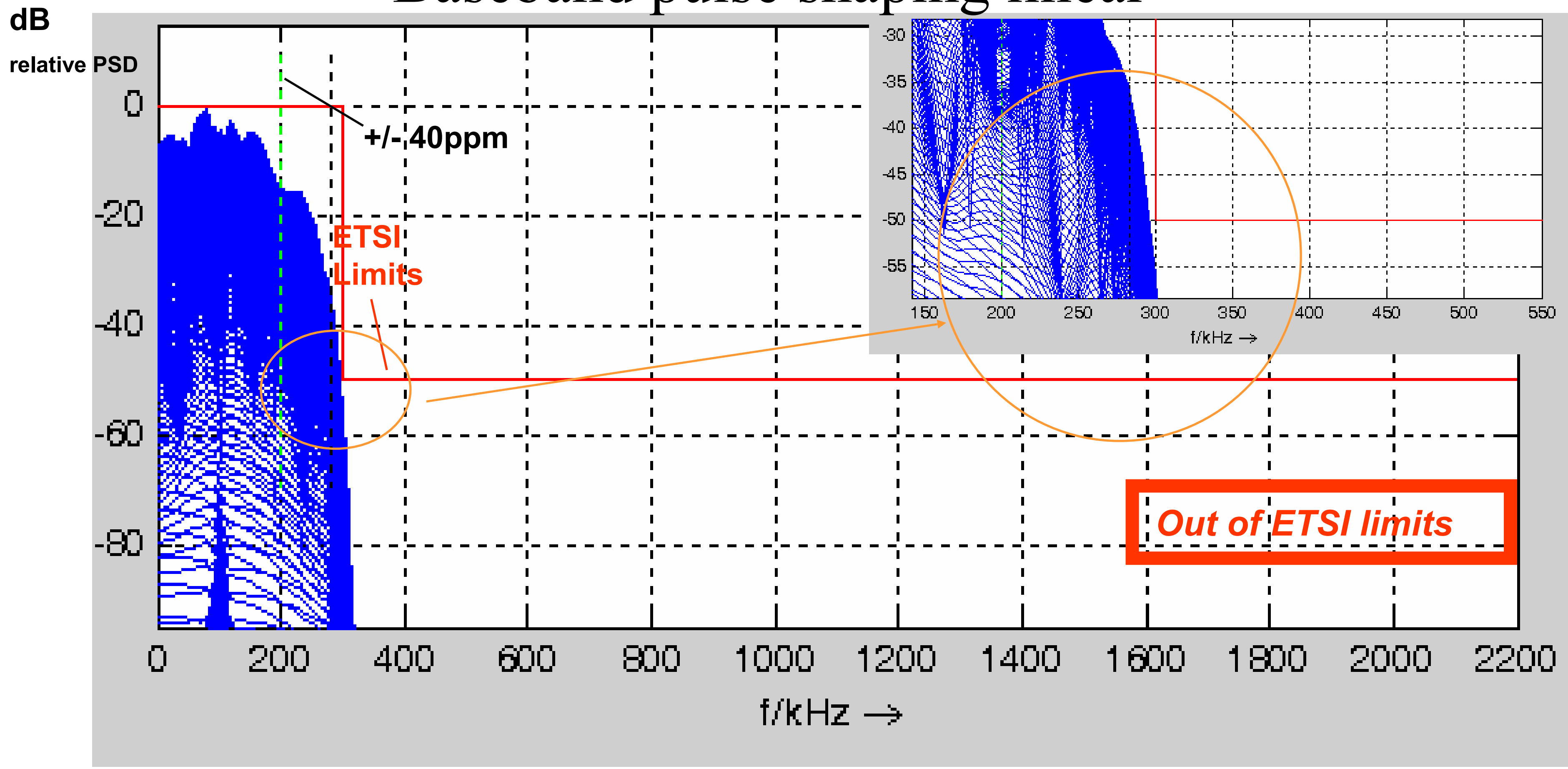
PSD for COBI8 in 600 KHz channel
Baseband pulse shaping non-linear



Simulations of the relative PSD in dB for the Cobi at 500 kchip/s, 250 kbit/s, r = 0.2, +/-40 ppm.

Reference for COBI 8: IEEE 802.15-04-0586-05-004b , slide 5

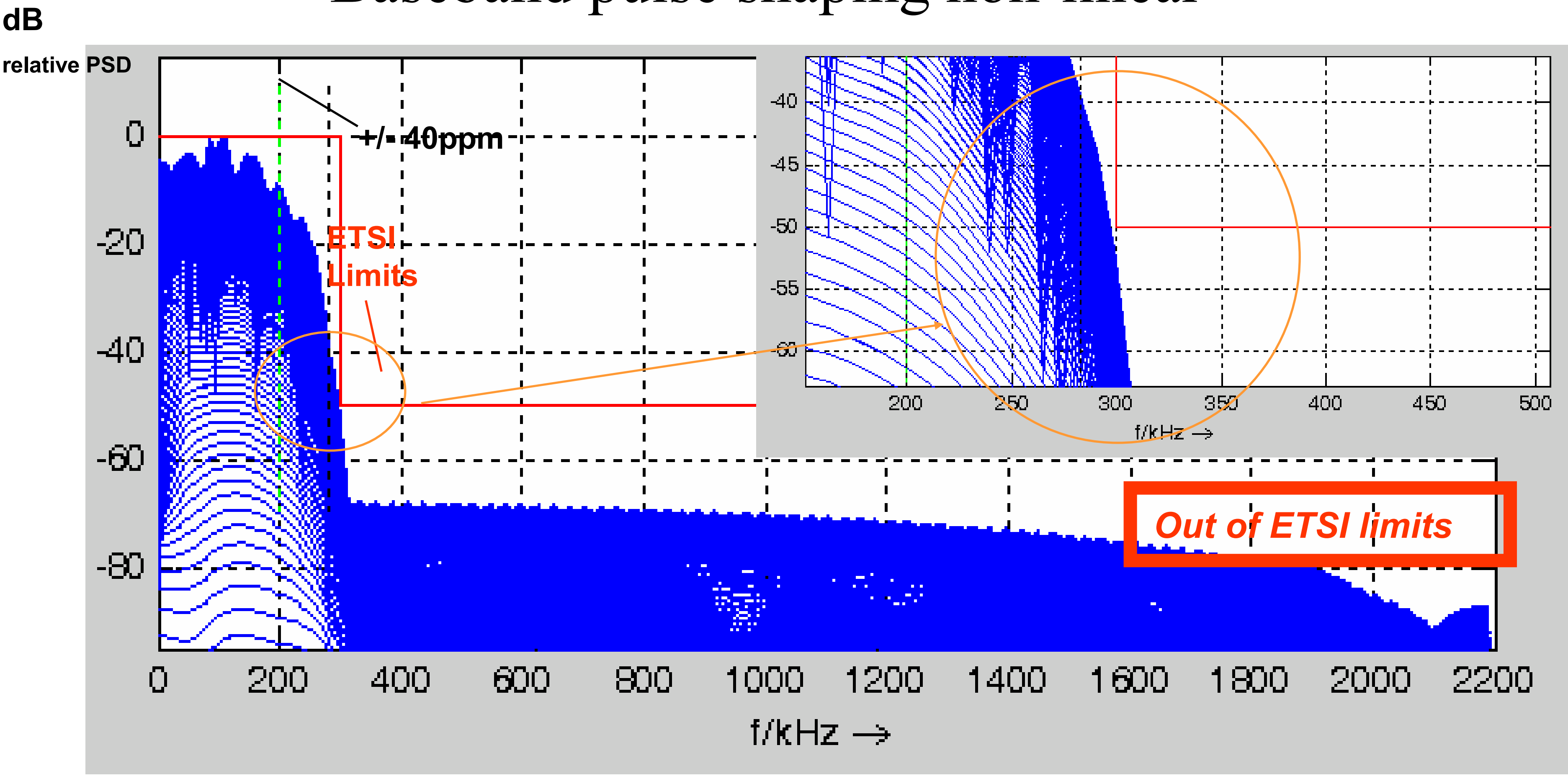
PSD for COBI8 in 600 KHz channel
Baseband pulse shaping linear



Simulations of the relative PSD in dB for the Cobi at 400 kchip/s, 200 kbit/s, r = 0.5, +/-40 ppm.

Reference for COBI 8: IEEE 802.15-04-0586-05-004b , slide 5

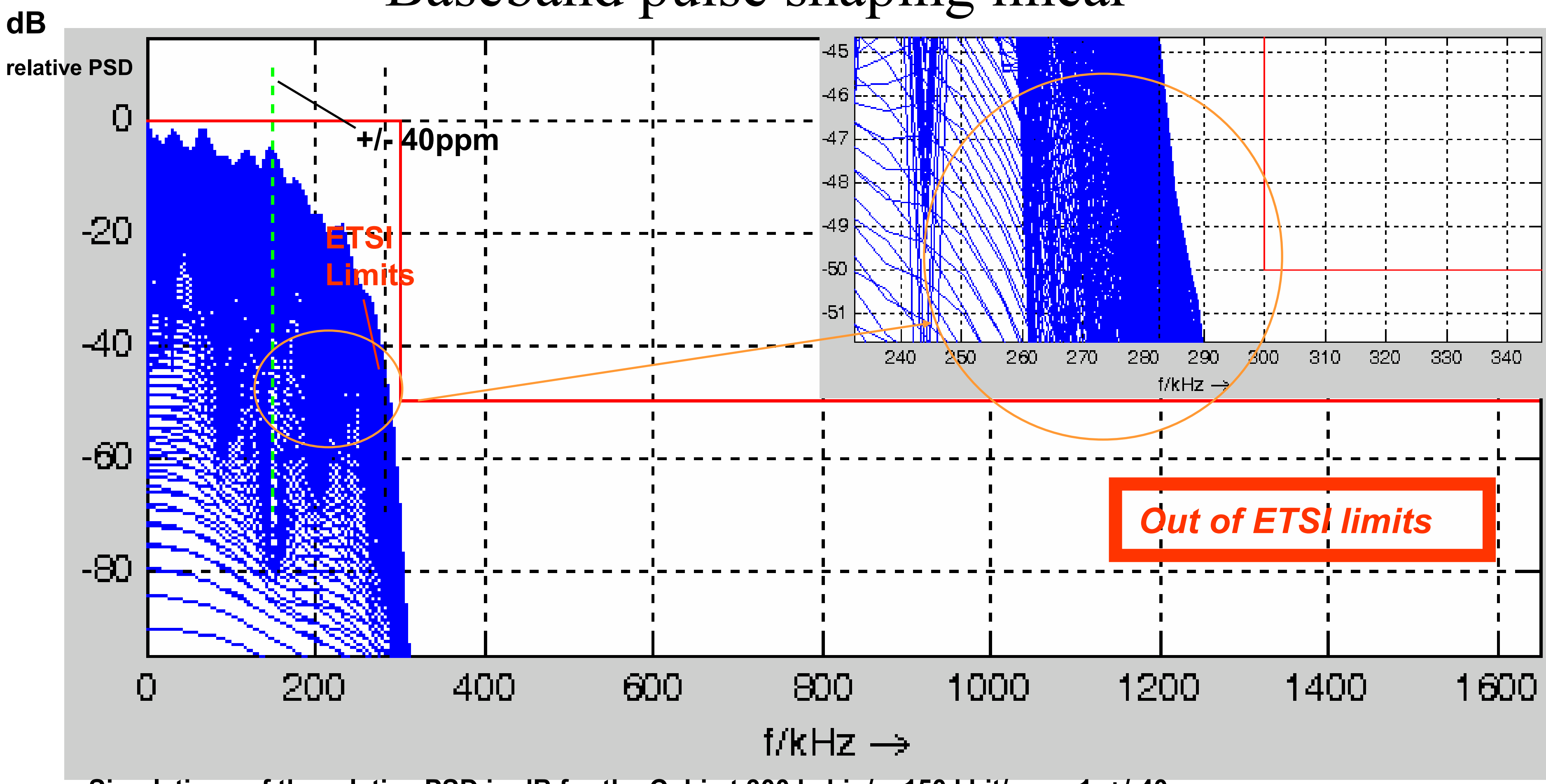
PSD for COBI8 in 600 KHz channel
Baseband pulse shaping non-linear



Simulations of the relative PSD in dB for the Cobi at 400 kchip/s, 200 kbit/s, r = 0.5, +/-40 ppm.

