## Designing with EMC in Mind

# How to Apply Shielded Cables to Solve More Problems than You Create

Thomas A. Jerse

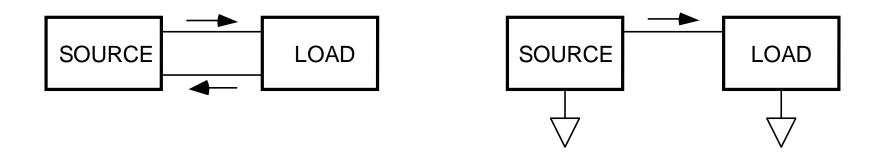




The Citadel Charleston, SC

#### The Job

Designing for electromagnetic compatibility (EMC) is primarily the orchestration of return current flow.

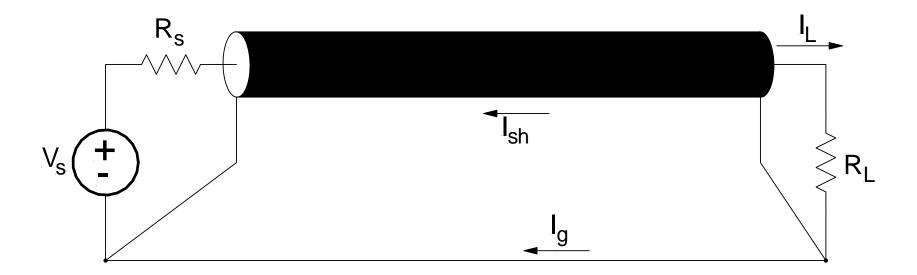


#### **Topics**

- Return Current Behavior
- Cable shields
- Cable grounding
- Ground-loop and common-mode coupling
- Common-mode emissions from cables
- Transfer impedance
- Bringing cables through shield walls

## The Key

The successful application of cables depends on a knowledge of the return current flow.



#### The Rules

 At low frequencies, the return current apportions itself among paths according to the resistances involved.

 At high frequencies, the return current apportions itself among paths according to the inductances involved.

#### The Good News

- The return current apportions itself in such a way that it incurs the smallest voltage drop possible on its way back to the source.
- That is, the return current seeks the minimum impedance.

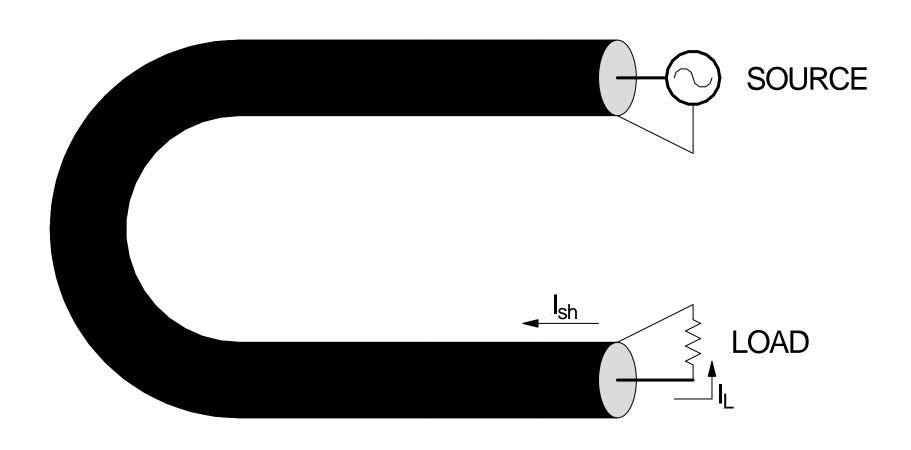
$$\vec{E} = -\nabla V \qquad P = \frac{1}{2} \Re e \left\{ V I^* \right\}$$

#### In Other Words

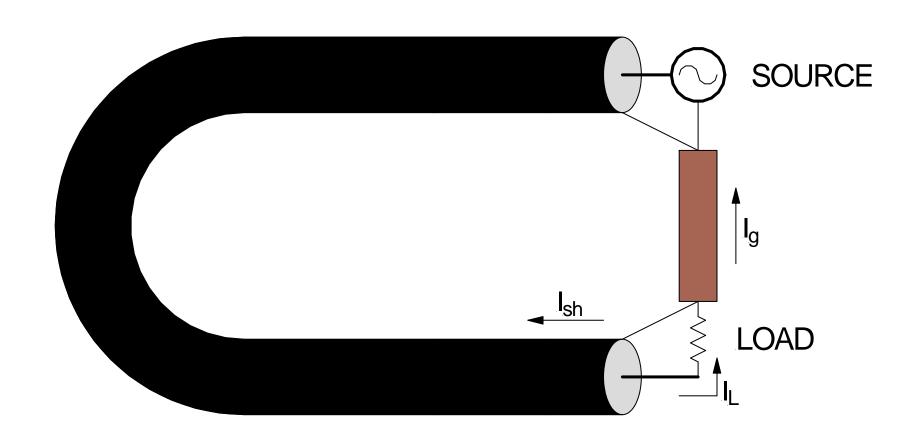
- At low frequencies, current apportions itself so that the net resistance seen in the aggregate of the paths it travels on the way back to the source is the smallest available.
- At high frequencies, current apportions itself so that the net inductance seen in the aggregate of the paths it travels on the way back to the source is the smallest available.

