



Low-Power, High-Bandwidth, and Ultra-Small Memory Module Design

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Overview: This work proposes a novel DRAM module and interconnect architectures in an attempt to improve computing energy use and performance. A low cost advanced packaging technology is used to propose an 8 die and 32-die memory module. The 32-die memory module measures less than 2 cm³. The size and packaging technique allow the memory module to consume less power than conventional module designs. A 4 Gb DRAM architecture utilizing 64 data pins is proposed. The DRAM architecture is inline with ITRS roadmaps and can consume 50% less power while increasing bandwidth by 100%. The large number of data pins are supported by a low power capacitive-coupled interconnect. The receivers developed for the capacitive interface were fabricated in 0.5 μm and 65 nm CMOS technologies. The 0.5 μm design operated at 200 Mbps, used a coupling capacitor of 100 fF, and consumed less than 3 pJ/bit of energy. The 65 nm design operated at 4 Gbps, used a coupling capacitor of 15 fF, and consumed less than 15 fJ/bit and order of magnitude smaller consumptions than previously reported receiver designs.

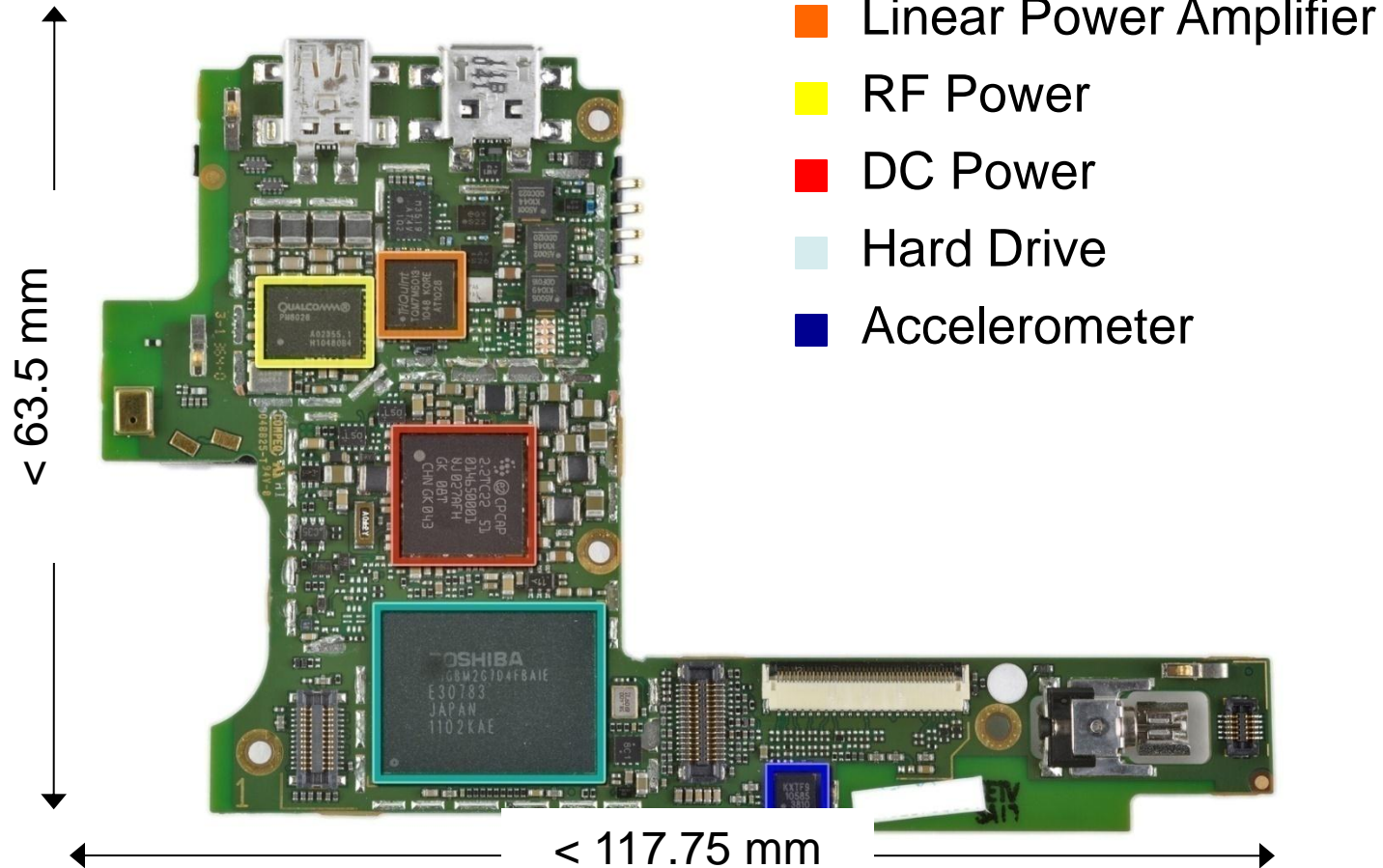


Why is Power Such a Big Deal?

- Let's say that at any given time there are, at least, 1,000,000 people playing World of Warcraft (WoW, a very reasonable assumption)
- Further let's say that the power consumed by each of these players for: processor, memory (DRAM), computer fan, hard disk drive, monitor(s), modem, remote servers, communication channels (e.g., satellites links), cooling, etc. is 1000 Watts (again a very reasonable assumption)
- More than 1 GW of power is needed at any time for people to play WoW. This is the amount of power generated by a small nuclear power plant!

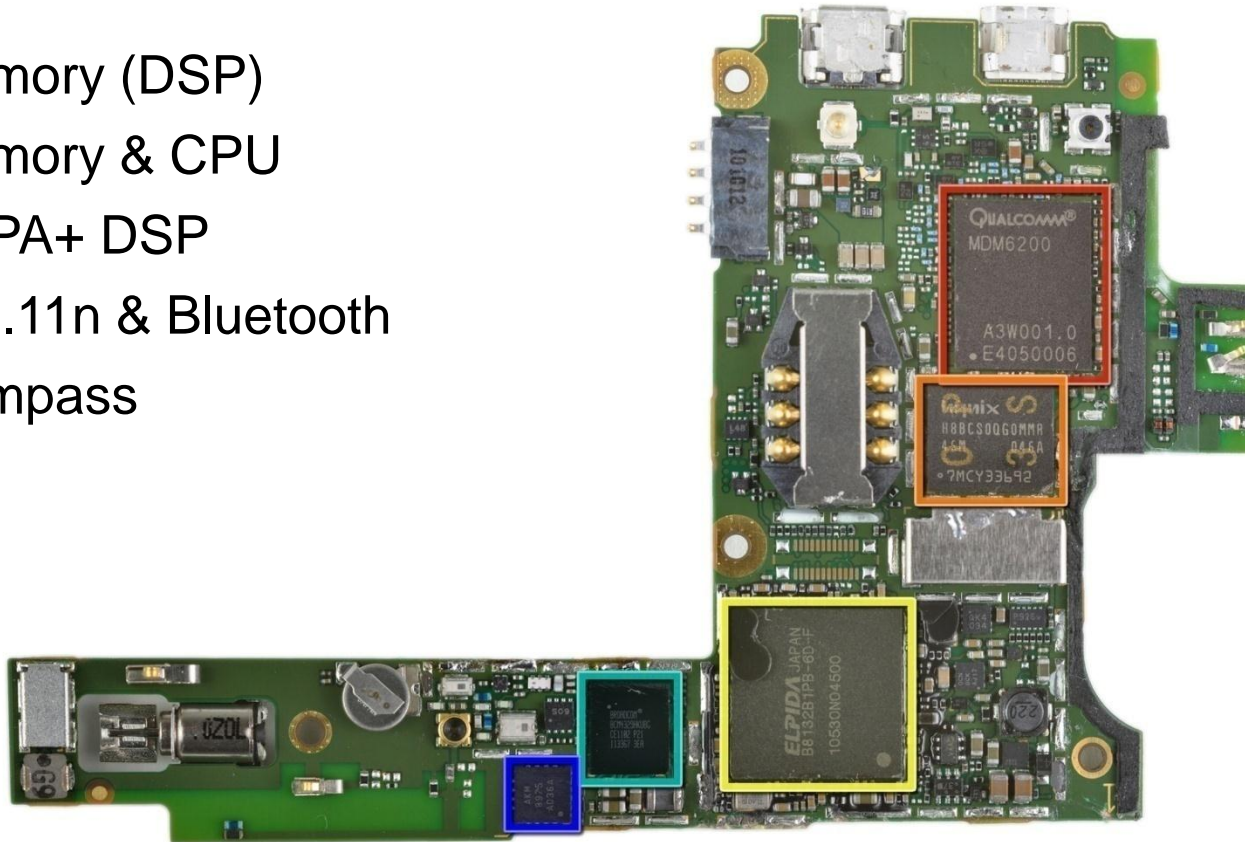
Mobile Platform

- Motorola Atrix (Front) found in the Google Droid



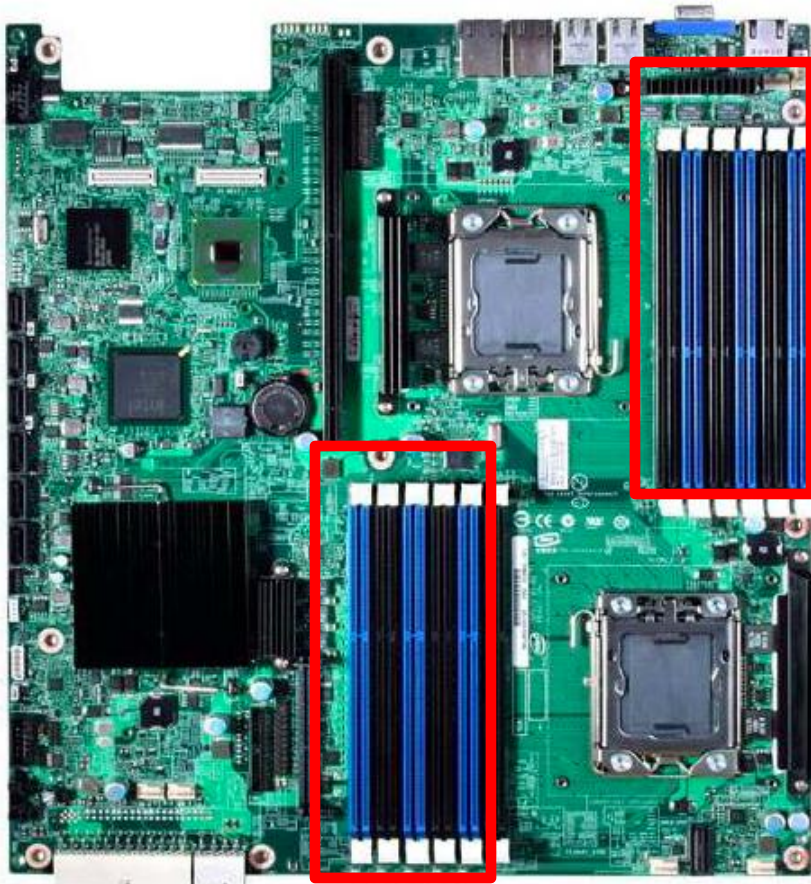
Mobile Platform

- Motorola Atrix (Back)
 - Memory (DSP)
 - Memory & CPU
 - HSPA+ DSP
 - 802.11n & Bluetooth
 - Compass