Reference for COBI 8: IEEE 802.15-04-0586-05-004b , slide 5

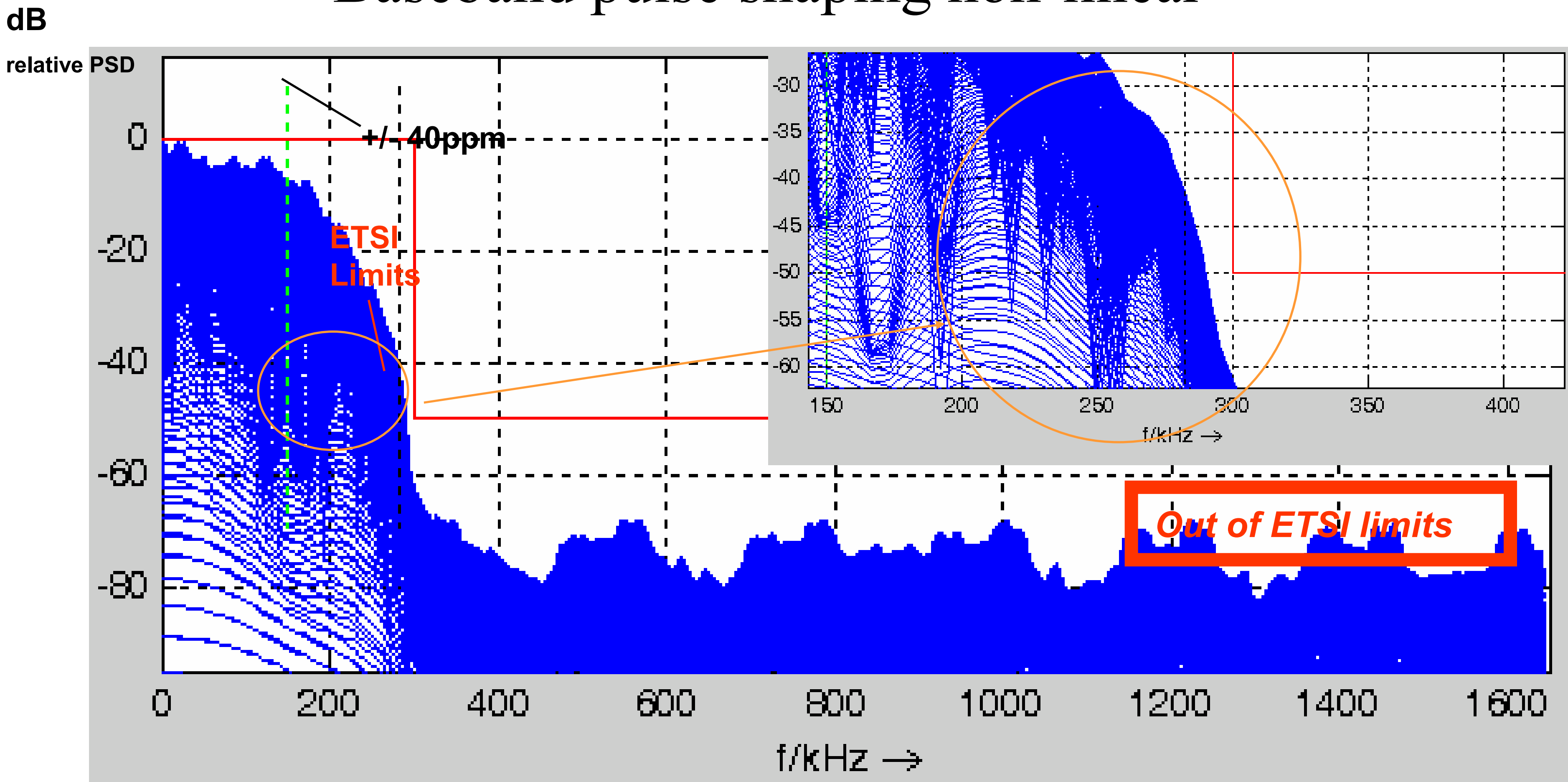
PSD for COBI8 in 600 KHz channel Baseband pulse shaping linear



Simulations of the relative PSD in dB for the Cobi at 300 kchip/s, 150 kbit/s, r = 1, +/-40 ppm.

Reference for COBI 8: IEEE 802.15-04-0586-05-004b , slide 5

PSD for COBI8 in 600 KHz channel
Baseband pulse shaping non-linear



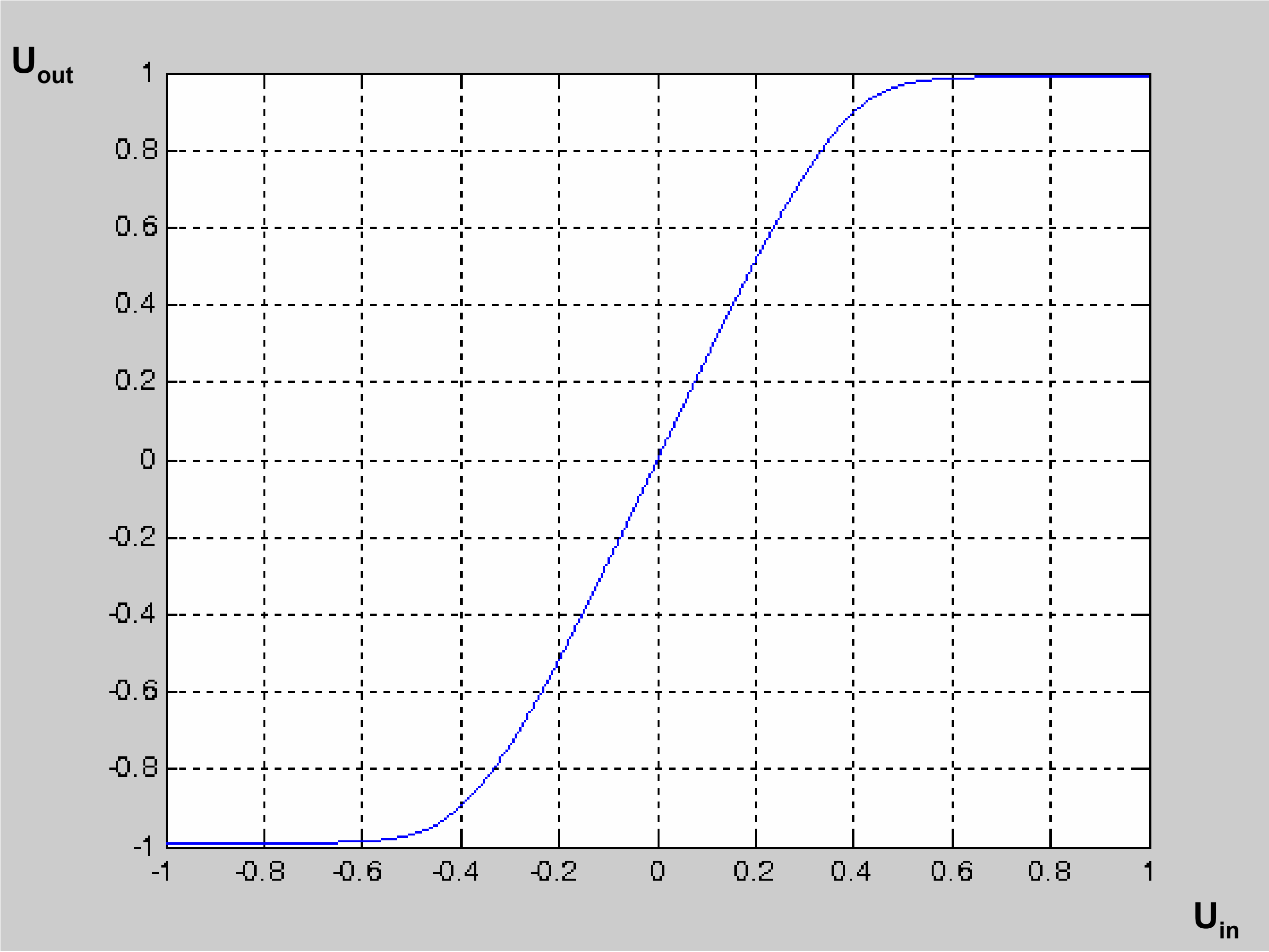
Simulations of the relative PSD in dB for the Cobi at 300 kchip/s, 150 kbit/s, r = 1 +/-40 ppm.