$$\psi = LI$$

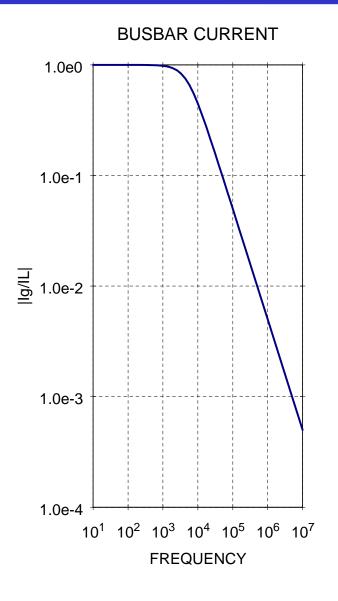
#### **Cable - One Return Path**

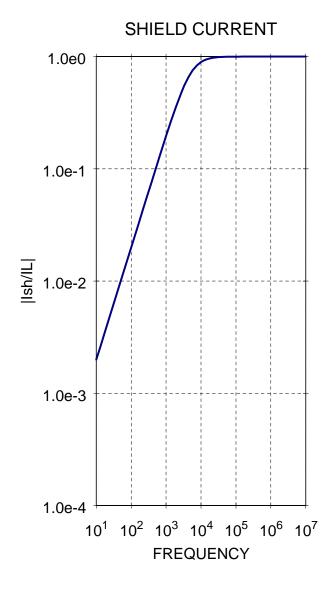


#### **Cable - Two Return Paths**

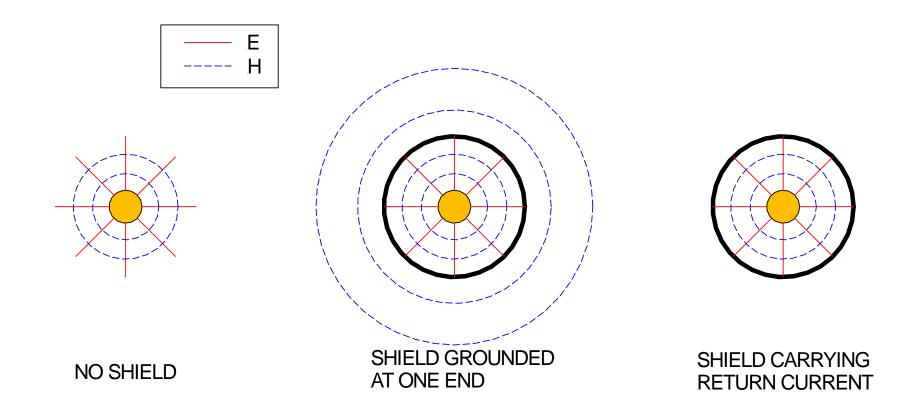


#### **Current Distribution**

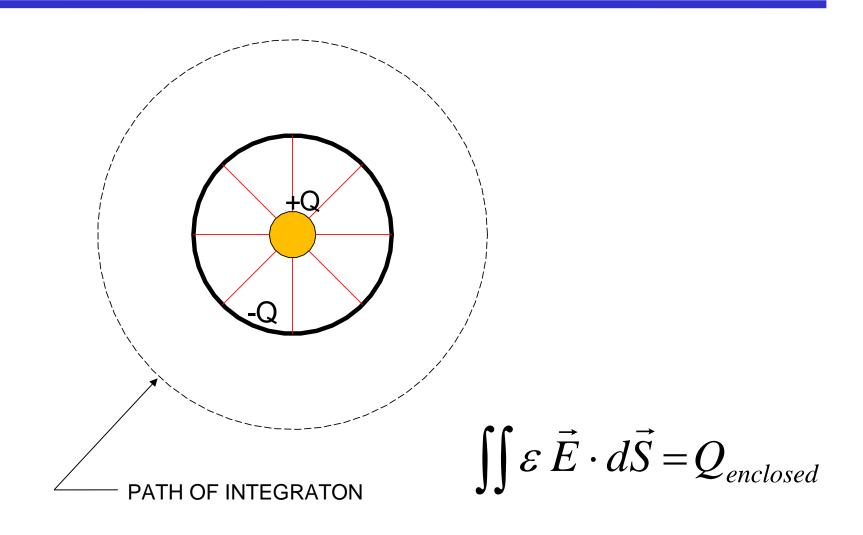




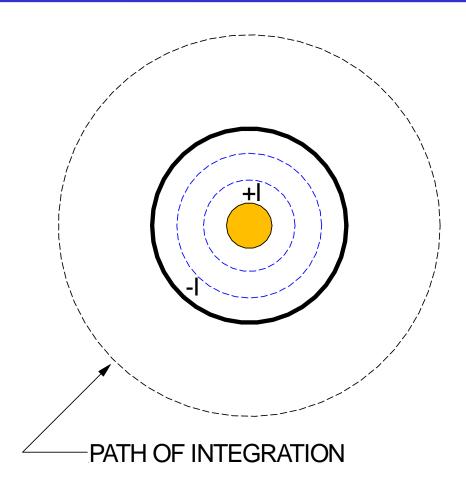
## **Shielded Cable**



## Gauss's Law

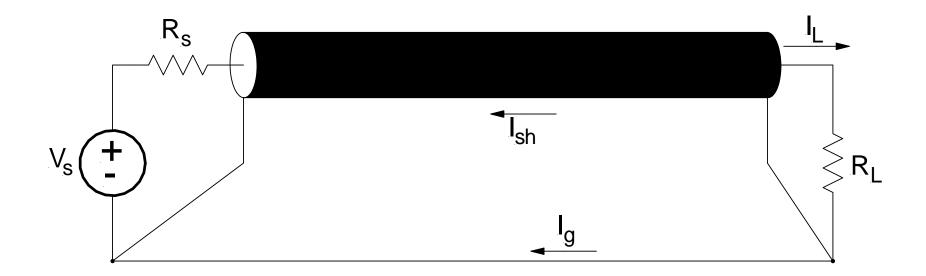


# Ampere's Law



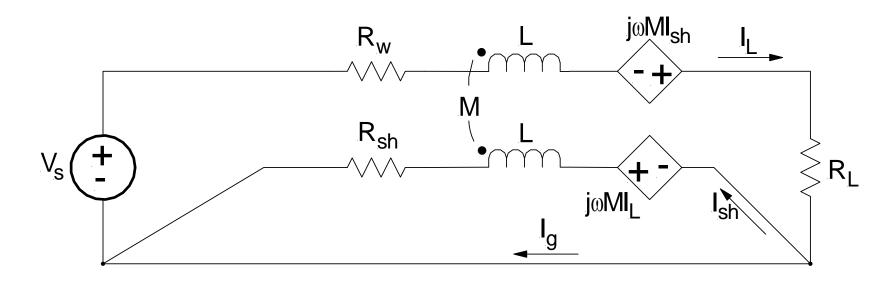
$$\oint \vec{H} \cdot d\vec{\ell} = I_{enclosed}$$

#### **Return Current Flow**



WHERE DOES THE RETURN CURRENT FLOW?