Server Platform

- Intel Server Board S5502UR



■ Memory Slots



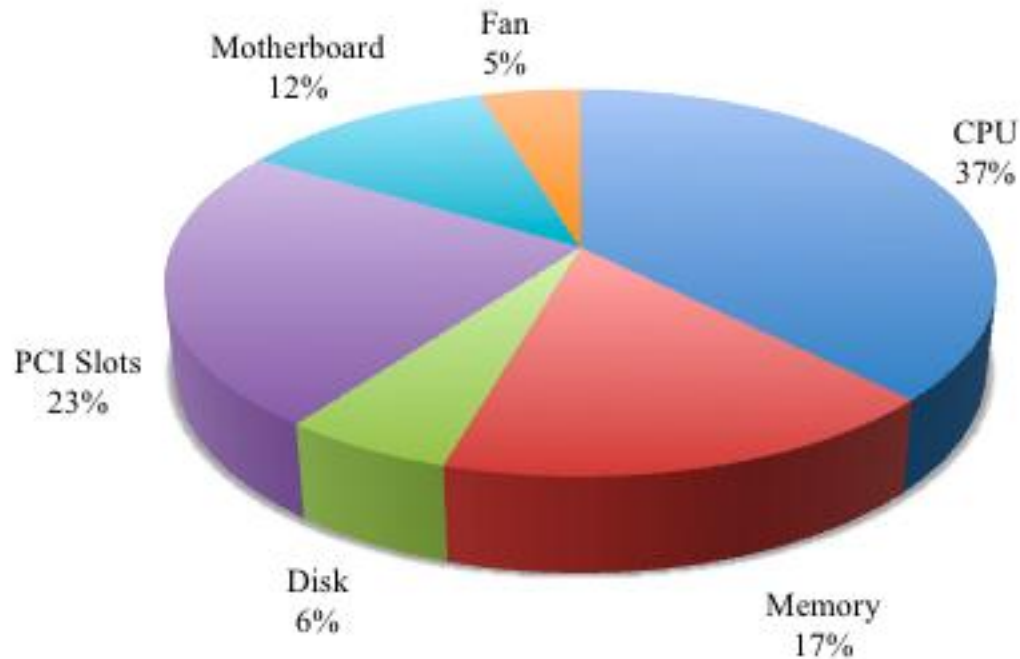
Organization

- Main Memory Limitations
- Nano-Module
- Wide I/O DRAM Architecture
- High Bandwidth Interconnect
- Hybrid Memory Cube

Main Memory Limitations

- Datacenter sparsity masked power limitations
 - Power trend: Energy consumption doubled every 5 years
- Historical server power
 - ~50 W in 2000
 - ~250 W in 2008
- Server power breakdown
 - CPU: 37%, Memory: 17%
 - Trend is Memory power > CPU power
- Main memory power
 - More die per module
 - Less modules per channel
 - Higher bandwidth

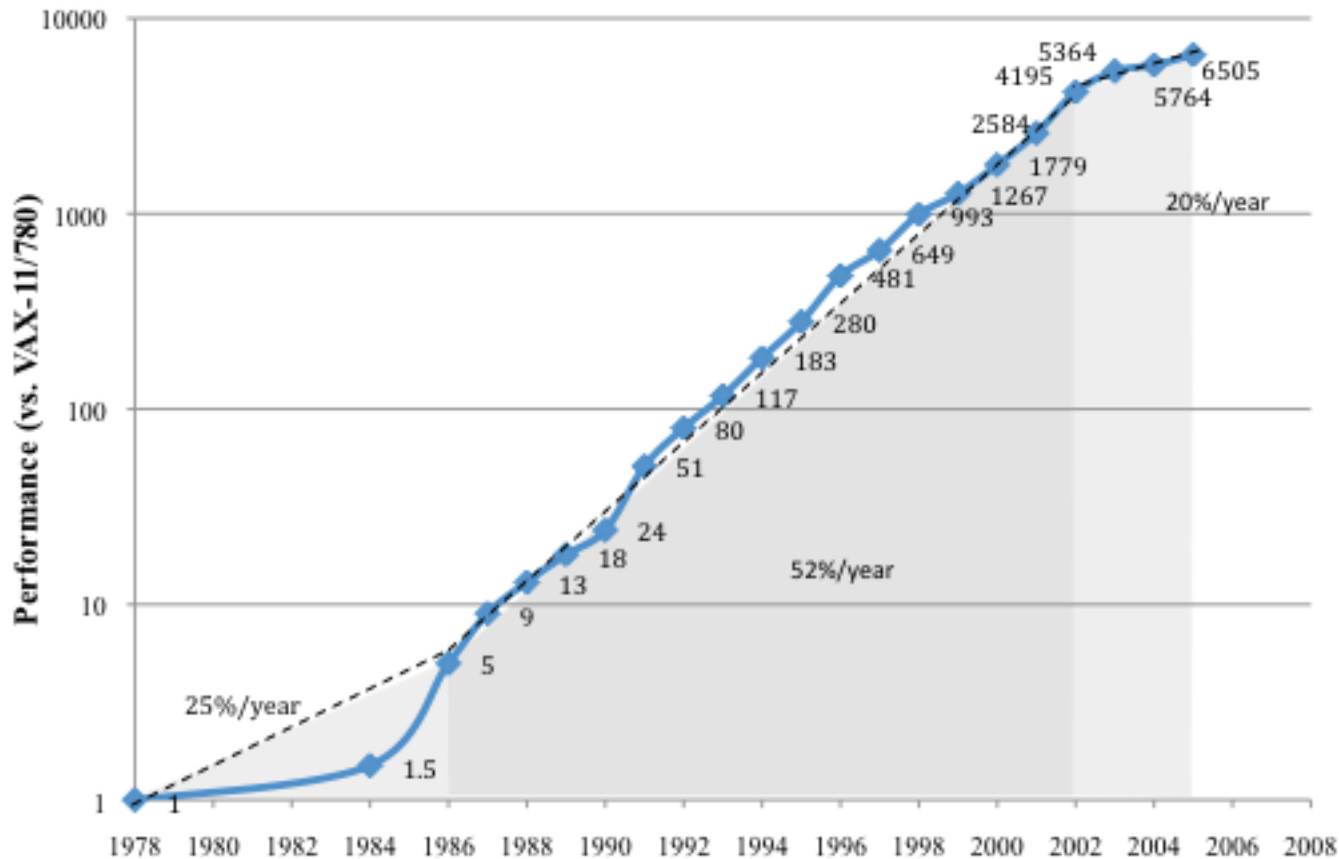
Main Memory Limitations



Main Memory Limitations

- CPU power wall
 - Voltage scaling reached its limit
 - Multiple cores supplement performance gains
 - No “multi-core” for DRAM
- DRAM voltage scaling reaching its limit
 - Current rate increase $>$ voltage reduction rate
 - Power increasing
- DRAM pre-fetch
 - Memory core operates at slower frequency
 - High power I/O devices and data-path

Main Memory Limitations



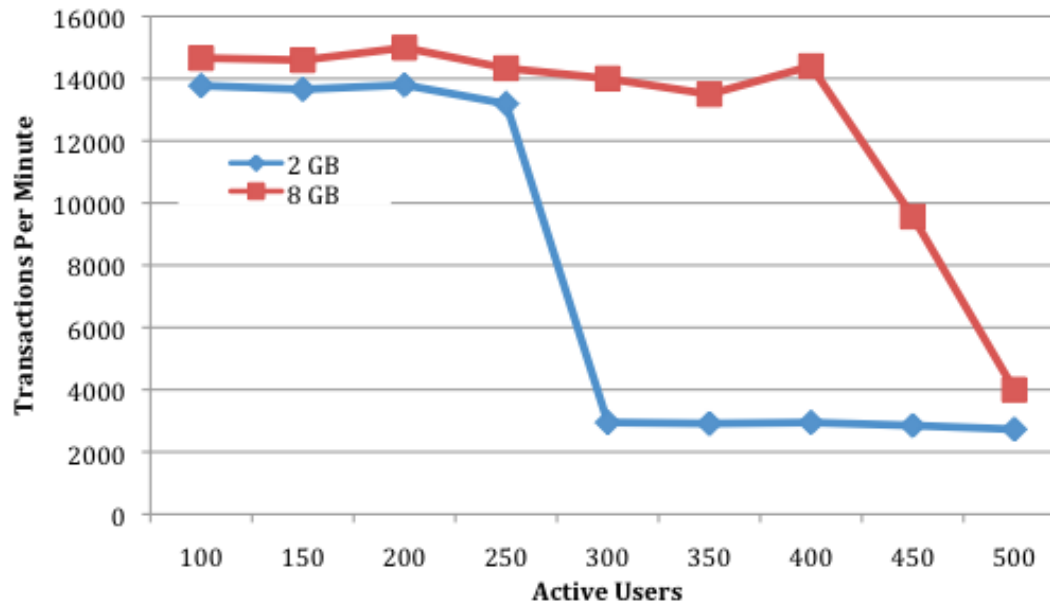
Main Memory Limitations

- DRAM inefficiencies increase cost and power
 - Processor cache increasing
 - Intel Nehalem processor
 - DRAM would need to have L3 BW and latency
 - “...create the illusion of a large memory that we can access as fast as a very small memory.” – Patterson & Hennessy