Reference for COBI 8: IEEE 802.15-04-0586-05-004b , slide 5

Enhanced PSD Simulations

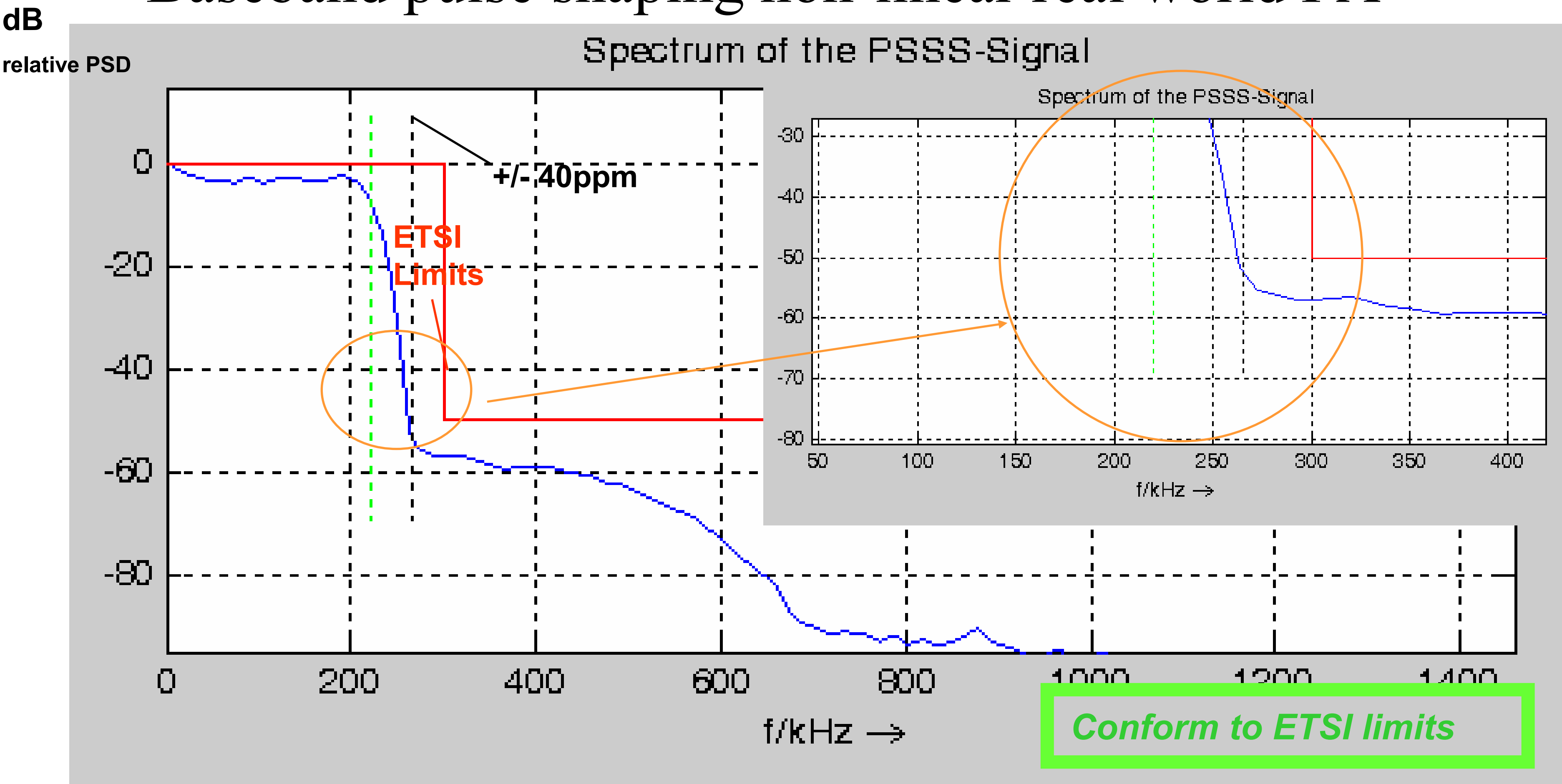
- PSD for real world non linear PA
- PSD for RAP model

Non Linear Transfer Function of a Real World PA



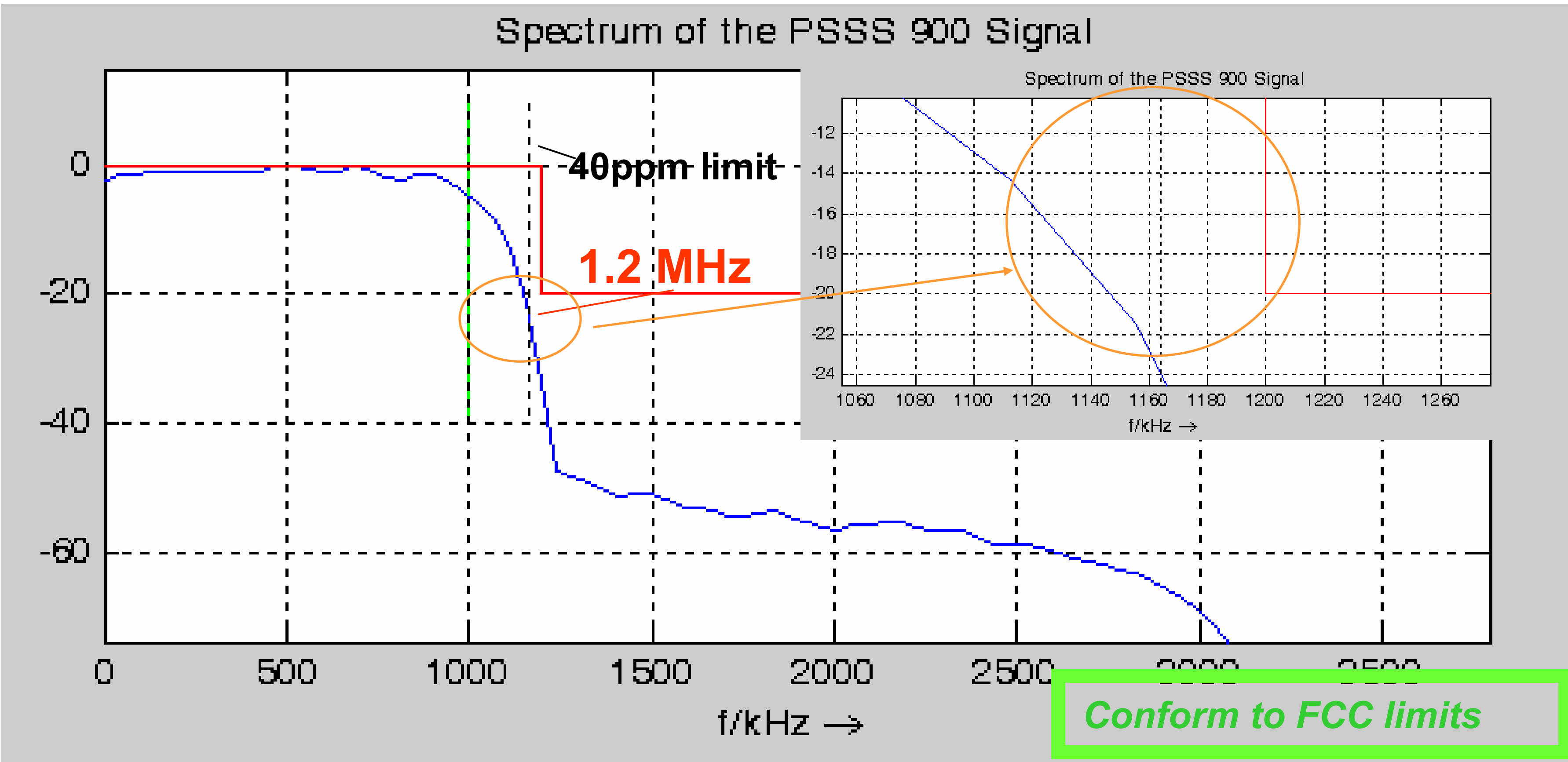
Normalized to 1

PSD for PSSS with Precoding in 600 KHz channel
Baseband pulse shaping non-linear real world PA



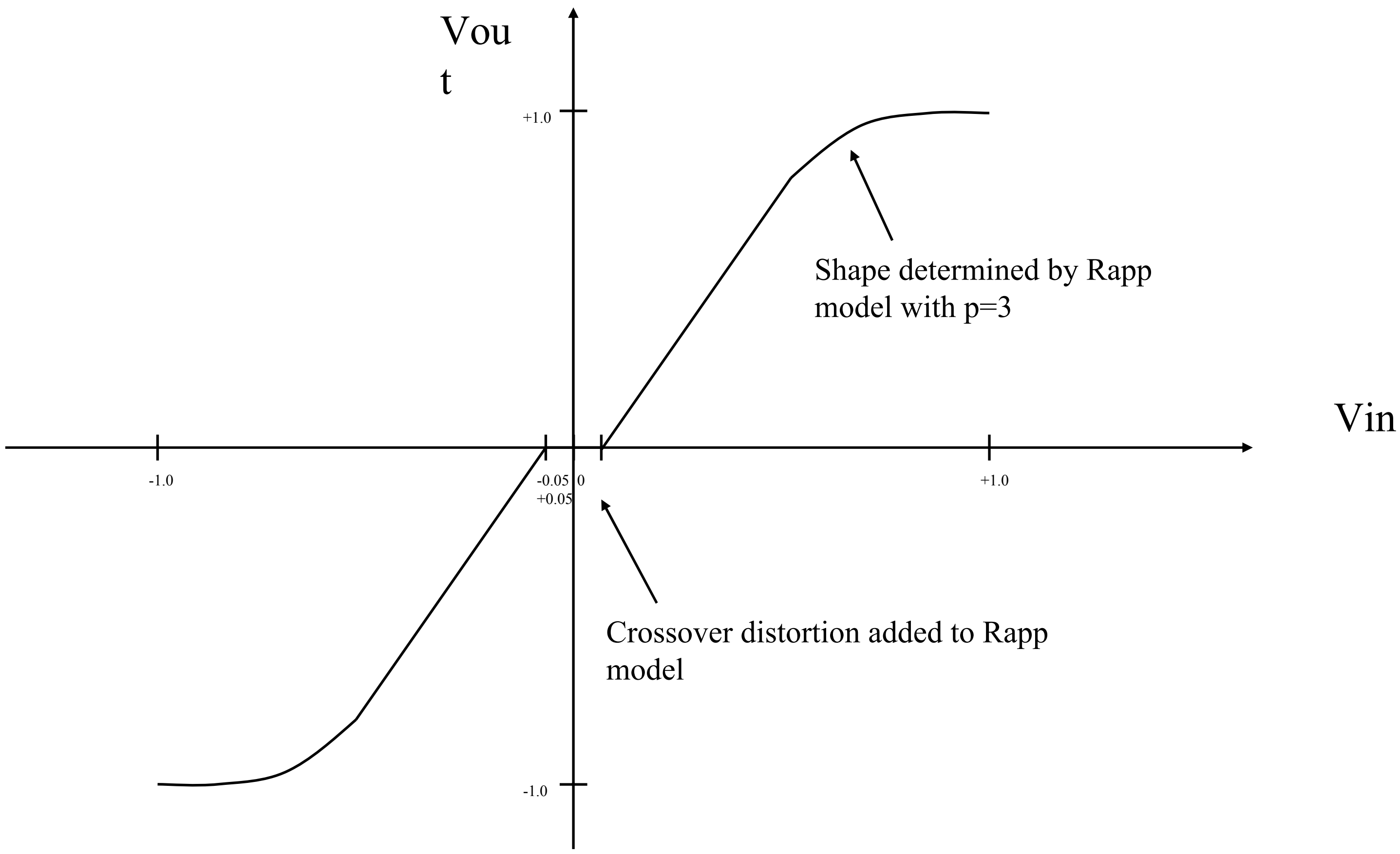
Simulations of the relative PSD in dB for the PSSS signal with precoding at 440 kchip/s 206 kbit/s, +/- 40ppm, 50% PA drive.