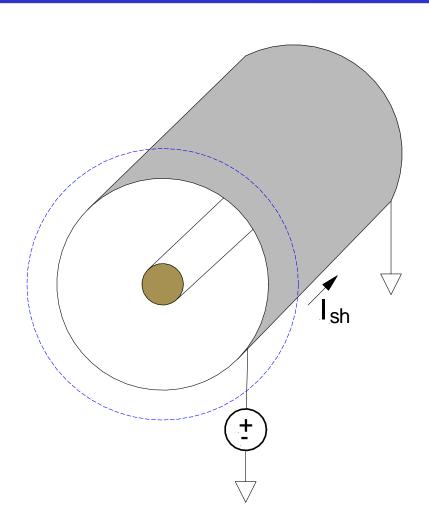
## **Analysis of Shield Cutoff**



$$I_{sh}R_{sh} + j\omega LI_{sh} - j\omega MI_{L} = 0$$

$$\frac{I_{sh}}{I_L} = \frac{j\omega M}{R_{sh} + j\omega L}$$

# **Magnetic Coupling**



M = L

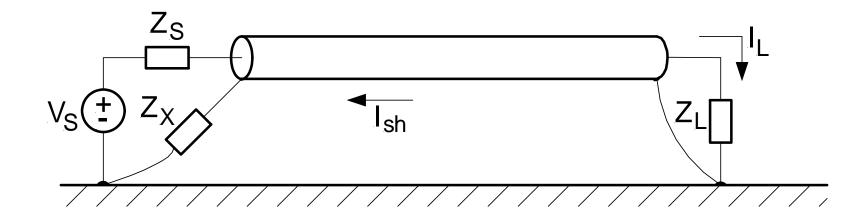
#### **Shield Current**

$$\frac{I_{sh}}{I_L} = \frac{j\omega L}{R_{sh} + j\omega L}$$

SHIELD CUTOFF FREQUENCY

$$f_c = \frac{R_{sh}}{2\pi L}$$

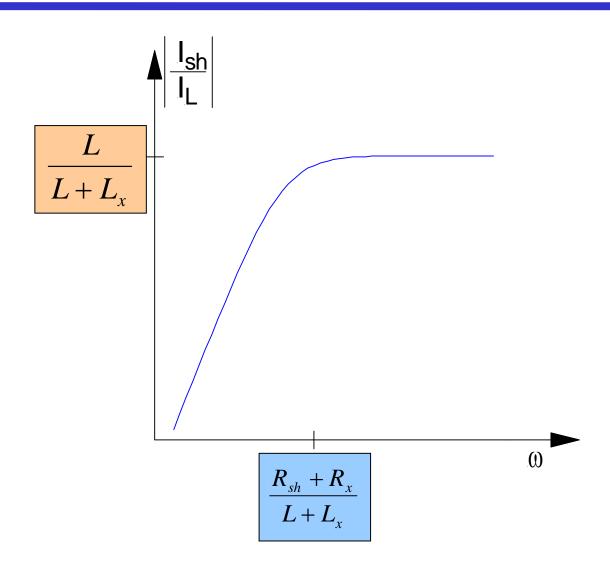
## **Poor Shield Grounding**



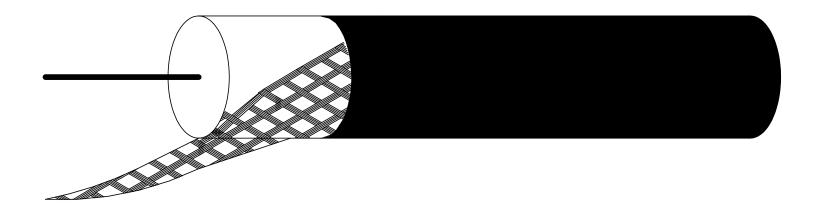
$$Z_{x} = R_{x} + j\omega L_{x}$$

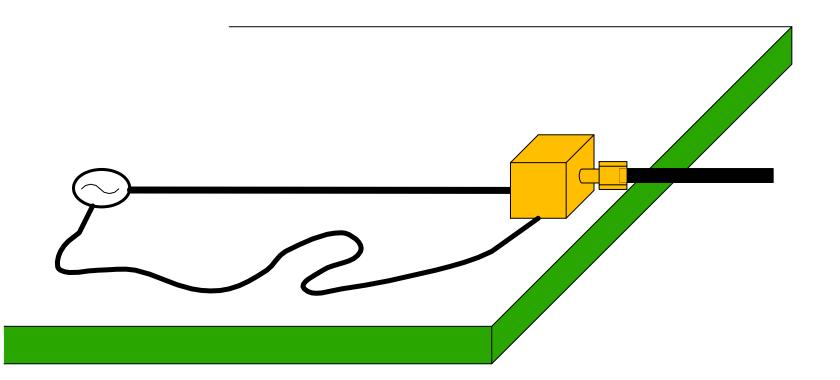
$$\frac{I_{sh}}{I_L} = \frac{j\omega L}{R_x + R_{sh} + j\omega (L + L_x)}$$