Local	L1	L2	L3	RAM
Read BW [GB/s]	45.6	31.1	26.2	10.1
Write BW [GB/s]	45.6	28.8	19.9	8.4
Latency [ns] (cycles)	1.3 (4)	3.4 (10)	13.0 (38)	65.1 (191)

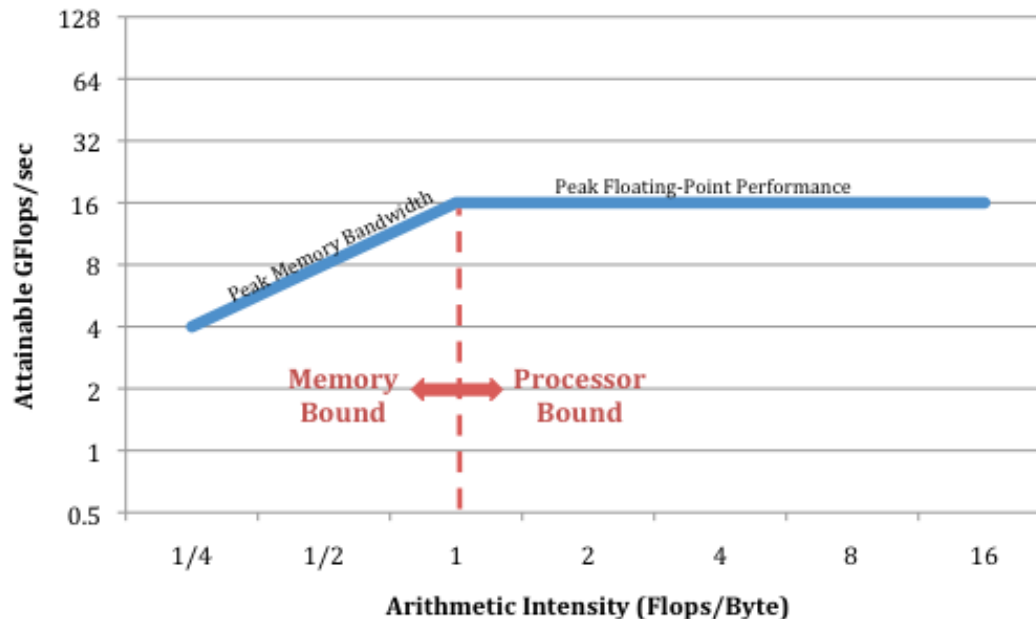
Main Memory Limitations

- DRAM efficiencies increase performance
- Capacity versus Performance
- Capacity costs power
 - Multiple memory channels
 - Each additional module increases power



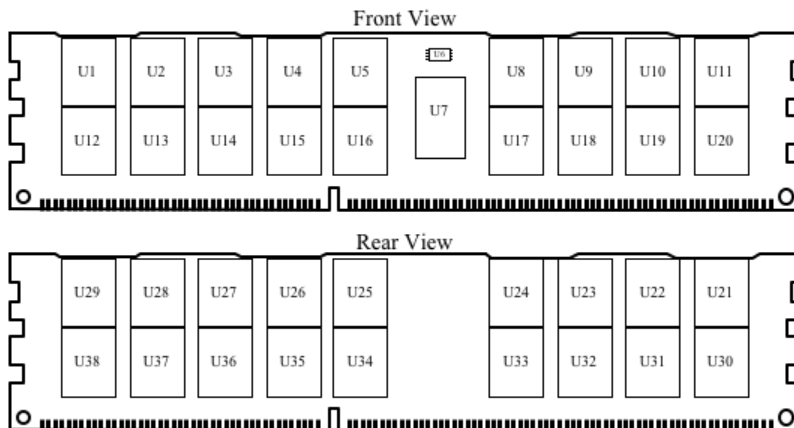
Main Memory Limitations

- Bandwidth versus performance
- Bandwidth costs power
 - Buffer on board
 - Multiple channels



Main Memory Limitations

- DRAM inefficiencies in practice
- Typical video/web server motherboard
 - 20+ layer PCB
 - 6 memory channels
- RDIMM
 - 10+ layer PCB
 - Maximum comp. count



Main Memory Limitations

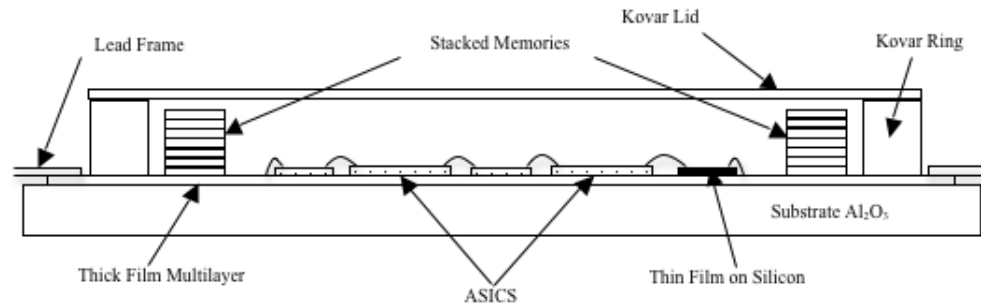
- 12 RDIMM
 - Termination
 - 36 components per DIMM
 - 8 I/O per component
 - 2.7 W of termination power for a read/write per module
 - 32.4 W total termination power
 - Wordline firing
 - 100 ns activation rate
 - 8126 page size
 - 200 fF per bitline
 - 11.2 W total bitline sense amplifier power
- Sustaining performance gains through capacity and bandwidth increases power and cost – innovation required.

Nano-Module

- Goals
 - Purpose was to move labs into prototype generation
 - Required low cost, high bandwidth, and low power memory solution that can be used with capacitive coupled interconnects in advanced server architectures
- Module component count trends required a new approach
- Nano-module proposed
 - Low cost advanced packaging technology
 - Off-the-shelf memory components
- Results can be leveraged
 - NAND
 - Mobile

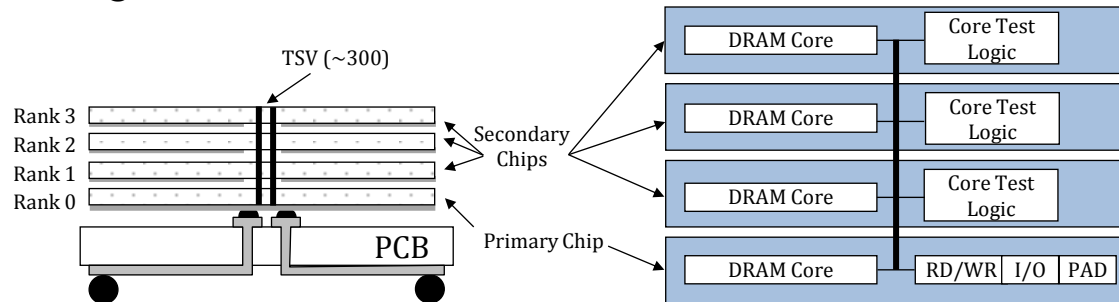
Nano-Module

- Literature review of high capacity memory stacks
- 1990's
 - Multichip Modules
 - Realized planar space limitations
 - Val & Lemione
 - Irvine Sensors
- Solutions proposed in research
 - No industry due to memory hierarchy effectiveness



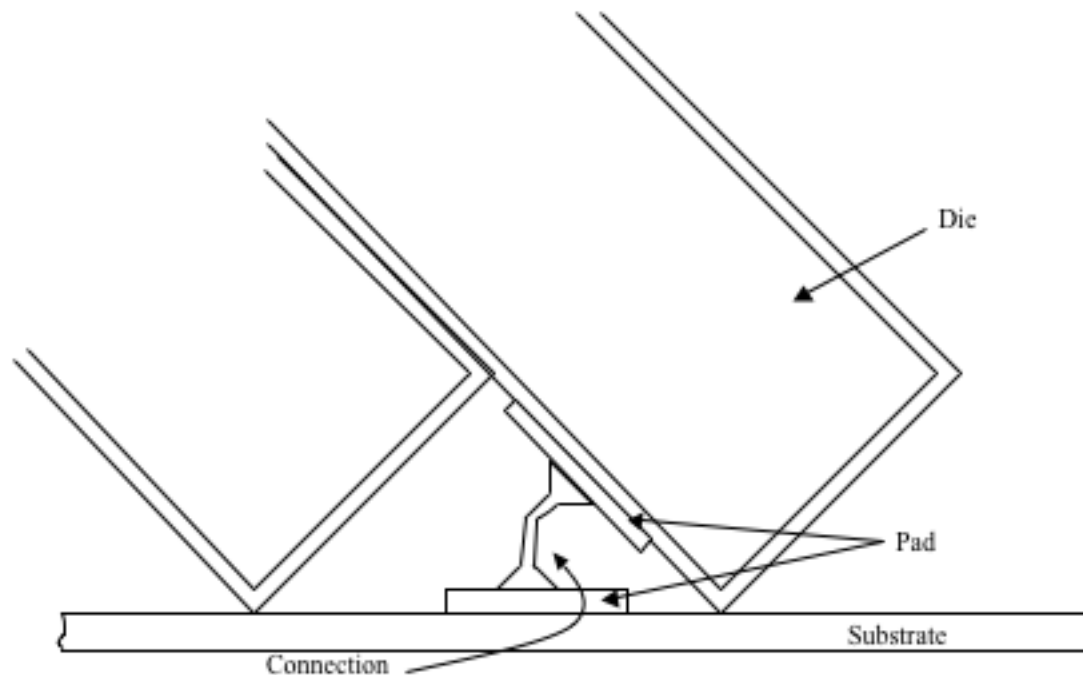
Nano-Module

- Memory stack technology gaining new attention
- Proposed in 2010 (more later about developments)
 - Samsung quad die with TSV
 - 80 μm pitch, 30 μm diameter, 300 TSV
 - $R_{TSV} = 5 \Omega$, $C_{TSV} = 300 \text{ fF}$
- Pros:
 - Lower power, higher bandwidth
- Cons:
 - Cost, integration