PSD for PSSS 900 MHz with Precoding in 2 MHz channel Baseband pulse shaping non-linear real world PA



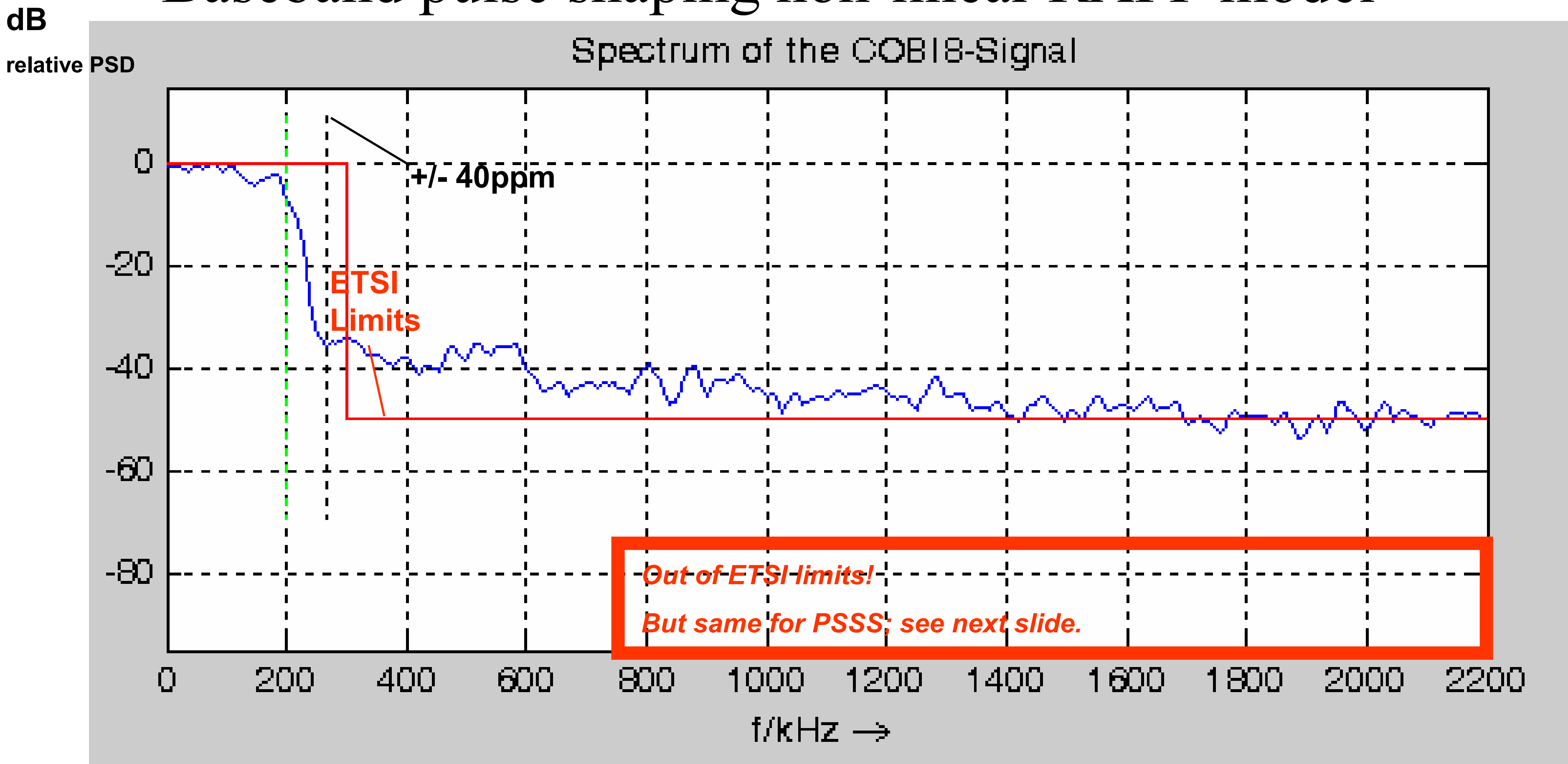
Simulations of the relative PSD in dB for the PSSS signal with precoding at 2000 kchip/s 250 kbit/s, +/- 40ppm, 50% PA drive, square root raised cosine r = 0.15.

Non Linear Transfer Function of a Rapp Model



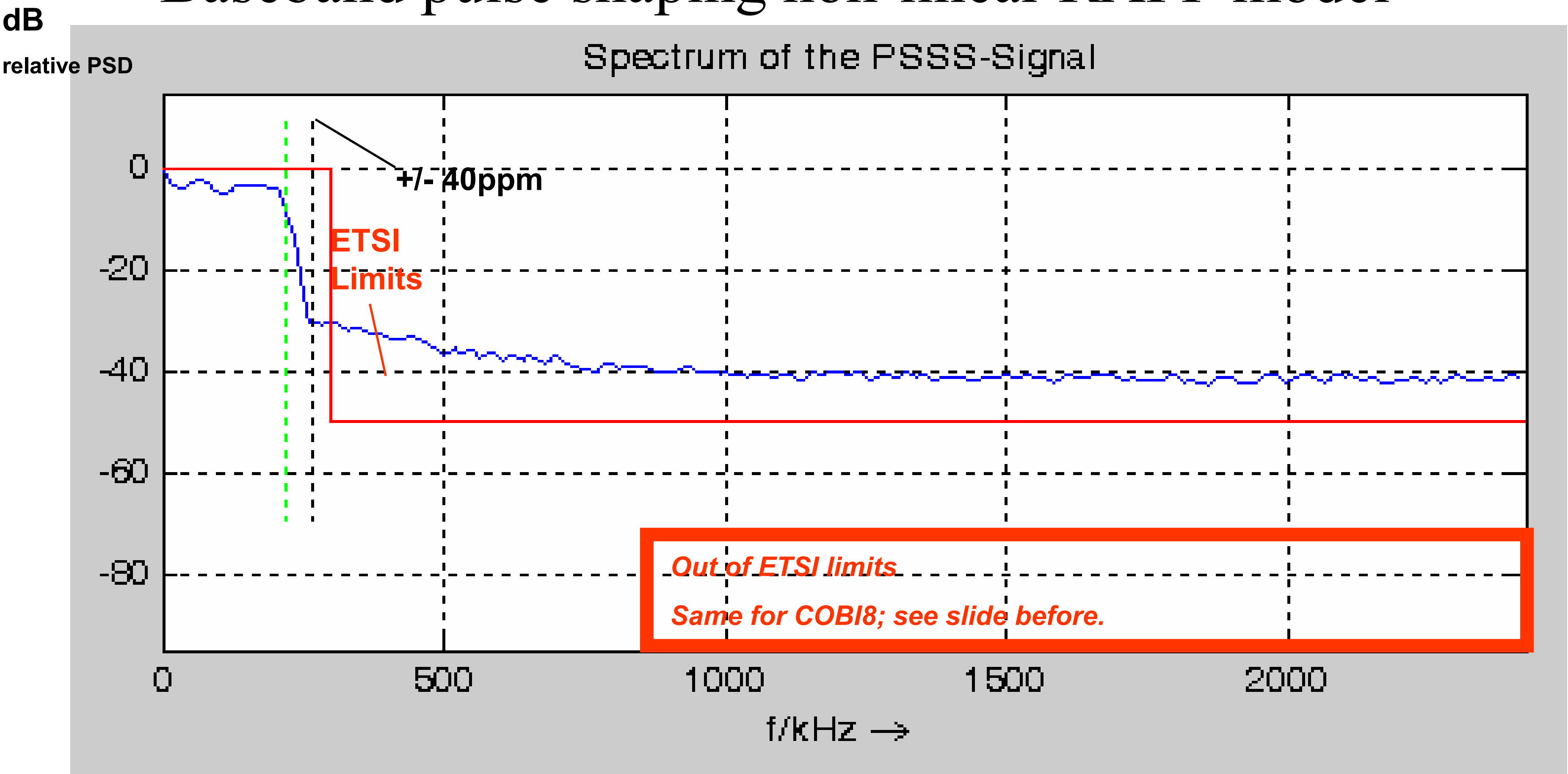
Source IEEE 802.15-04/663r0, Colin Lanzl, Ember, slide 3

PSD for COBI8 in 600 KHz channel Baseband pulse shaping non-linear RAPP model



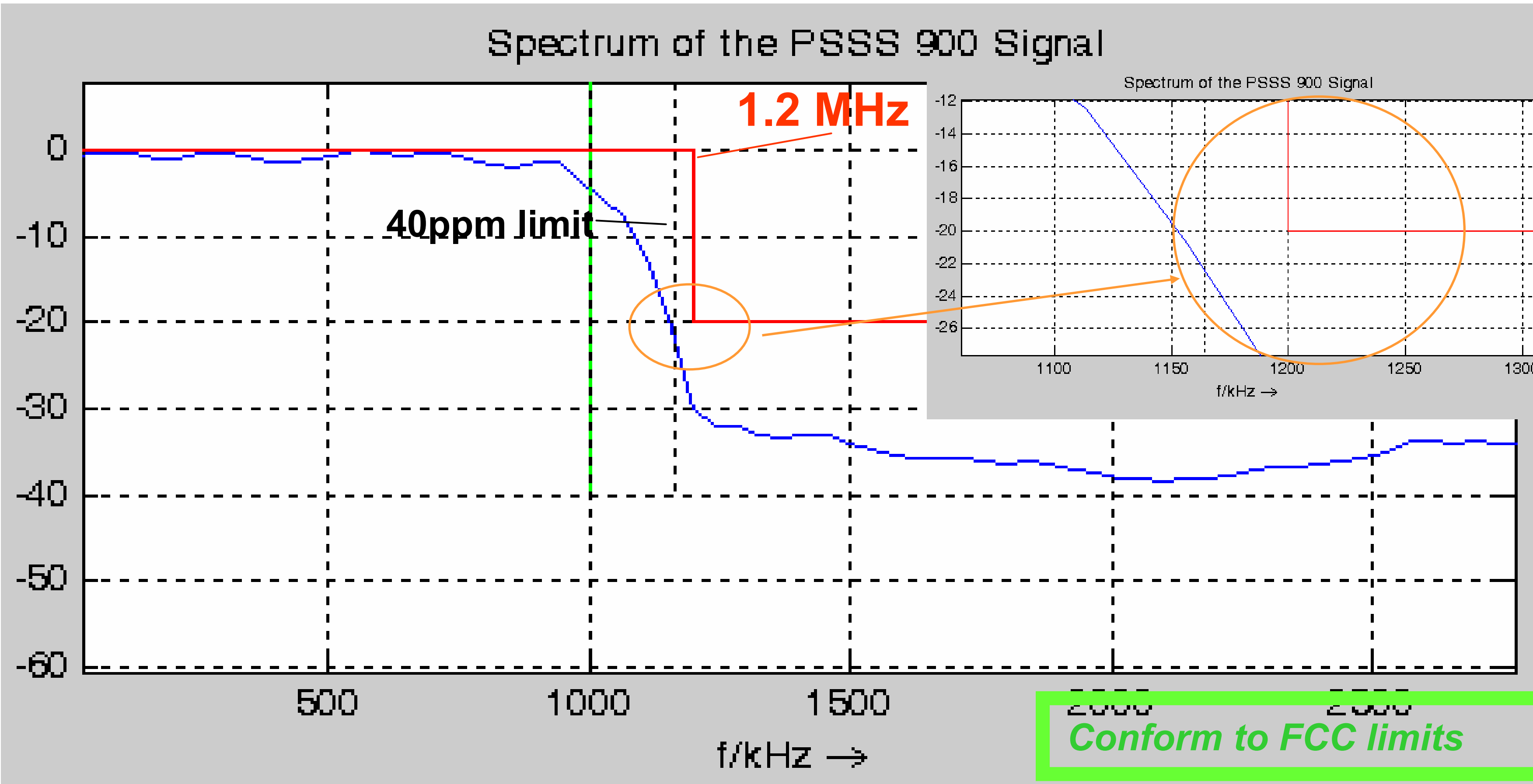
Simulations of the relative PSD in dB for the Cobi8 at 400 kchip/s, 200 kbit/s, r = 0.25 +/-40 ppm, 100% PA drive. Reference for COBI 8: IEEE 802.15-04-0586-08-004b , slide 5

PSD for PSSS with Precoding in 600 KHz channel Baseband pulse shaping non-linear RAPP model



Simulations of the relative PSD in dB for the PSSS signal with precoding at 440 kchip/s 206 kbit/s, +/- 40ppm, 100% PA drive.