# **Shield Current**

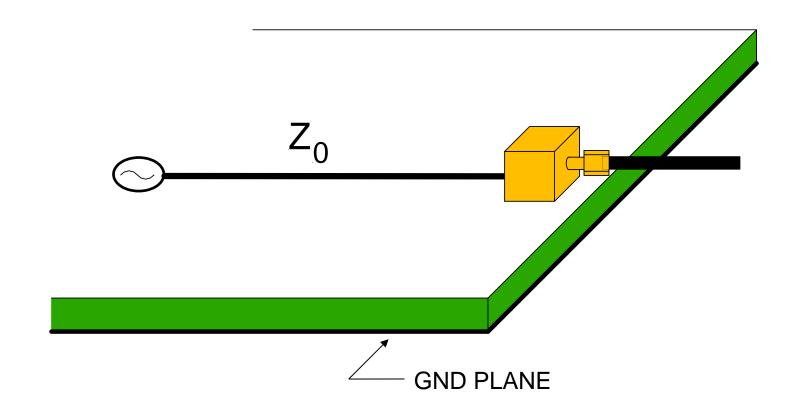


# **Pigtail**





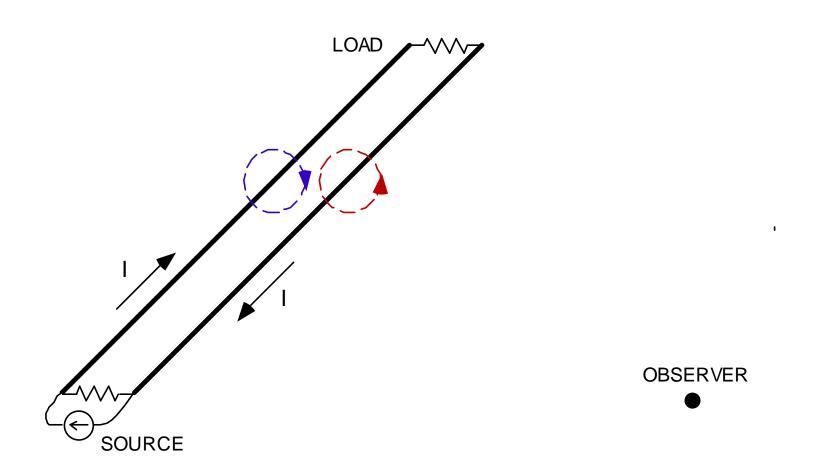
# Minimum Z<sub>x</sub>



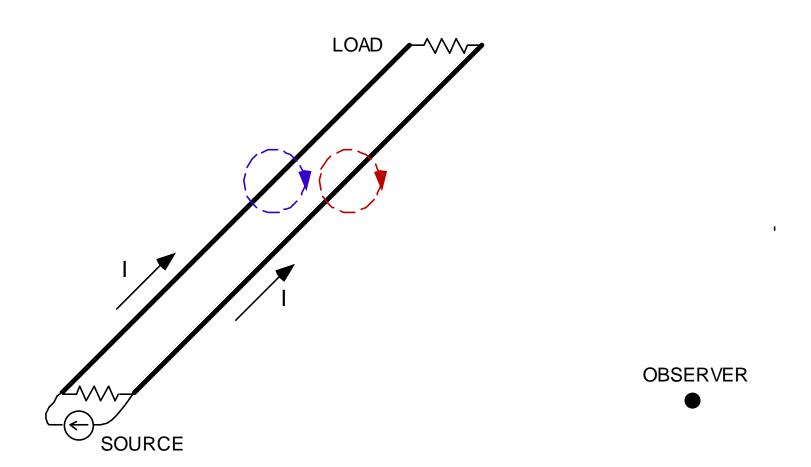
#### **Common-Mode Radiation**

- In a system, cables are often the principal radiators of CM emissions.
- Sources of CM current
  - Poor cable grounding
  - Connection to a noisy ground

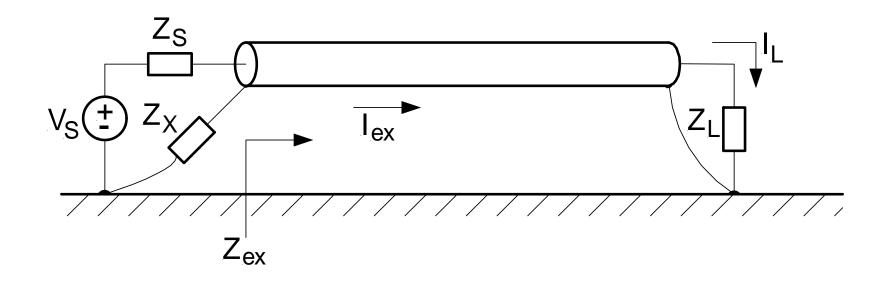
#### **Differential Mode Radiation**



## **Common-Mode Radiation**



# **Poor Shield Grounding**



I<sub>ex</sub> IS A COMMON-MODE CURRENT.

## At High Frequencies

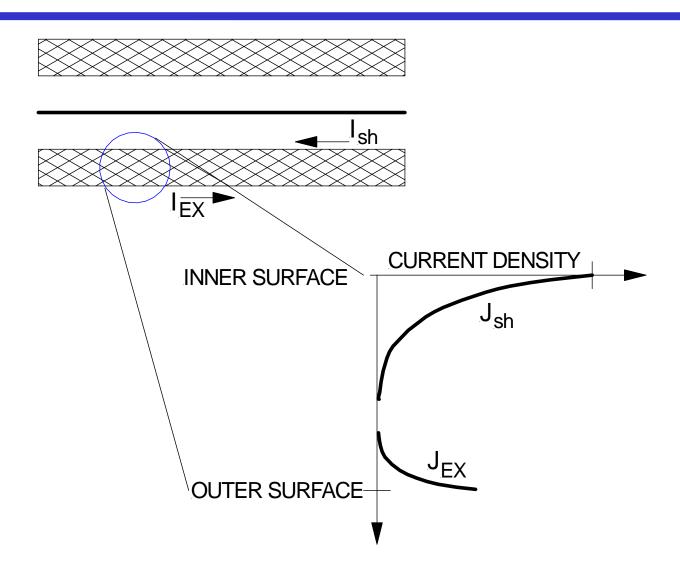
- The skin effect causes the coax to behave like a triax.
- I<sub>ex</sub> will flow on the outside of the shield.

#### **Skin Effect**

- Describes tendency of high frequency current to ride on conductor surface
- Skin depth,  $\delta$ , represents distance where current density is attenuated by  $e^{-1}$ .

$$\delta = \sqrt{\frac{1}{\pi f \sigma \mu}}$$

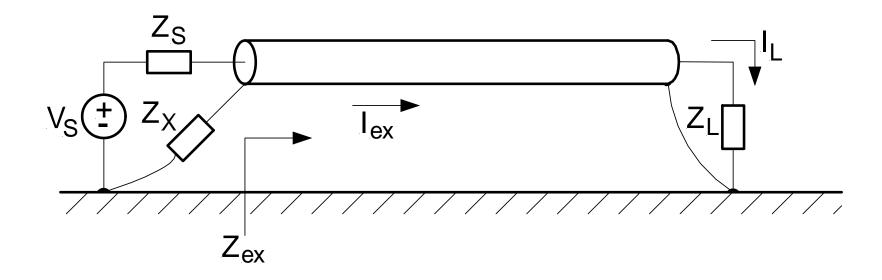
# **Separation of Currents**



# 80-dB Separation Frequency



# **Poor Shield Grounding**



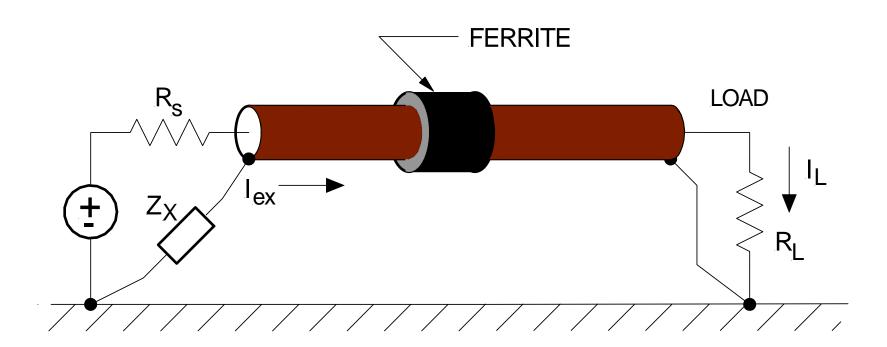
$$I_{ex} = I_L \frac{Z_x}{Z_x + Z_{ex}}$$

# $Z_{ex}$

- Function of frequency with numerous resonances.
- Depends on the cable layout, particularly its position with respect to any ground planes.