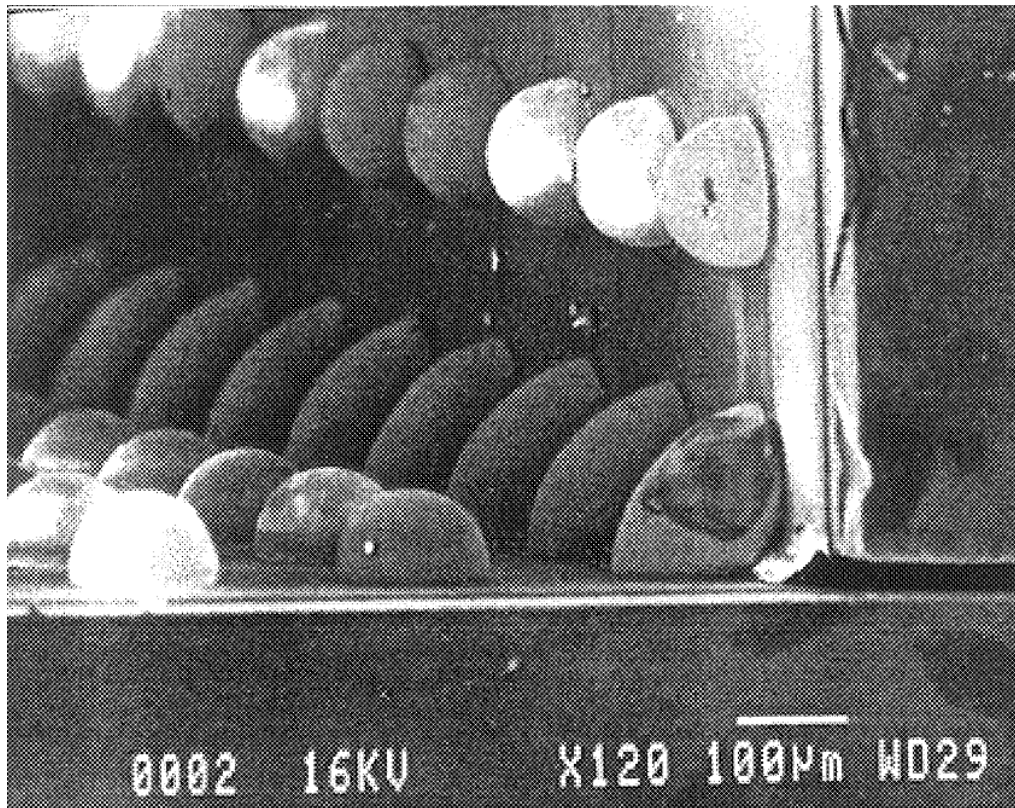
Nano-Module

- Literature review revealed novel solutions
- Slant the die!
- Applicable to capacitive-coupled interconnects



Nano-Module

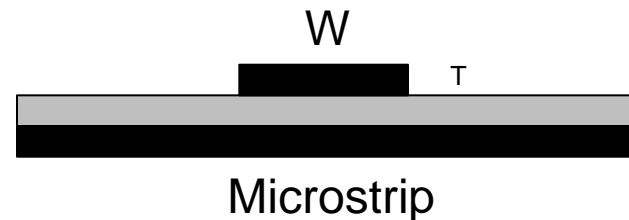
- Not the first to try it:



Nano-Module

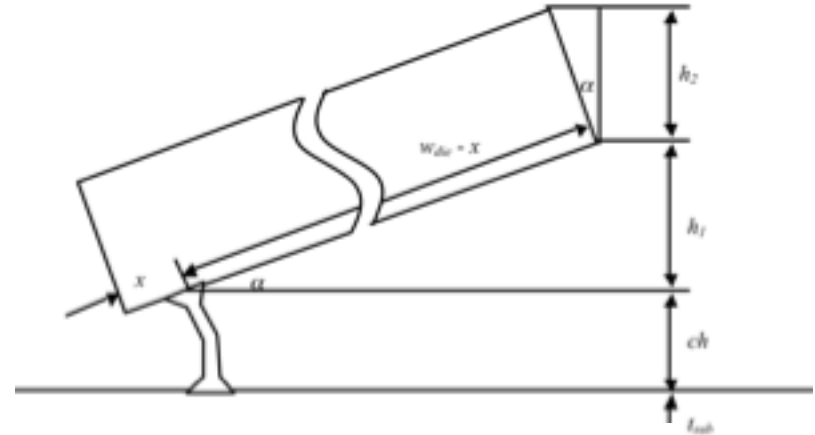
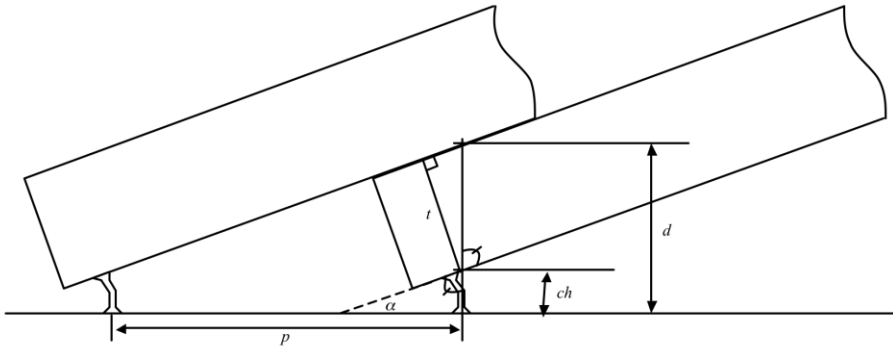
- Controlled Impedance
 - All Signals 50 Ω controlled impedance
 - DQS and CLK 120 Ω differential impedance
- Trace Length Matching
 - All Data matched to worst case
 - All CLK matched to worst case
 - All Address/Command matched to worst case

$$Z_0 = \frac{87}{\sqrt{\epsilon_r + 1.41}} \ln \left(\frac{5.98H}{0.8W + T} \right)$$



Nano-Module

- Size calculations



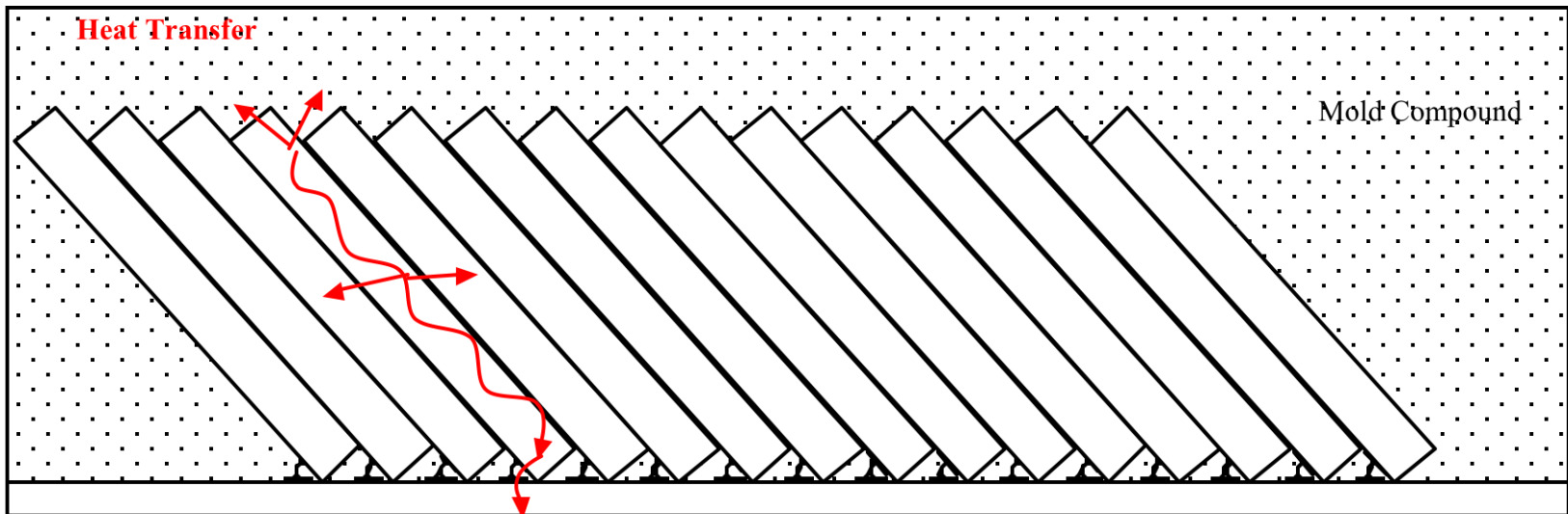
$$height = t_{sub\ thickness} + t_{connection} + \frac{t_{die\ thickness}}{\cos \alpha} + (t_{die\ width} - t_{die\ pad\ to\ edge}) \sin \alpha$$

$$width = (no.\ signals - 1)t_{con\ pitch} + t_{con\ diameter} + 2(t_{die\ to\ pad\ edge} + t_{wb_1} + t_{wb_2})$$

$$length = 2(t_{wb_1} + t_{wb_2}) + \sin \alpha \cdot t_{die\ thickness} + \frac{(\# die - 1)t_{die\ thickness}}{\sin \alpha} + \cos \alpha \cdot t_{die\ width}$$

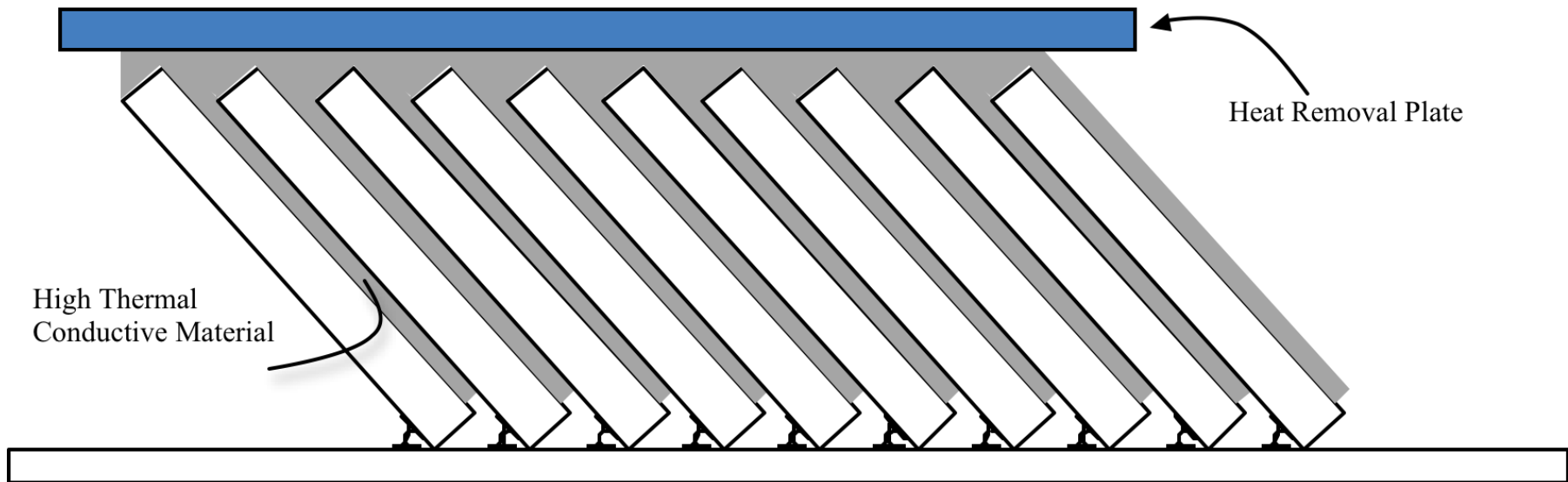
Nano-Module

- Thermal option
 - Thermal conductivity
 - Silicon, Metals \gg Mold Compound
 - Hot spots
 - Temperature gradient