PSD for PSSS 900 MHz with Precoding in 2 MHz channel Baseband pulse shaping non-linear RAPP model



Simulations of the relative PSD in dB for the PSSS signal with precoding at 2000 kchip/s 250 kbit/s, +/- 40ppm, 100% PA drive, square root raised cosine r = 0.15.

Question

- Shall we use also use the real world PA transfer function?
- We could offer it as matlab workspace file.

Crystal quality, Linearity, PSD – Conclusions

- **Crystal Quality conclusions**
  - PSSS could work in ETSI mask with +/-40ppm tolerance up to 250 kbit/s, depending of used coding
- **PSD Conclusions**
  - PSSS matches with with up to 450/480 kchip/s (40/20 ppm) the ETSI recommendations
  - Depending on pulse shaping passband / baseband Non-Linearity 20% / 1% has nearly no effect to PSD
  - PSD for COBI8[1] at 250 kbit/s violates ETSI recommendations
  - Non linearity increases also outband PSD for COBI
- **General Linearity Conclusions**
  - PSSS works even with 20% non linear PA and LNA
  - PA designs are available off-the-shelf with
    - No increase in chip cost even for linearity of 2%
    - No additional power consumption compared to C class PA used in IEEE802.15.4-2003 today
  - No impact of linearity requirements on power consumption
    - Reviewed and confirmed with two large semiconductor manufacturers
  - No implementation risk due to increased linearity required for PSSS !
- **Non-linearity simulations are confirmed with PSSS prototype**

1) Reference: IEEE 802.15-04-0586-05-004b , slide 5

Chip size and power consumption

Chip size

- High tolerance towards non-linearity and simplicity of PSSS minimizes increase in analog part
  - Estimate 0.25 mm^2 max.
- Digital part: No increase expected due to reduced complexity.
- **Total increase**: 7-10 % PHY max.
  4-6 % TRx die
  2-3 % SoC die
  < 2% SoC cost !
- Larger increase in size expected for COBI for Rake receiver etc.

Assumption: 0.18 μ CMOS process

Power consumption

- High tolerance against non-linearity and simplicity of PSSS minimizes increase in power consumption
  - Estimate Rx/Tx: 5-10% max.
    Sleep: <0.05 μA
- 15.4 2.4 Ghz chips today spread between 15...55 mA Rx
  - Effect of implementation + process is large vs. increase from PSSS (if any)
- **No visible change in battery lifetime**
  - Most energy for sleep+discharge
  - Longer battery life vs. current 868/915
- Visible increase expected for COBI due to Rake receiver etc.

Presentation Contents

- Introduction
  - Summary of OEM requirements for the TG4b PHY
- PSSS variants reviewed in this document
- PSSS Performance
  - BPSK / ASK modulation
  - O-QPSK / I/Q modulation
- PSSS Implementation aspects
  - Crystal quality – frequency offset tolerance
  - PSD
  - Chip size and power consumption
- Status
- Summary

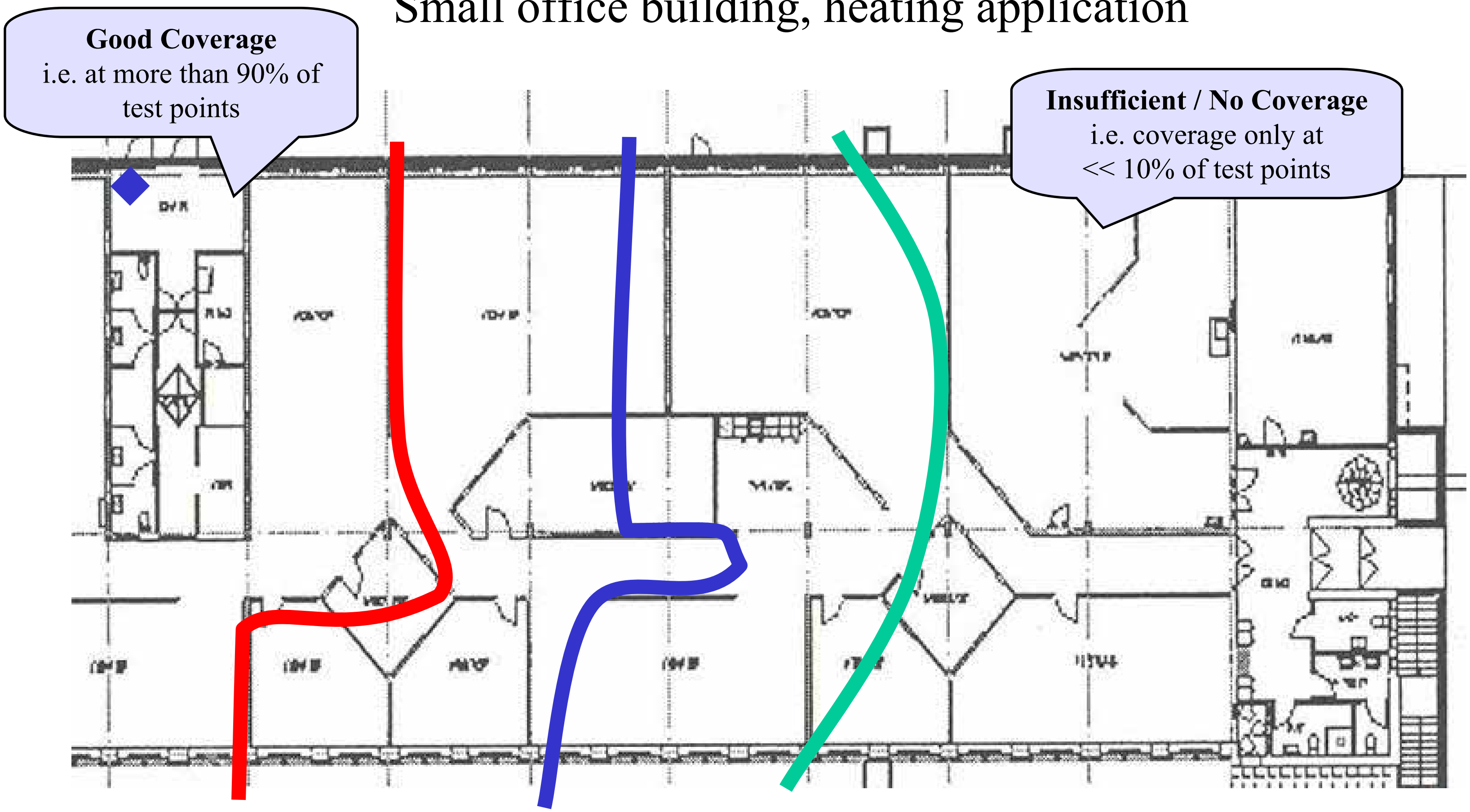
- Attachments
  - PSSS PHY Tx operation
  - Selected Rx implementation options
  - Linearity

Status

- Comprehensive research and development on PSSS has been performed based on:
  - **Full simulation**
  - **Configurable prototype for PSSS**
  - **Analytical model for PSSS**

➡ Minimal risk for implementation due to well understood technology and all building blocks being widely available

Results of first field measurements with PSSS and COBI16: Residential / light commercial environments – Small office building, heating application



- Test site: Office building (brick, sheetrock walls), rms delay spreads typ. 200 ... 400 ns
- Tested RF technology:
  - – IEEE802.15.4-2003 (2.4 GHz), 0dBm Tx
  - – PSSS 225-600, 225 kbit/s (600 kHz) in 2.4 Ghz, 0dBm Tx
  - – COBI16+1, 235 kbit/s (600 kHz) in 2.4 GHz, 0 dBm Tx

◆ Test transmitter

Comparison of PHY technologies

	PSSS 225-600	PSSS 210-600	PSSS 250-600 a/b	PSSS[1)] 250-2000	COBI16[2)]	COBI8[2)]
Bandwidth	600 kHz	600 kHz	600 kHz	2,000 kHz	2,000 kHz	600 kHz
Chiprate	480 cps	450 cps	266.6 / 400 cps	800 kcps	1 Mchip/s	500 kcps
Bitrate	225 kbit/s	210 kbit/s	250 kbit/s	250 kbit/s	250 kbit/s	250 kbit/s
Spreading	15x 32-chip seq.	15x 32-chip seq.	10/15x 32-chip seq.	5x 32 chip seq.	16x16 chip seq.	16x8 chip seq.
Pulse shape	Square root raised cosine r = 0.2	Square root raised cosine r = 0.2	Square root raised cosine r = **0.5** / 0.2	Square root raised cosine ?	Halfsine	**Raised cosine R = 0.2 Not possible[3)]**
Rake	**Not required**	**Not required**	**Not required**	**Not required**	**Required[1]**	**Required[1]**
Modulation	**BPSK + ASK**	**BPSK + ASK**	**BPSK + I/Q**	**BPSK + ASK**	OQPSK	BPSK
Complexity	**small**	**small**	**Small to medium**	**small**	**high**	**high**
MP performance E_bN_0 @ PER=10^{-3}	**31dB**	**31dB**	**27dB/30dB**	?	**>>40dB**	**>>>40dB**
Conclusion	**Attractive**	**Highly Attractive**	**Attractive**	**Highly Attractive**	**Less Attractive**	**Not Attractive**

Joint PHY (Sept. 2004) Advantage Disadvantage **Blocking point**

1): Not yet fully simulated, may still not provide required MP performance

2) Reference: IEEE 802.15-04-0586-05-004b

3): Also other proposed COBI8 versions are not conform to ETSI rec.

Summary

- PSSS is the only proposal that fulfills all OEM requirements
  - Provides very high robustness against MP fading – up to 2 µs
    i.e. visibly stronger MP fading robustness than current 2.4 GHz PHY,
    provides required higher range in many attractive, high volume target areas
  - Data rate of > 200 kbit/s at low complexity with highly backward compatible PHY,
    250 kbit/s with even simpler pulse shaping with I/Q modulation/demodulation
  - Suitable for existing and upcoming regulatory environment in Europe (ETSI)
- Analysis in TG4b has shown that PSSS is implementable at low risk
  - High confidence in results due to very comprehensive simulation model
  - Simulation results match first measurements with lab prototype
  - Full understanding of PSD shows compliance with stringent ETSI requirements
- PSSS offers highly attractive performance and increases market opportunities
  - Performance of COBI is lower than with current 2.4 GHz PHY coding
  - PSSS is competitive with Bluetooth radios in industrial / commercial environments
- PSSS provides for Europe significantly more attractive solution than COBI
  - Lower COBI16 performance is acceptable for US
    if higher permitted Tx power is used (only if feasible with regard to PSD!)
  - Use of Rake receiver is inconsistent with IEEE802.15.4 objectives

Attachments

Changes vs. PSSS presentation at March 2004 meeting (Orlando)

- Unchanged basic proposal for parallel reuse of 2.4 GHz PHY!
  - Added option of use of BPSK/ASK instead of O-QPSK
    - Based on OEM and semiconductor manufacturers requirements
    - To avoid added complexity and cost for two radio cores
    - To avoid doubling required bandwidth for O-QPSK
  - Added option to reduce 868 Mhz bandwidth to 500 Khz
    - Changed to reduce implementation complexity and cost
    - Bitrate of 234 kbit/s changed to 225 kbit/s based on input from September 2004 meeting to have "more even" bit rate
    - 210 kbit/s and 250 kbit/s variants added based on chip manufacturer's inputs in TG4b PHY subcommittee to even further reduce implementation cost
  - Details of combining provided that were not shown in March 2004
    - Coding gain through simple precoding in combiner