**FIX CABLES IN POSITION** 

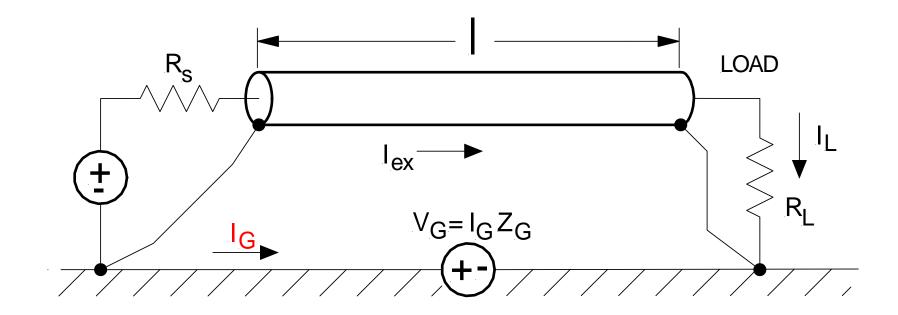
# Raising Z<sub>ex</sub>



## **Differential Ground Voltage**

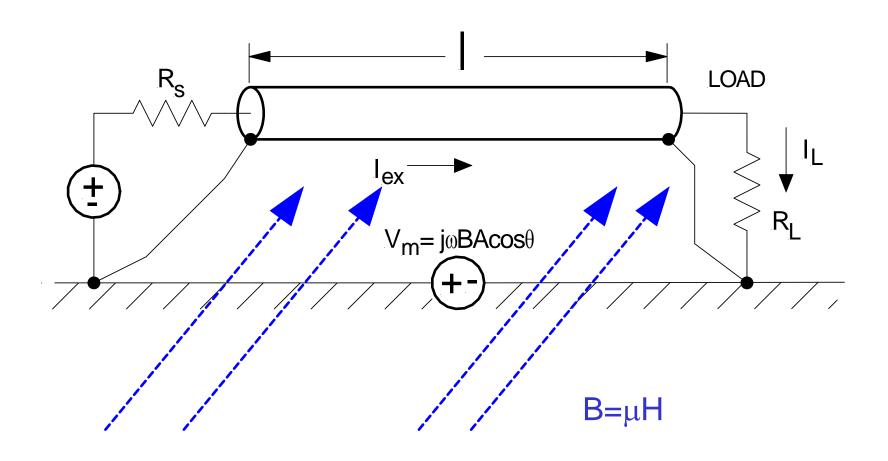
- The ground potential at the two ends of the cable may differ.
- Symptom—noise voltage coupled across load
- Causes
  - Ground loop coupling
  - Common-mode coupling

# **Ground-Loop Coupling**

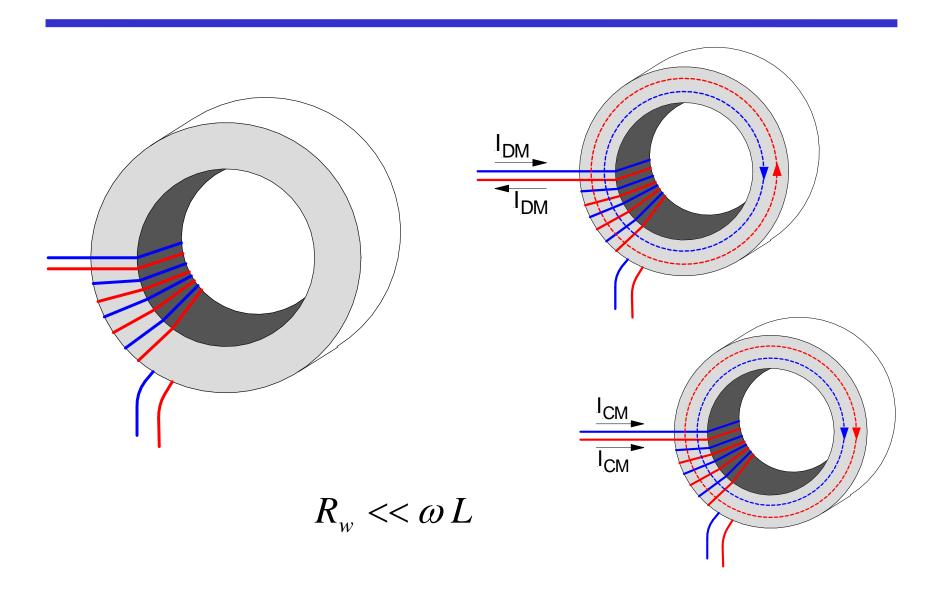


$$I_{L} = \frac{V_{g}}{\left(R_{s} + R_{L}\right)\left(1 + j\omega \frac{L}{R_{sh}}\right)}$$