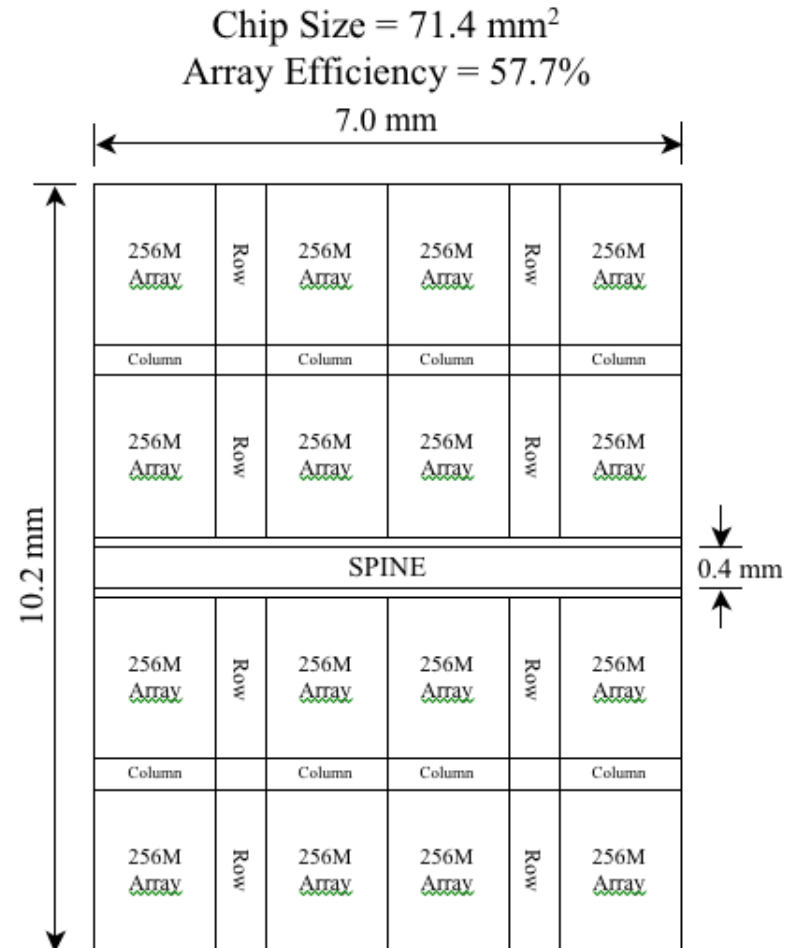
Nano-Module

- Thermal option
 - Heat plate



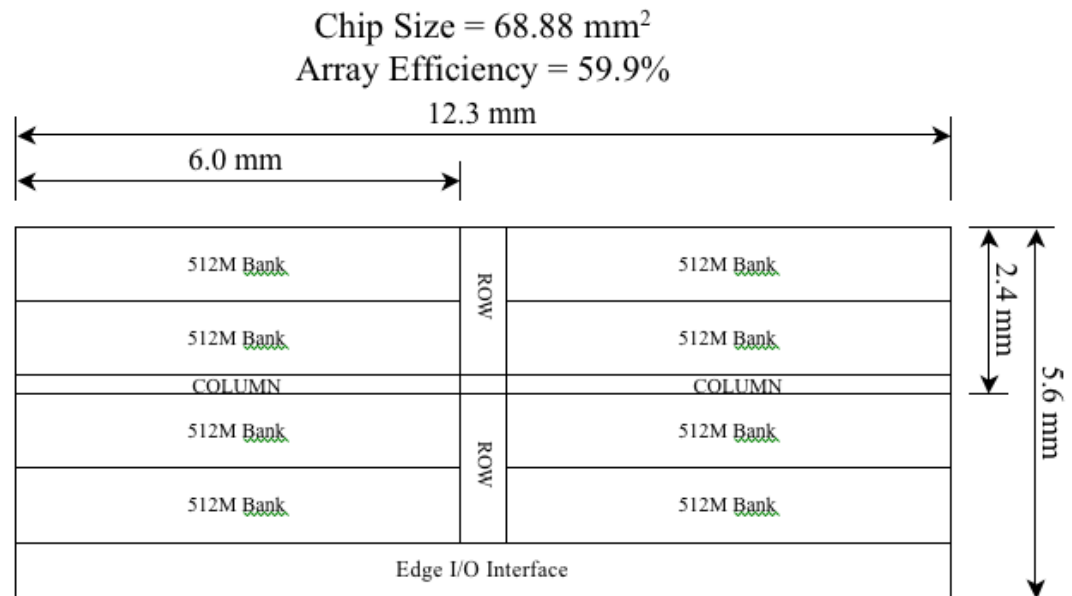
Wide I/O DRAM Architecture

- 4 Gb DRAM
 - Meets ITRS predictions
- Edge aligned pads
- Page size reduction
- Low cost process
 - < 4 levels of metal
 - No impact to die size
 - No impact to array efficiency
- Move to 64 data pins
 - Report challenges
 - Propose innovations



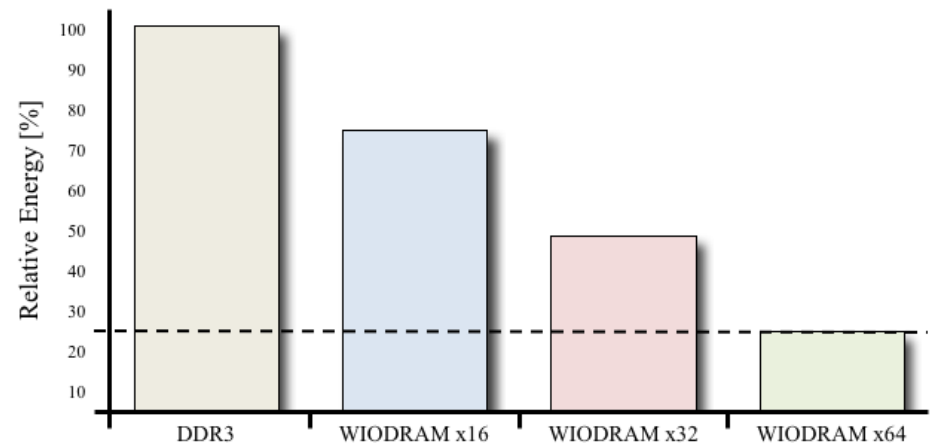
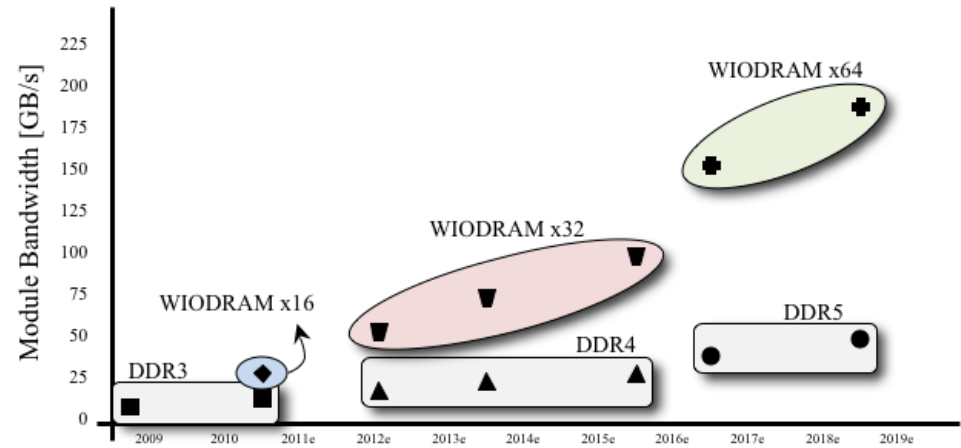
Wide I/O DRAM Architecture

- 4 Gb Edge DRAM
 - Centralized Row and Column
 - Smaller die
 - Higher efficiency
 - < 4 levels of metal



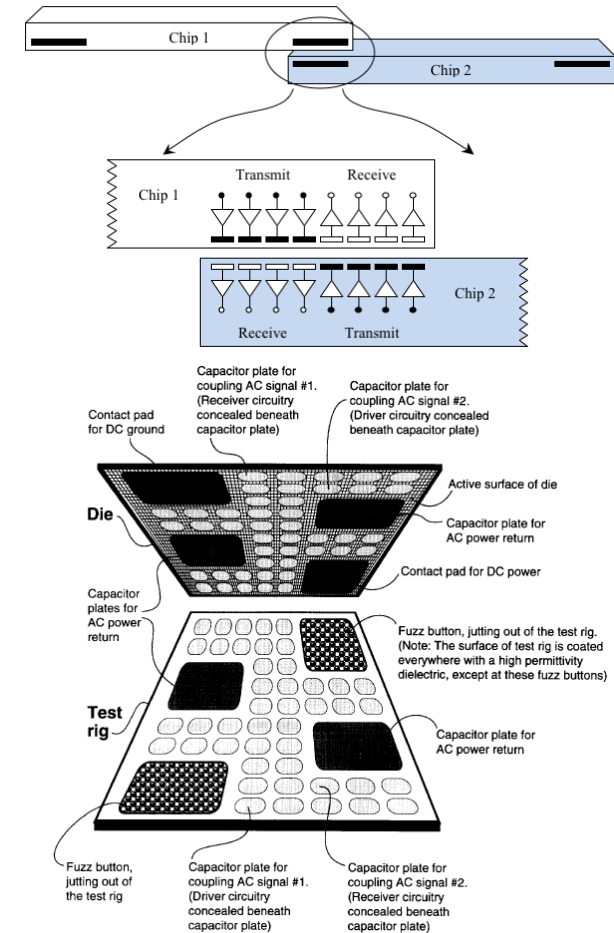
Wide I/O DRAM Architecture

- Challenges
 - Number of metal layers
 - Global data routing
 - Local data routing
- Proposals
 - Split bank structure
 - Data-path design
 - Through bitline routing
 - SLICE architecture
 - Capacitive-coupled I/O