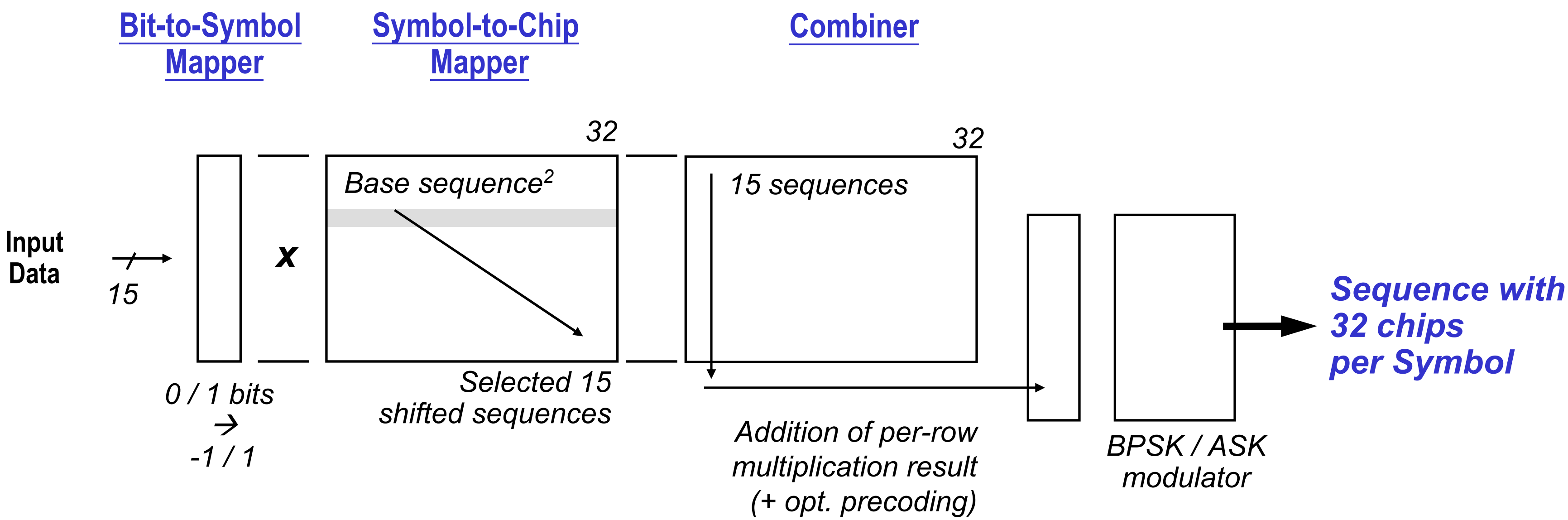
- Added new results on PSSS
  - Solution performance
  - Implementation aspects
  - Status

Used Matlab Code for Discrete Channel

```
L=2
% L=2 equal 370 ns RMS Delay Spread
profile = zeros(1,10*L+1);
profile(1:L:end) = exp(-(0:10)/2);
profile = profile/(sum(profile));
channel = sqrt(profile/2).*(randn(size(profile))+j*randn(size(profile)));
signal_out = zeros(size(signal_in));
for k = 0:10
  signal_out=signal_out+channel(k+1)*[zeros(1,k*L) signal_in(1:length(signal_in)-k*L)];
end
```

Source:
Paul Gorday Freescale IEEE 802.15-04-0585-00-004b, slide 9

PSSS – Tx – BPSK/ASK variant (15/32 bit/s/Hz)[1]



...addition of multiple parallel sequences instead of selection of single sequence

1: PSSS 225-600 + PSSS 210-600

2: Use of single base sequence simplifies implementation in Rx

PSSS –BPSK/ASK option (15/32 bit/s/Hz) – Coding table

Symbol-to-Chip Mapper

# Bit	Chip Values																															
1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1
2	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1
3	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1
4	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1
5	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1
6	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1
7	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1
8	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1
9	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1
10	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1
11	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1
12	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1
13	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1
15	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1
15	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	24	25	26	27	28	29	30	31	32

PSSS –BPSK/ASK option (15/32 bit/s/Hz) – Coding example

# Bit	Data
1	-1
2	-1
3	-1
4	-1
5	-1
6	-1
7	-1
8	1
9	-1
10	-1
11	-1
12	-1
13	-1
15	-1
15	-1

Code Words

1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	24	25	26	27	28	29	30	31	32
-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1
-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1
-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1
1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1
-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1
1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1
-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1
1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1
1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1
1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1
-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1
1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1
1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1
-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1
-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1

x =

PSSS Symbol

1	1	1	1	-1	1	1	-1	1	-1	-1	1	1	-1	-1	-1	-1	-1	1	1	1	-1	-1	1	-1	-1	-1	1	-1	1	-1	1
1	-1	1	1	1	1	-1	1	1	-1	1	-1	-1	1	1	-1	-1	-1	-1	-1	1	1	1	-1	-1	1	-1	-1	-1	1	-1	1
1	-1	1	-1	1	1	1	1	-1	1	1	-1	1	-1	-1	1	1	-1	-1	-1	-1	-1	1	1	1	-1	-1	1	-1	-1	-1	1
-1	-1	1	-1	1	-1	1	1	1	1	-1	1	1	-1	1	-1	-1	1	1	-1	-1	-1	-1	-1	1	1	1	-1	-1	1	-1	-1
1	-1	-1	-1	1	-1	1	-1	1	1	1	1	-1	1	1	-1	1	-1	-1	1	1	-1	-1	-1	-1	-1	1	1	1	-1	-1	1
-1	-1	1	-1	-1	-1	1	-1	1	-1	1	1	1	1	-1	1	1	-1	1	-1	-1	1	1	-1	-1	-1	-1	-1	1	1	1	-1
1	1	-1	-1	1	-1	-1	-1	1	-1	1	-1	1	1	1	1	-1	1	1	-1	1	-1	-1	1	1	-1	-1	-1	-1	-1	1	1
1	-1	-1	-1	1	1	-1	1	1	1	-1	1	-1	1	-1	-1	-1	-1	1	-1	-1	1	-1	1	1	-1	-1	1	1	1	1	1
-1	-1	-1	1	1	1	-1	-1	1	-1	-1	-1	1	-1	1	-1	1	1	1	1	-1	1	1	-1	1	-1	-1	1	1	-1	-1	-1
-1	-1	-1	-1	-1	1	1	1	-1	-1	1	-1	-1	-1	1	-1	1	-1	1	1	1	1	-1	1	1	-1	1	-1	-1	1	1	-1
1	1	-1	-1	-1	-1	-1	1	1	1	-1	-1	1	-1	-1	-1	1	-1	1	-1	1	1	1	1	-1	1	1	-1	1	-1	-1	1
-1	-1	1	1	-1	-1	-1	-1	-1	1	1	1	-1	-1	1	-1	-1	-1	1	-1	1	-1	1	1	1	1	-1	1	1	-1	1	-1
-1	1	-1	-1	1	1	-1	-1	-1	-1	-1	1	1	1	-1	-1	1	-1	-1	-1	1	-1	1	-1	1	1	1	1	-1	1	1	-1
1	1	-1	1	-1	-1	1	1	-1	-1	-1	-1	-1	1	1	1	-1	-1	1	-1	-1	-1	1	-1	1	-1	1	1	1	1	-1	1
1	-1	1	1	-1	1	-1	-1	1	1	-1	-1	-1	-1	-1	1	1	1	-1	-1	1	-1	-1	-1	1	-1	1	-1	1	1	1	1

\+

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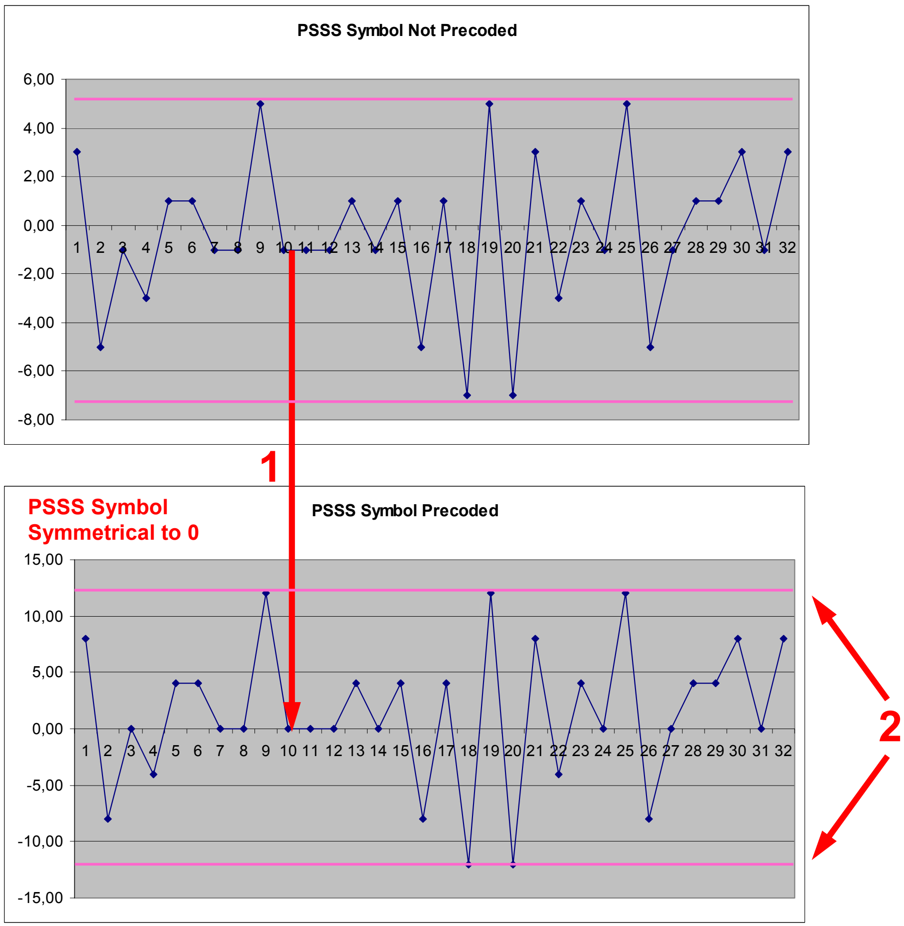
3	-5	-1	-3	1	1	-1	-1	5	-1	-1	-1	1	-1	1	-5	1	-7	5	-7	3	-3	1	-1	5	-5	-1	1	1	3	-1	3