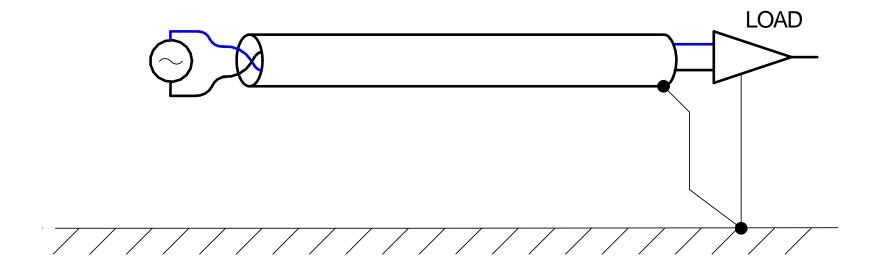
# **Common-Mode Coupling**



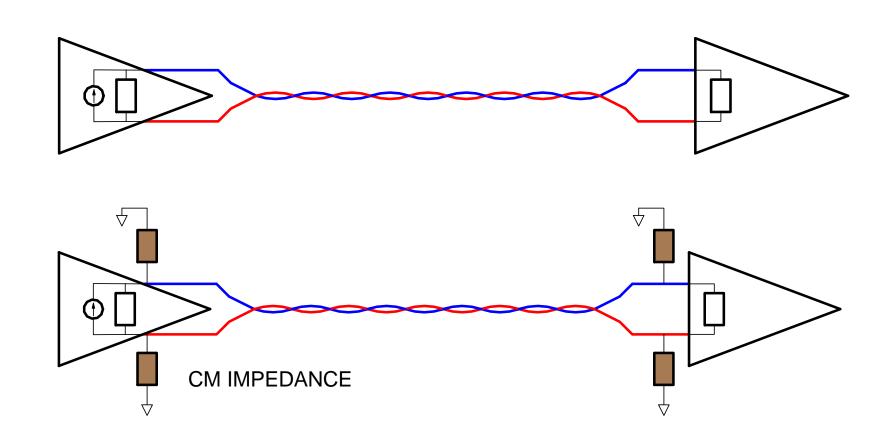
## **Common-Mode Choke**



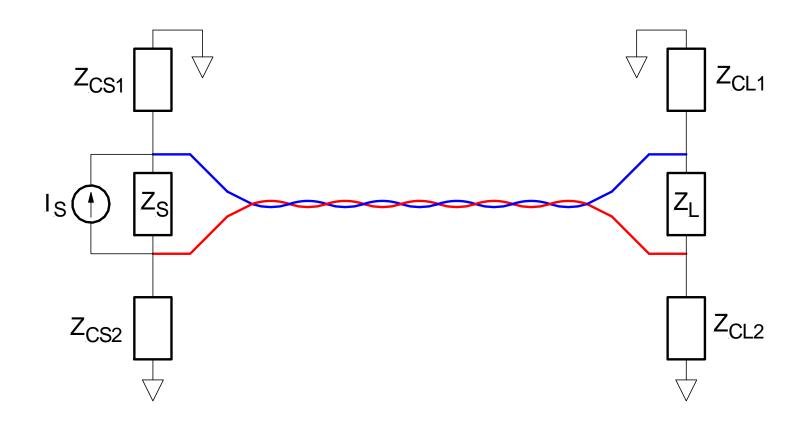
## **Shielded Twisted Pair**



# Differential to Common-Mode Conversion

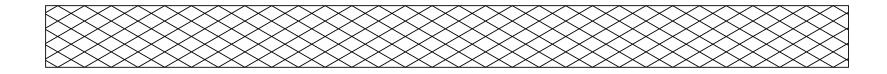


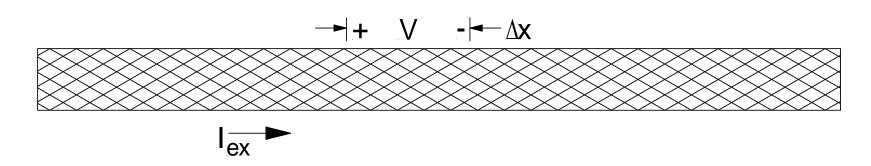
## **DM-to-CM Conversion**



$$I_{\mathit{CM}} = 0$$
 when  $Z_{\mathit{CS1}} Z_{\mathit{CL2}} = Z_{\mathit{CS2}} Z_{\mathit{CL1}}$ 

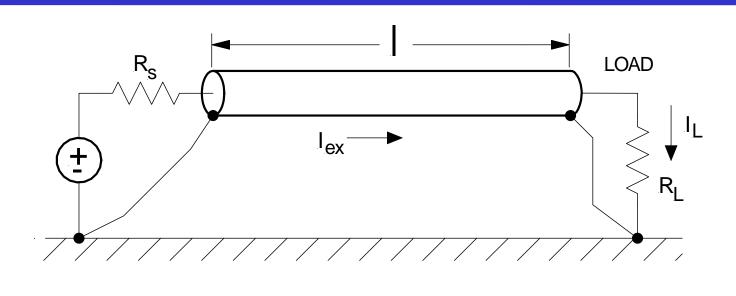
# Cable Transfer Impedance

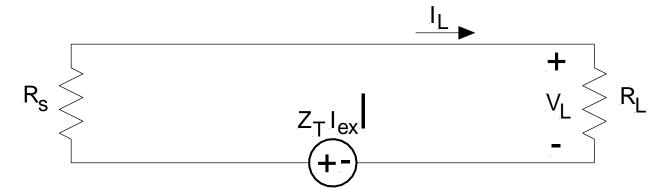




$$Z_T = \frac{V}{I_{ex} \Delta x}$$

# **Coupling Calculation**





# **Shielding Effectiveness Measures**

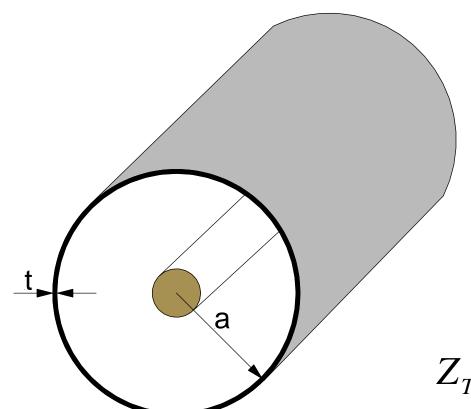
#### Traditional

$$S = 20 \log_{10} \left( \frac{I_{ex}}{I_L} \right) \text{ [dB]} \qquad S = 20 \log_{10} \left( \frac{Z_T \ell}{R_s + R_L} \right)$$

#### Transfer

$$Z_{TL} = 20\log_{10}\left(\frac{V_L}{I_{ex}}\right) \text{ [dB}\Omega] \qquad Z_{TL} = 20\log_{10}\left(\frac{Z_T \ell R_L}{R_s + R_L}\right)$$

## **Z**<sub>T</sub> of a Solid Shield



DC RESISTANCE

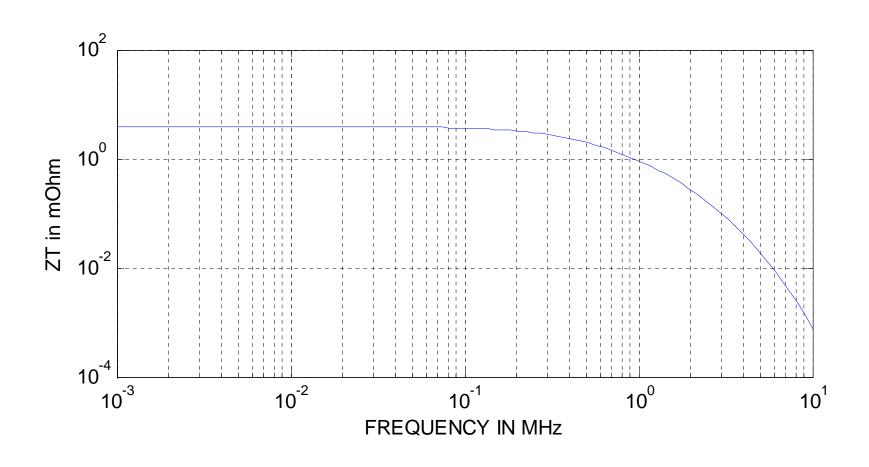
$$R_0 = \frac{1}{2\pi \, at \, \sigma} \quad [\Omega/\mathrm{m}]$$

a = MEAN RADIUS

$$Z_{T} = R_{0} \frac{(1+j)\frac{t}{\delta}}{\sinh\left[(1+j)\frac{t}{\delta}\right]}$$