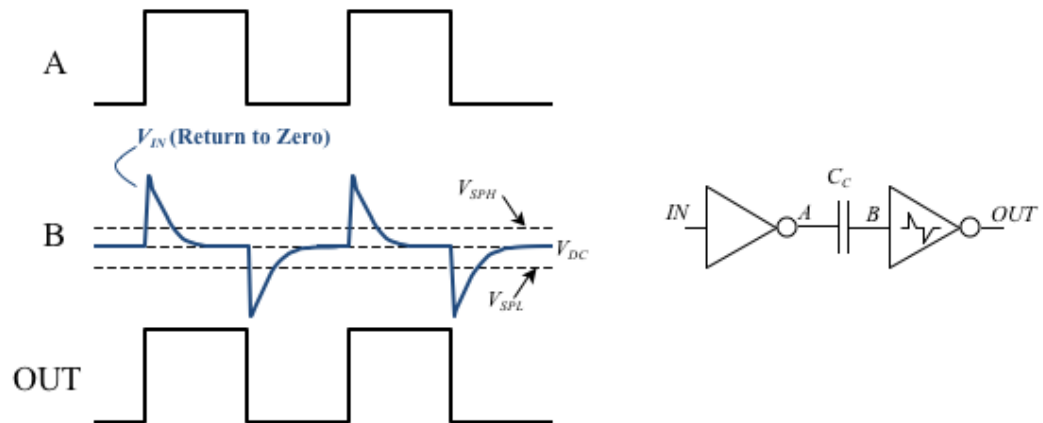
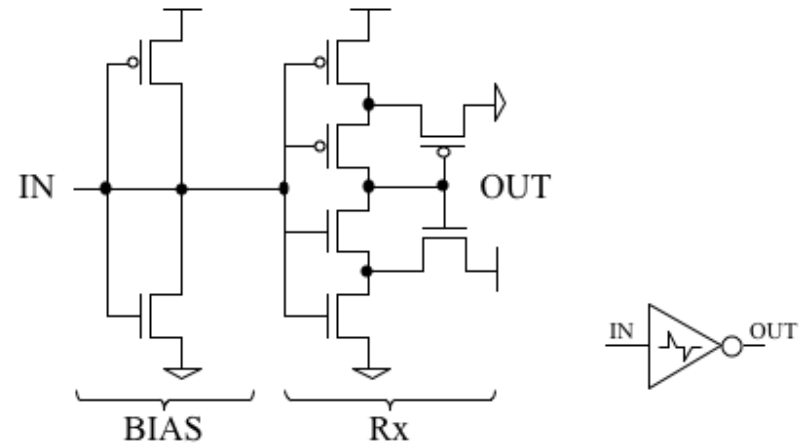
High Bandwidth Interconnect

- Capacitive-coupling
 - Increased bandwidth
 - Reduced ESD capacitance
 - Smaller I/O channel = more I/O
 - Removal of inductive channel
 - Low power
 - Reduced ESD capacitance
 - Low power Tx & Rx
 - Low cost
 - Simple
 - Alignment required
- Literature review
 - Revealed inefficiencies and lack of application



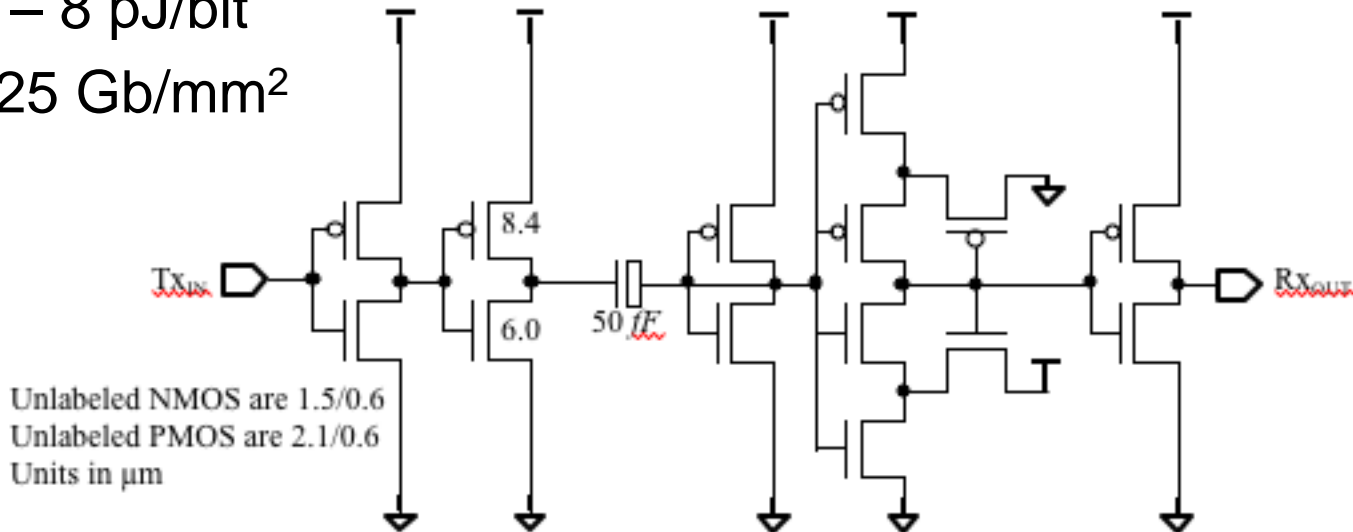
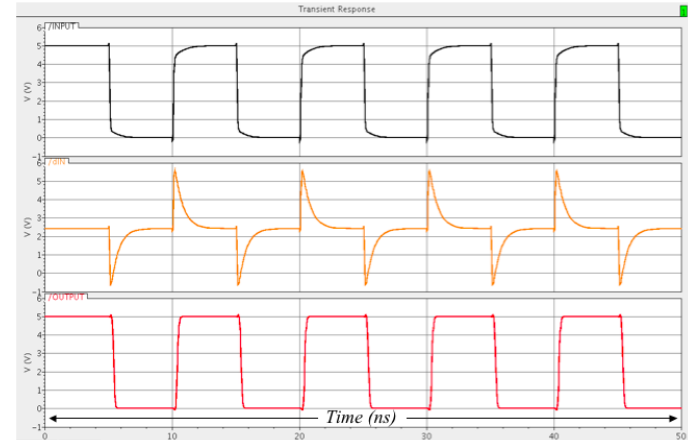
High Bandwidth Interconnect

- Proposed receiver design
 - Extreme low power
 - ~1 gate delay latency
 - 'DC' transmission
 - RTZ → NRZ



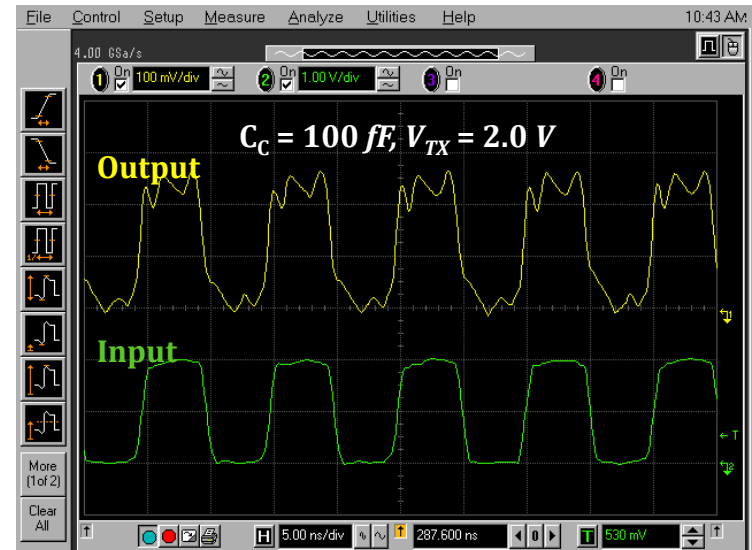
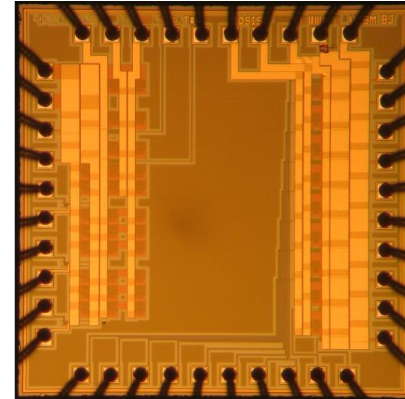
High Bandwidth Interconnect

- ❑ 0.5 μm CMOS design (proof of concept)
 - ✓ 5.0 V process
 - ✓ 50 fF poly-poly capacitor
 - ✓ 200 Mbps
 - ✓ 3 – 8 pJ/bit
 - ✓ 325 Gb/mm²