PSSS Symbol with 32 Chips

PSSS –BPSK/ASK option (15/32 bit/s/Hz) – Precoding

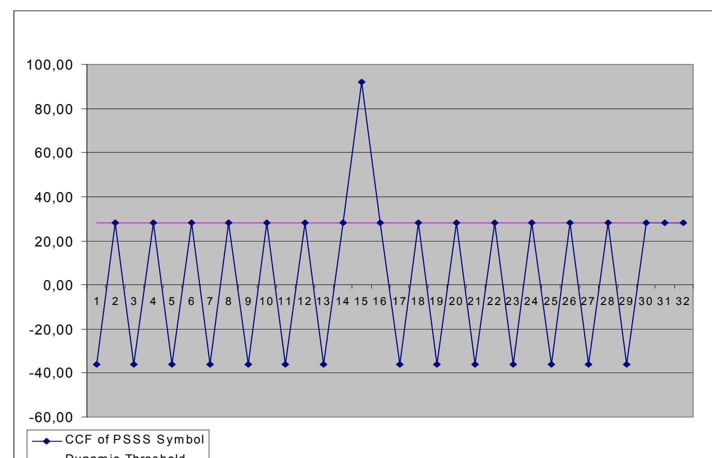


1. Align PSSS symbol maxima symmetrical to 0
2. Scale PSSS symbol to amplitude limit

Minimal Resolution after precoding: 5 bit

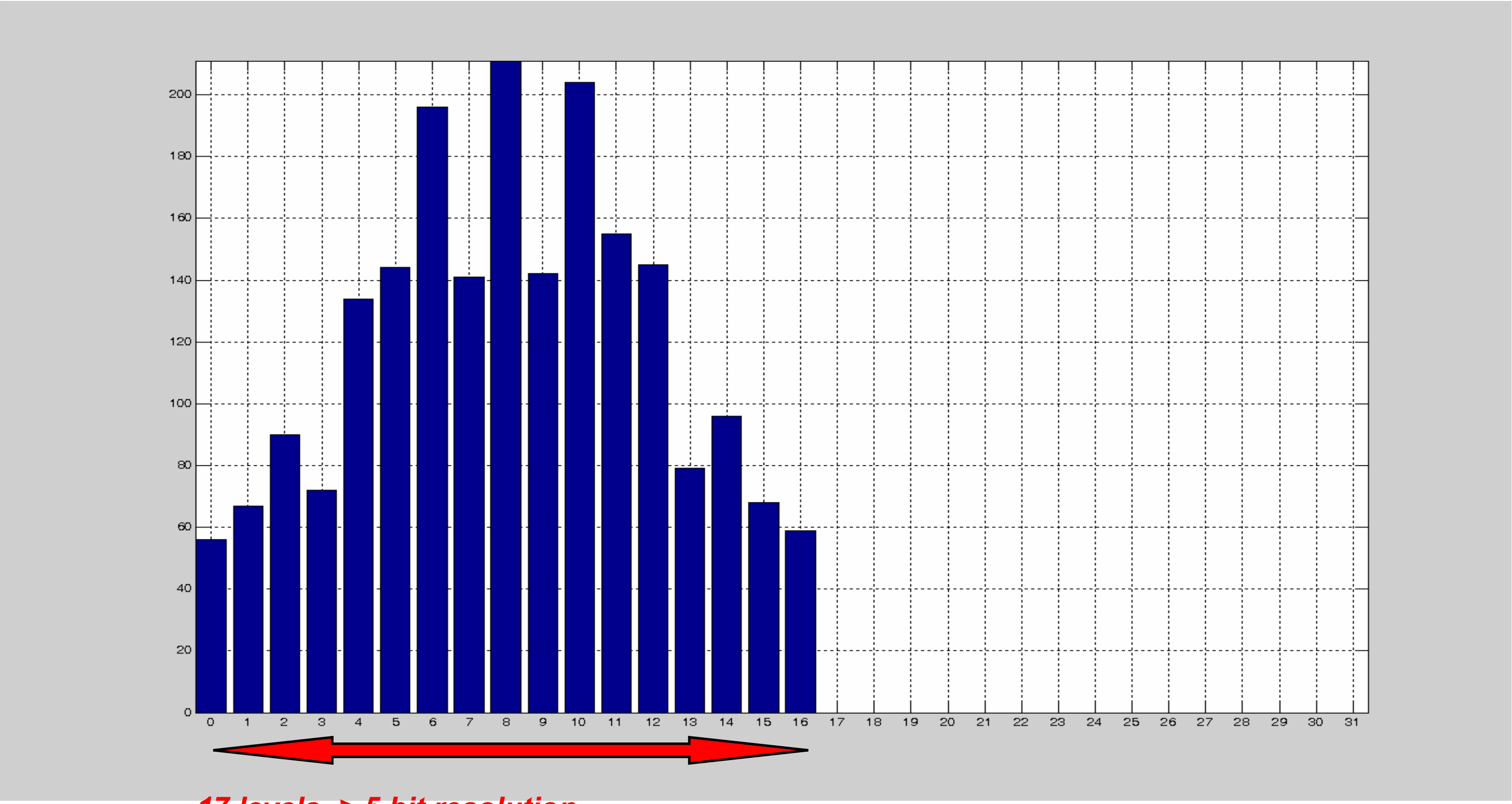
Note:

Higher resolution further improves performance, but does not limit interoperability



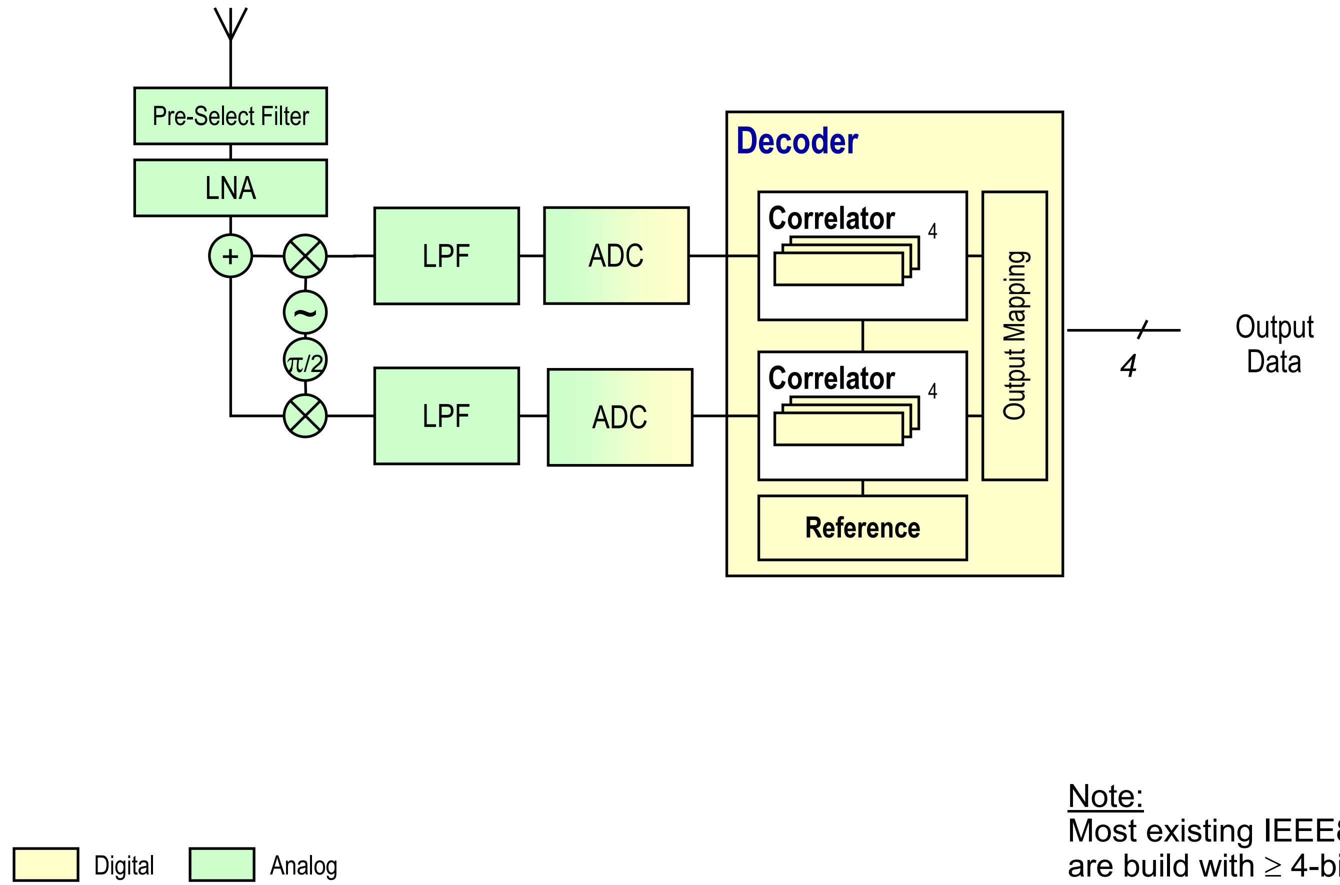
Tx **Rx**

PSSS Amplitude Histogram With Precoding



17 levels -> 5 bit resolution

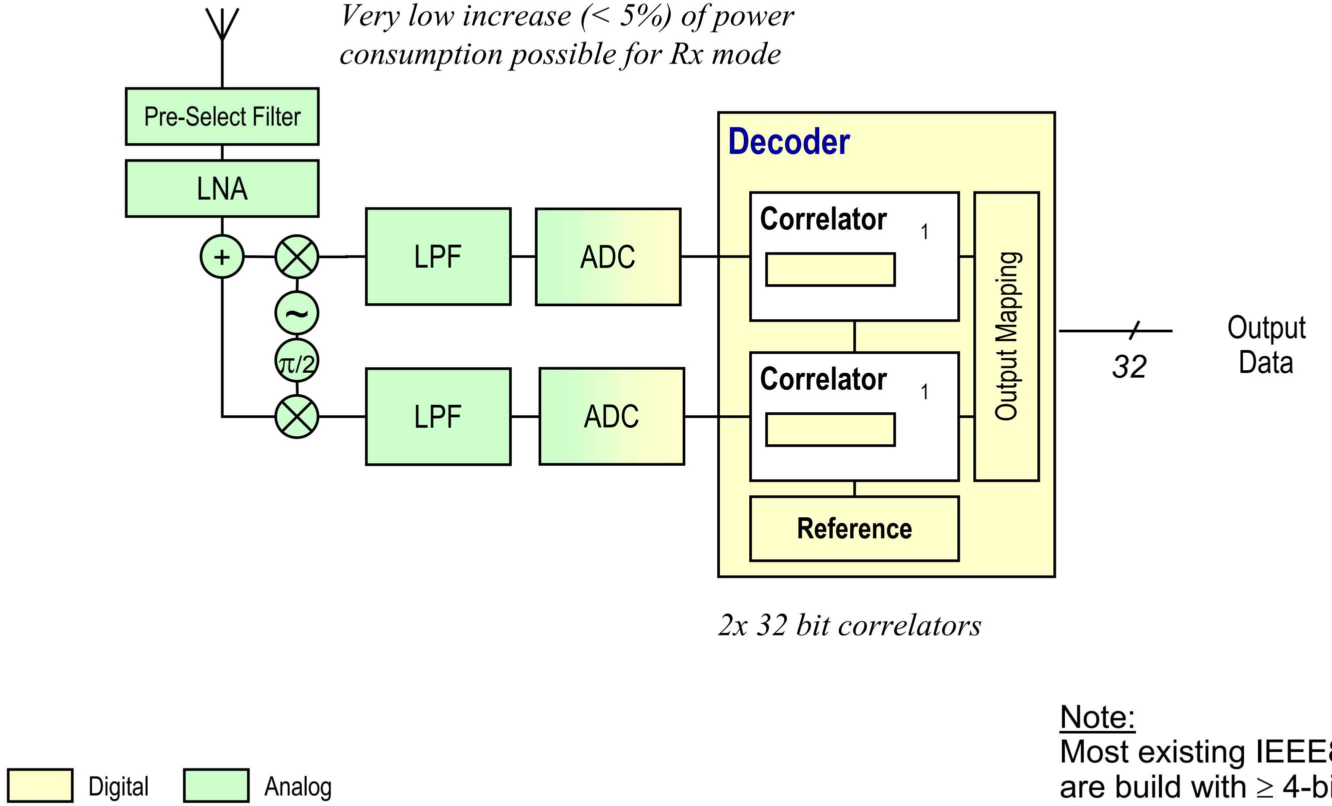
IEEE802.15.4-2003 2.4 GHz PHY – Rx architecture example (1/16 Bit/s/Hz)