 $\delta$  = SKIN DEPTH

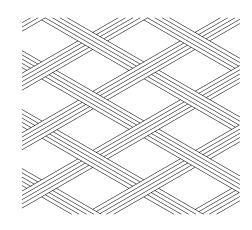
# Solid Shield Example



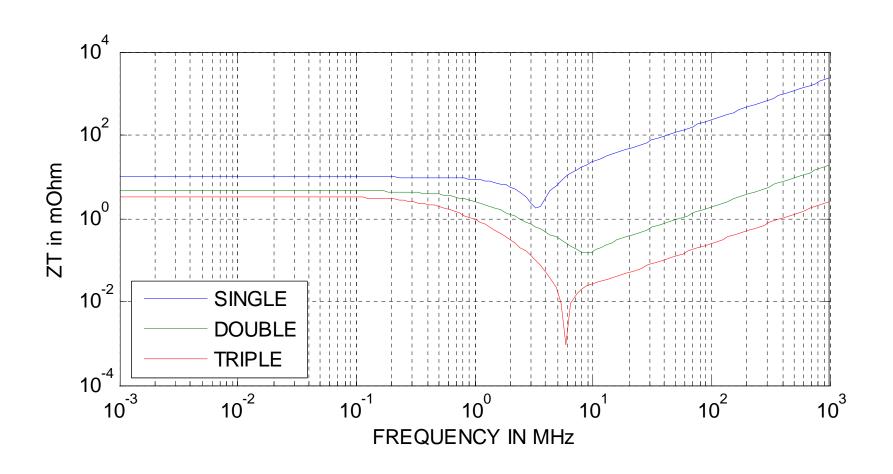
#### **Braided Shield**

- Coupling through rhombic apertures modeled by mutual inductance,  $M_{12}$ .
- More susceptible to corrosion

$$Z_{T} = R_{0} \frac{(1+j)\frac{t}{\delta}}{\sinh\left[(1+j)\frac{t}{\delta}\right]} + j\omega M_{12}$$

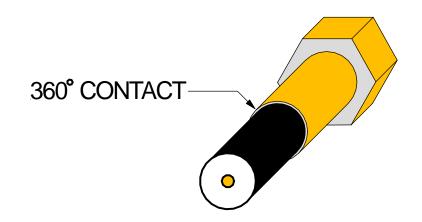


#### **Braided Shield**

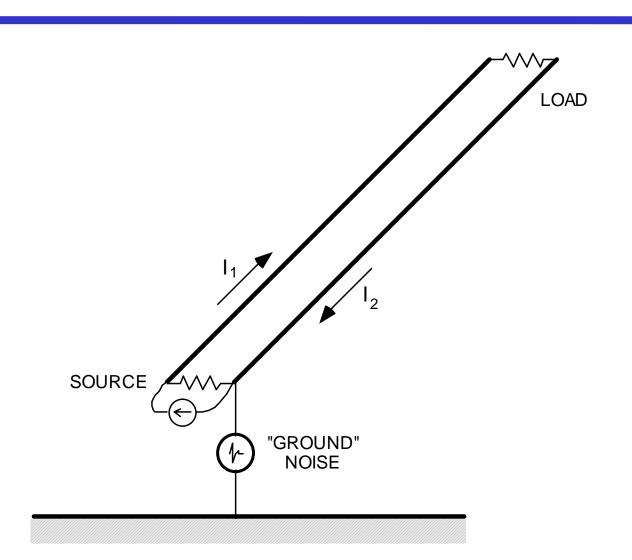


#### **Connectors**

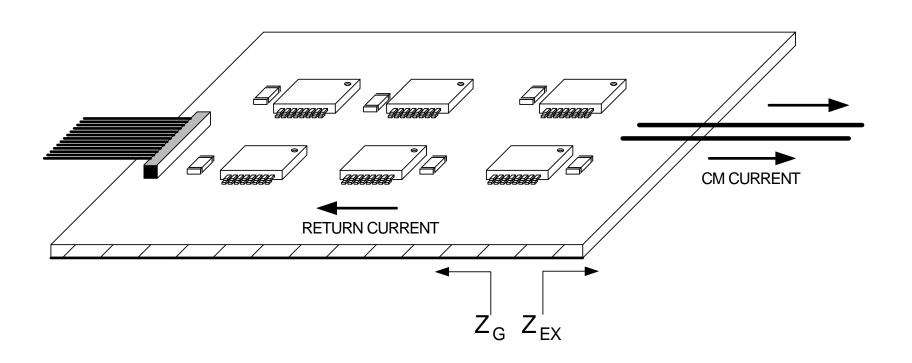
- Connectors also have transfer impedances;
  - e.g., BNC,  $R_0 = 2 \text{ m}\Omega$  ,  $M_{12} = 5 \text{ pH}$
- Cable shield must have a 360° lowresistance contact to its connectors.



### "Ground" Noise



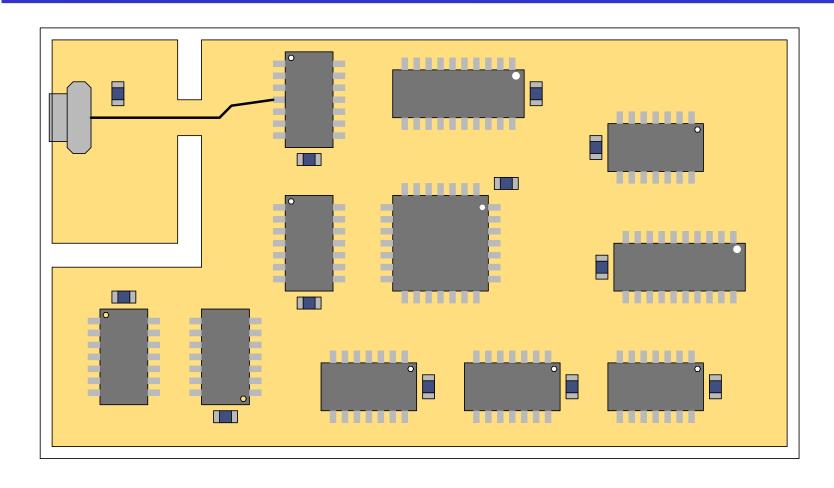
#### **Common-Mode Excitation**



ATTACH CABLE TO QUIETER GROUND REGIONS.