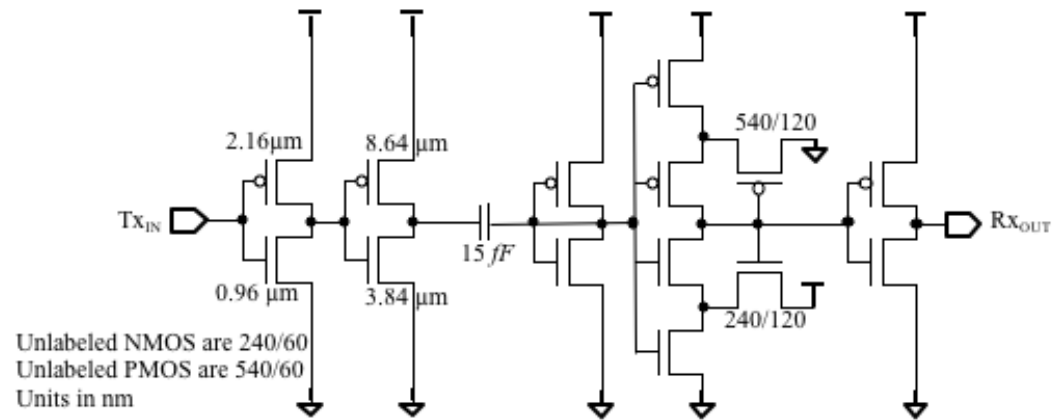
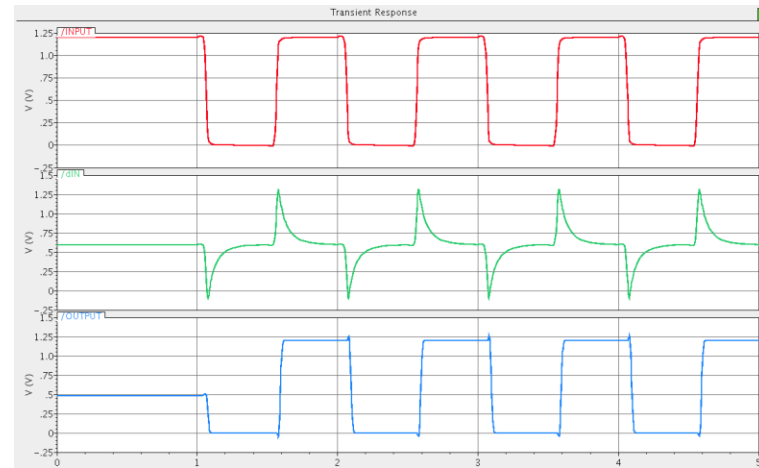
High Bandwidth Interconnect

- Chip micrograph
 - 1.5 mm x 1.5 mm
 - 9 structures
- Experimental results
 - Operate at $V_{TX} = 2.0$ V
 - 3 pJ/bit at 200 Mbps



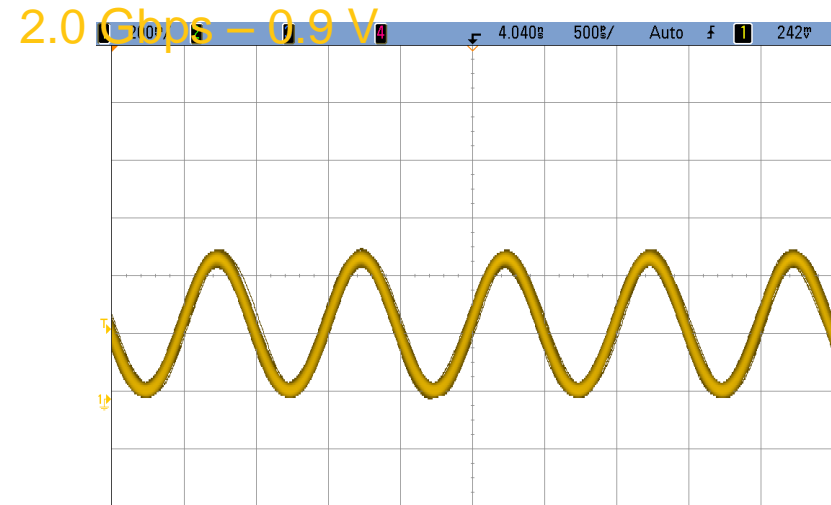
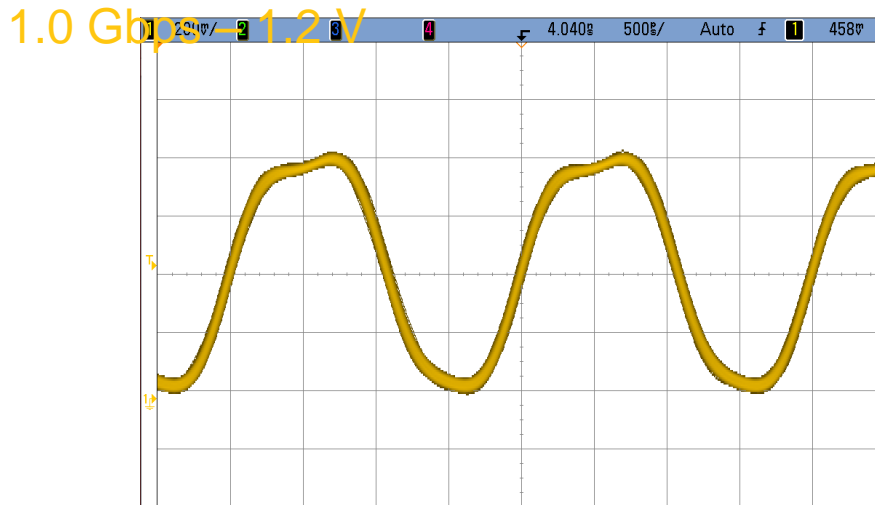
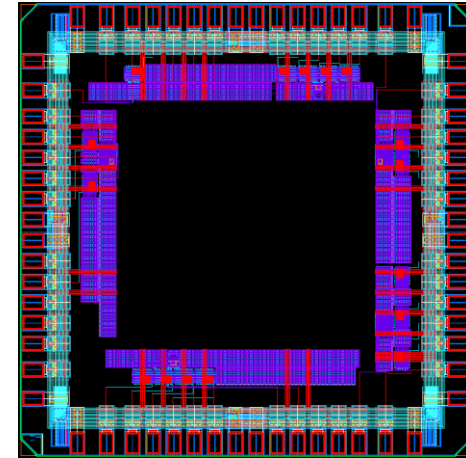
High Bandwidth Interconnect

- 65 nm CMOS design (proof of scalability)
 - 1.2 V process
 - 15 fF metal-metal capacitor
 - 4 Gbps
 - $17 \mu\text{m}^2$
 - 227 Tbps/mm²



High Bandwidth Interconnect

- Die micrograph
 - 2 mm x 2 mm
- Experimental results
 - 2 Gbps @ 0.9V
 - 50 fF coupling capacitor



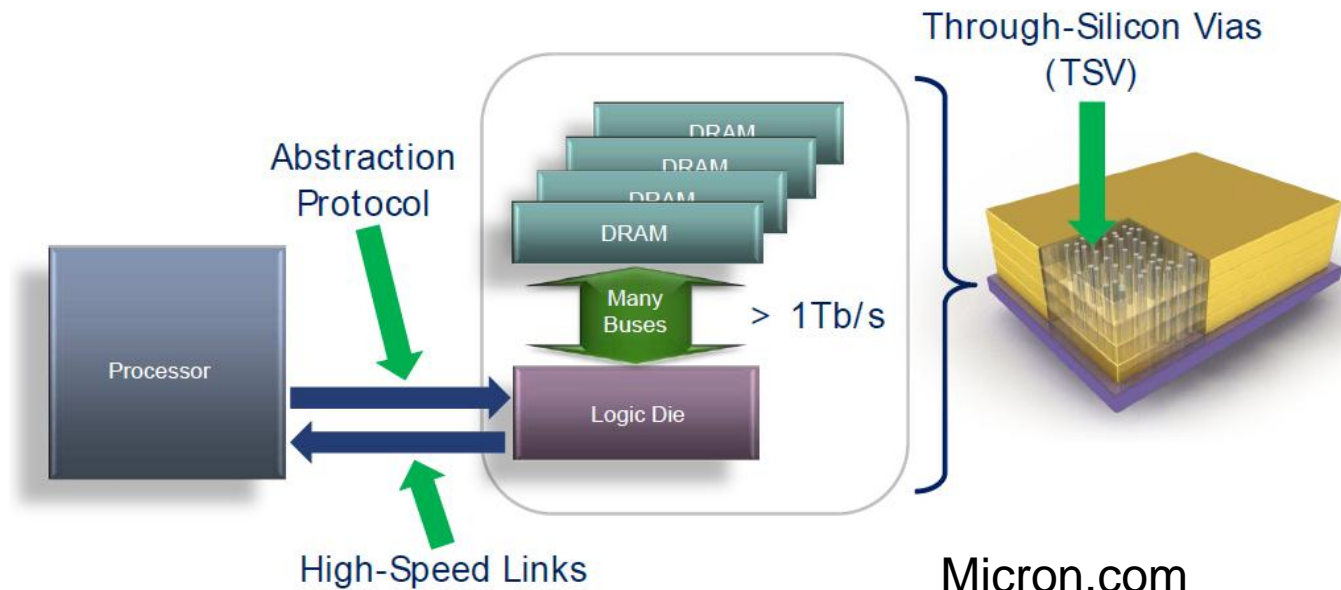
High Bandwidth Interconnect

Work	Process	Supply (V)	Data Rate (Gbps)	Coupling (fF)	Bandwidth (Gbps/mm ²)	Energy (pJ/bit)	Requires CLK
Kanda 2003 [31]	0.35 μm	3.3	1.27	10	2.117×10^3	2.4	Yes
Franzon 2006 [27]	0.18 μm	1.8	3	150	5.55×10^2	5.0	No
Fazzi 2007 [32]	0.13 μm	1.2	1.23	10	1.922×10^4	0.14	Yes
Kim 2009 [33]	0.18 μm	1.8	2	600	6.90×10^2	0.8	Yes
This Work	0.5 μm	5.0	0.2	50	3.25×10^2	8	No
This Work	65 nm	1.2	4	15	2.268×10^5	0.015	No

So What is the Industry Moving Towards?