Note:
Most existing IEEE802.15.4 2.4 GHz chips are build with ≥ 4-bit ADCs

PSSS - 8 Times parallel 2.4 GHz PHY derivate – Rx: Original O-QPSK / I/Q proposal (1/2 bit/s/Hz) – Digital correlation example

Very low increase (< 5%) of power consumption possible for Rx mode

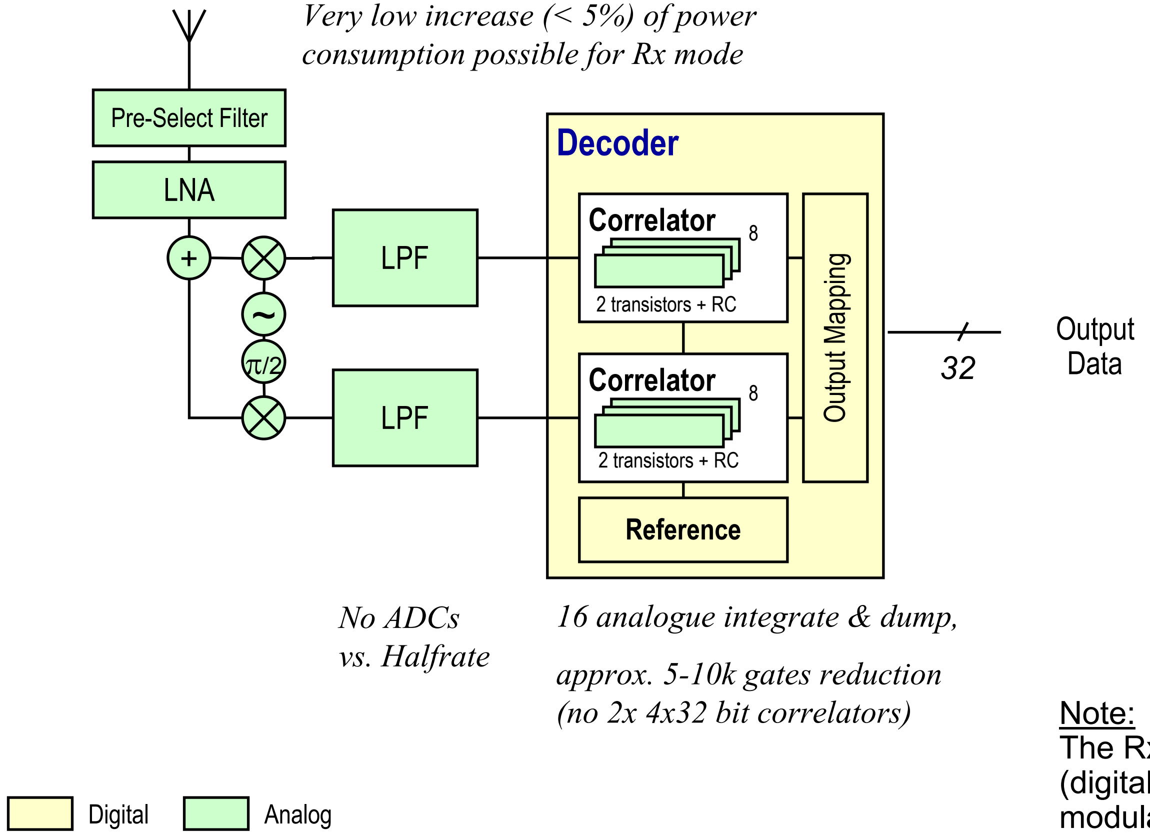


2x 32 bit correlators

Note:
Most existing IEEE802.15.4 2.4 GHz chips are build with ≥ 4-bit ADCs

PSSS - 8 Times parallel 2.4 GHz PHY derivate – Rx: Original O-QPSK / I/Q proposal (1/2 bit/s/Hz) – Analog correlation example

Very low increase (< 5%) of power consumption possible for Rx mode



No ADCs vs. Halfrate

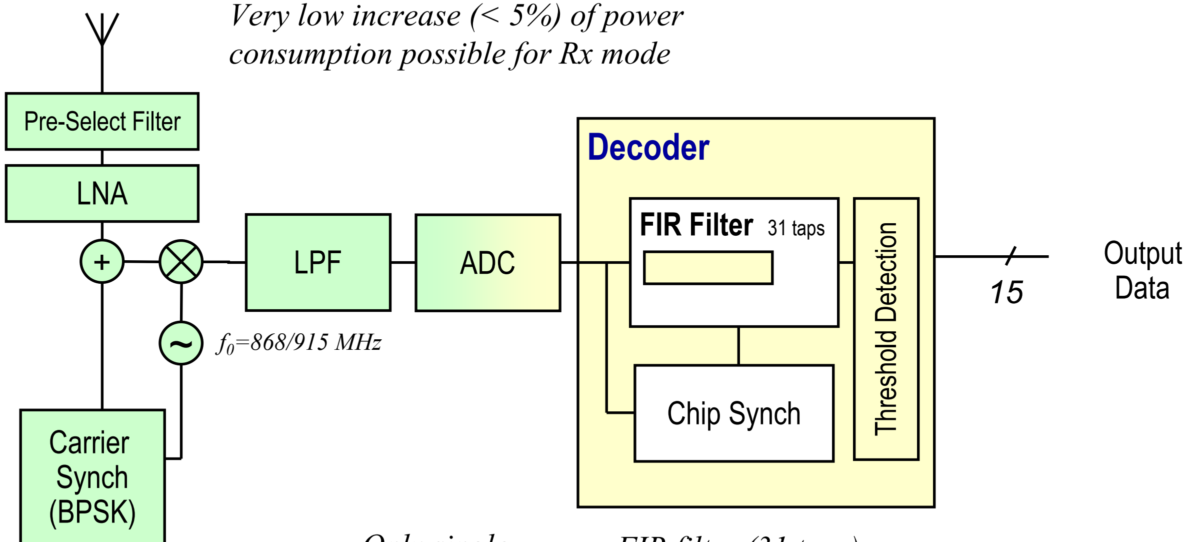
16 analogue integrate & dump,

approx. 5-10k gates reduction (no 2x 4x32 bit correlators)

Note:
The Rx example architectures shown (digital, analog, FIR correlator) and the modulation variant can be freely combined

PSSS - 8 Times parallel 2.4 GHz PHY derivate – Rx - BPSK/ASK option (15/32 bit/s/Hz) – FIR filter correlation example

Very low increase (< 5%) of power consumption possible for Rx mode



Only single ADC vs. Halfrate

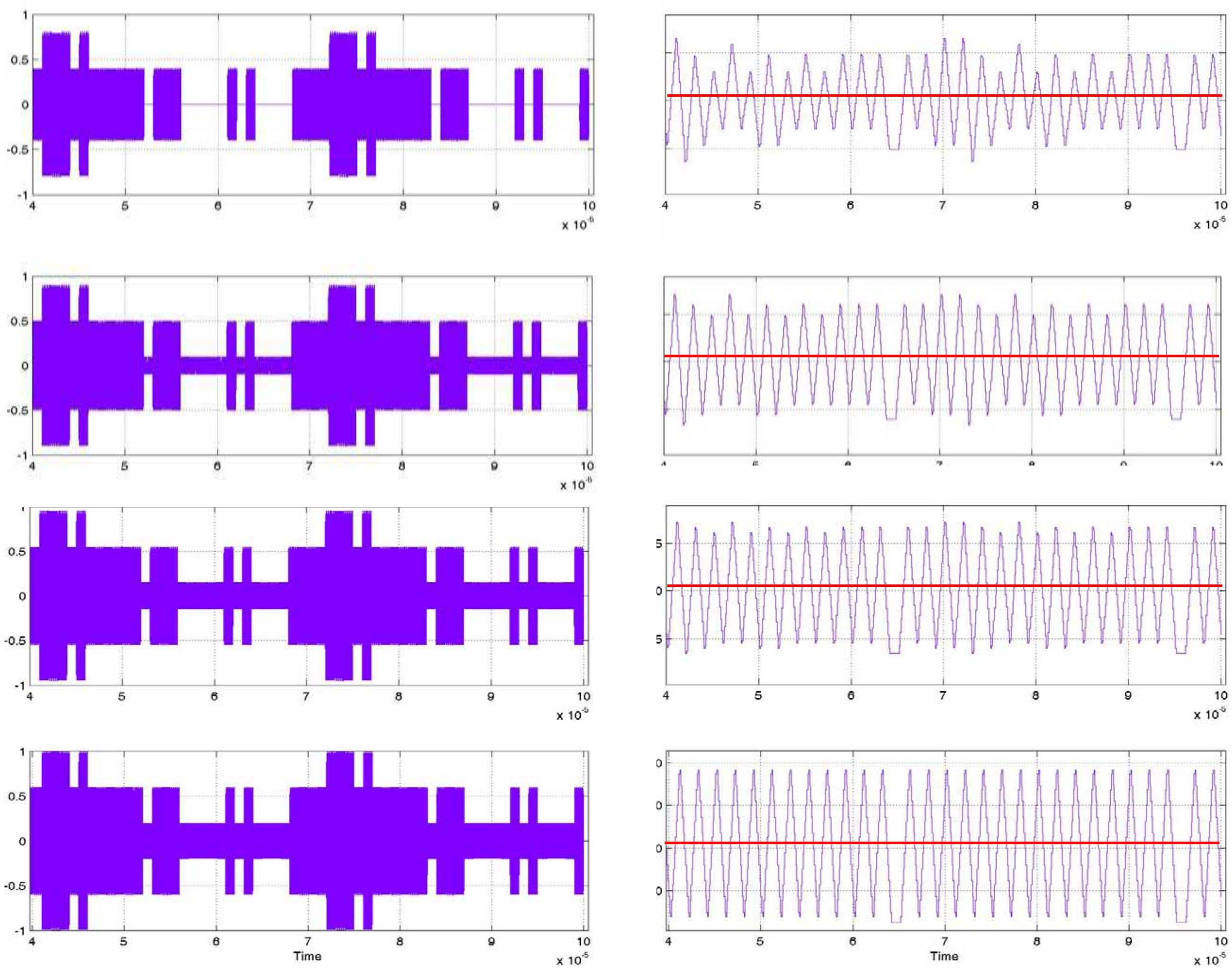
FIR filter (31 taps)

approx. 5-10k gates reduction vs. halfrate (no 2x 4x32 bit correlators)

Digital Analog

Linearity –
Transfer function for non-linear system simulated

Linearity – Simulation results



20% non-linearity

10% non-linearity

5% non-linearity

0% non-linearity

Detection threshold (for ‚0‘ or ‚1‘ data bits)