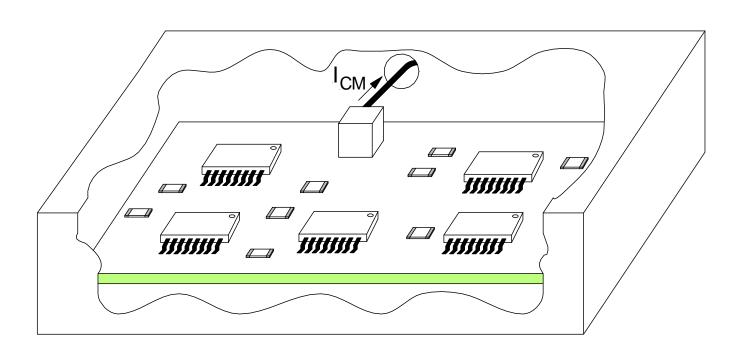
$$I_{CM} = I_{RTN} \frac{Z_{G}}{Z_{G} + Z_{EX}}$$

# Cable Grounding "Island"

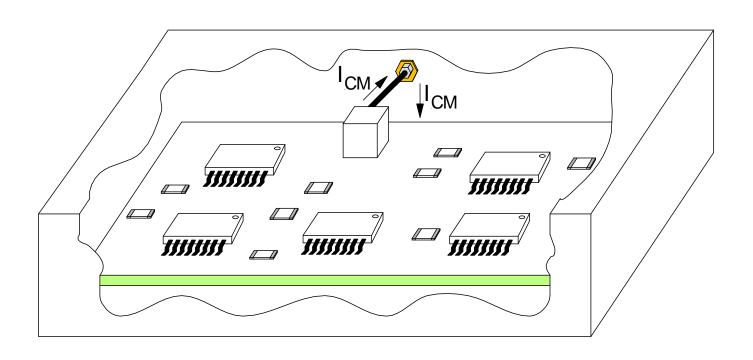


NOT AS EASY AS IT LOOKS. SIGNAL TRACES SHOULD NOT CROSS GROUND CUTOUT.

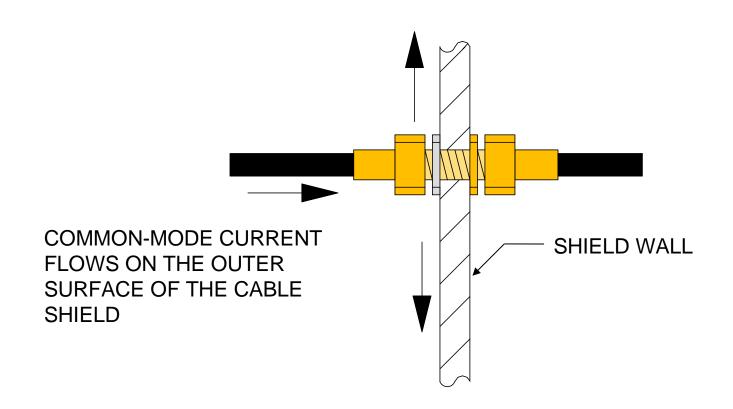
## Bringing a Cable through a Shield Wall



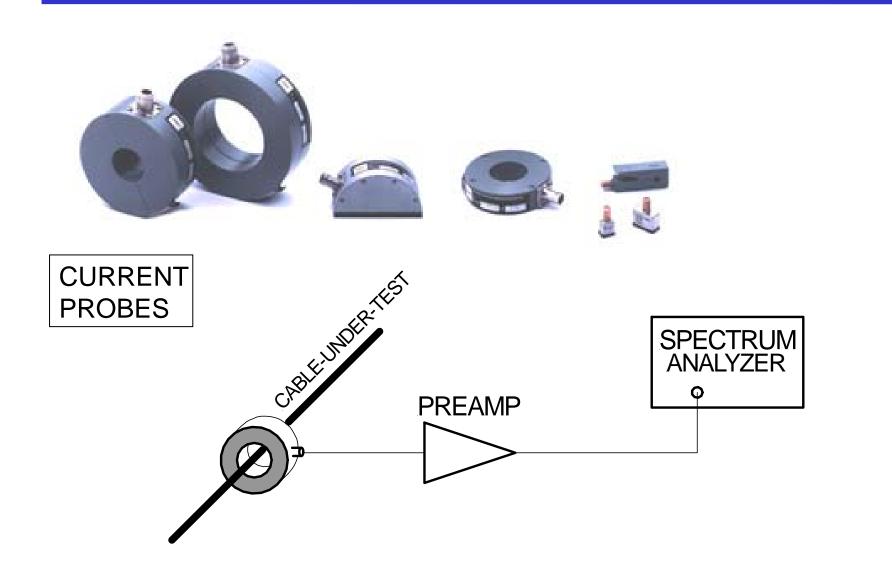
## **Bulkhead Connector**



# **Stripping Common-Mode Current**



## **Measuring Common-Mode Current**



## **Summary of Concepts**

- Return current seeks the minimum impedance.
- Above the 5x shield cutoff frequency, nearly all the return current flows in the shield of a well grounded cable.
- Common-mode current induced by poor cable grounding flows on the outer surface of a cable shield.

## **Summary of Concepts**

- A cable performs no better than its connectors.
- Terminate cable shields 360° to their connectors.
- Connecting a shield to a noisy ground produces CM current on the outer surface of the shield.
- Cables are best brought through a shield wall using a bulkhead connector.

#### **Books**

- Coupling to Shielded Cables
  E.F. Vance, John Wiley & Sons, 1978.
- Coupling of External Electromagnetic Fields to Transmission Lines
   A.A. Smith, Jr., John Wiley & Sons, 1977.

#### Checklist

- Where does the return current flow?
- Where does the return current flow?