- Hybrid Memory Cube

Hybrid Memory Cube (HMC)

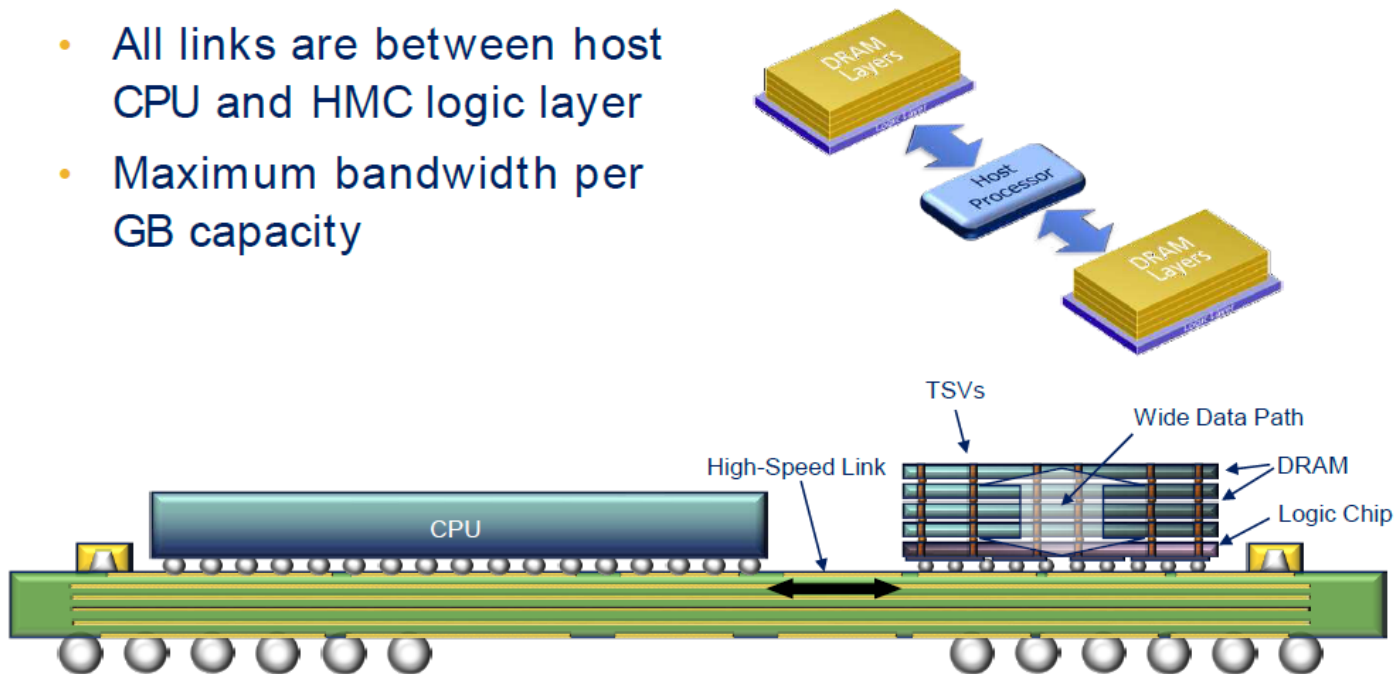


Notes: Tb/s = Terabits / second
HMC height is exaggerated

Hybrid Memory Cube

HMC Near Memory – MCM Configuration

- All links are between host CPU and HMC logic layer
- Maximum bandwidth per GB capacity

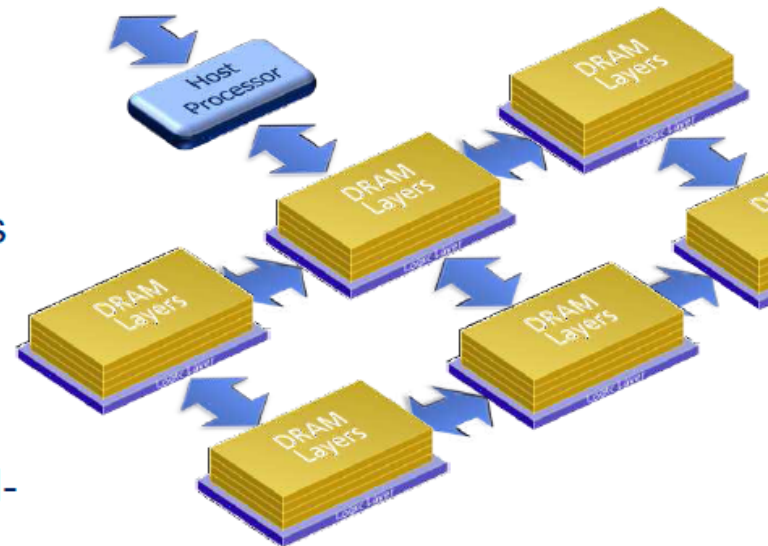


Notes: MCM = multi-chip module
 Illustrative purposes only; height is exaggerated

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HMC “Far” Memory

- Far memory
 - ▶ Some HMC links connect to host, some to other cubes
 - ▶ Serial links form networks of cubes
 - ▶ the memory = the network
 - ▶ Scalable to meet system requirements
 - ▶ Can be in module form or soldered-down
 - ▶ Can form a variety of topologies e.g., tree, ring, double-ring, mesh
- Future interfaces
 - ▶ Higher speed electrical (SERDES)
 - ▶ Optical
 - ▶ Whatever the most appropriate interface for the job



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HMC_{Gen1}: Technology Comparison

Generation 1 (4 + 1 memory configuration)

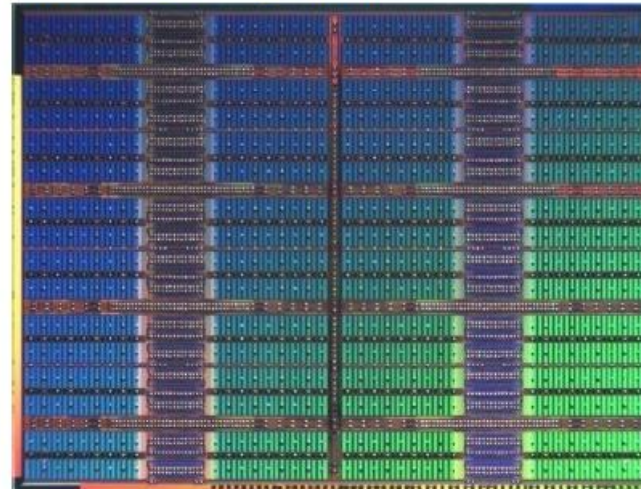
Technology	VDD	IDD	BW GB/ s	Power (W)	mW/ GB/ s	pj/ bit	real pJ/ bit
SDRAM PC133 1GB Module	3.3	1.50	1.06	4.96	4664.97	583.12	762
DDR-333 1GB Module	2.5	2.19	2.66	5.48	2057.06	257.13	245
DDRII-667 2GB Module	1.8	2.88	5.34	5.18	971.51	121.44	139
DDR3-1333 2GB Module	1.5	3.68	10.66	5.52	517.63	64.70	52
DDR4-2667 4GB Module	1.2	5.50	21.34	6.60	309.34	38.67	39
HMC, 4 DRAM w/ Logic	1.2	9.23	128.00	11.08	86.53	10.82	13.7

Simple calculation from IDD7 (SDRAM IDD4)

Real system,
some with lower
density modules

- 1Gb 50nm DRAM Array
- 90nm prototype logic
- 512MB total DRAM cube
- 128GB/s Bandwidth
- 27mm x 27mm prototype
- Functional demonstrations!
- Reduced host CPU energy

HMC Gen 1 DRAM



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Conclusions

- Nano-Module
 - Developed a new research direction for industry research labs
 - Developed initial motivation
 - Developed initial prototype
- DRAM Architecture
 - Demonstrated benefits of wide I/O topologies
 - Proposed several low power innovations
 - Provided application for novel interconnect technologies
- Capacitive-Coupled Receiver
 - Demonstrated low power receiver designs
 - Achieved 2 Gbps at < 15 fJ/bit in 65 nm
- Summarized industry direction – Hybrid Memory Cube

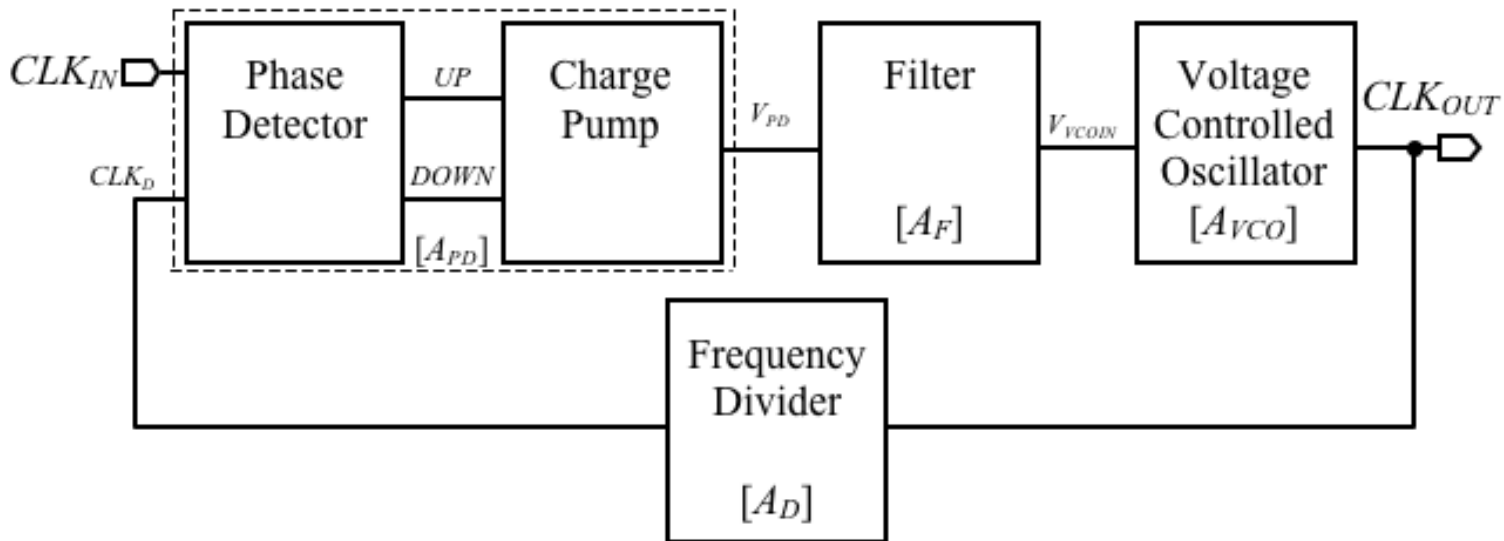
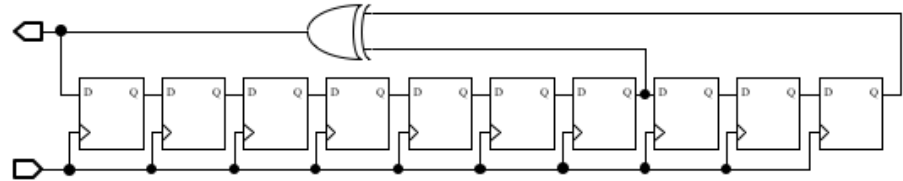


Questions

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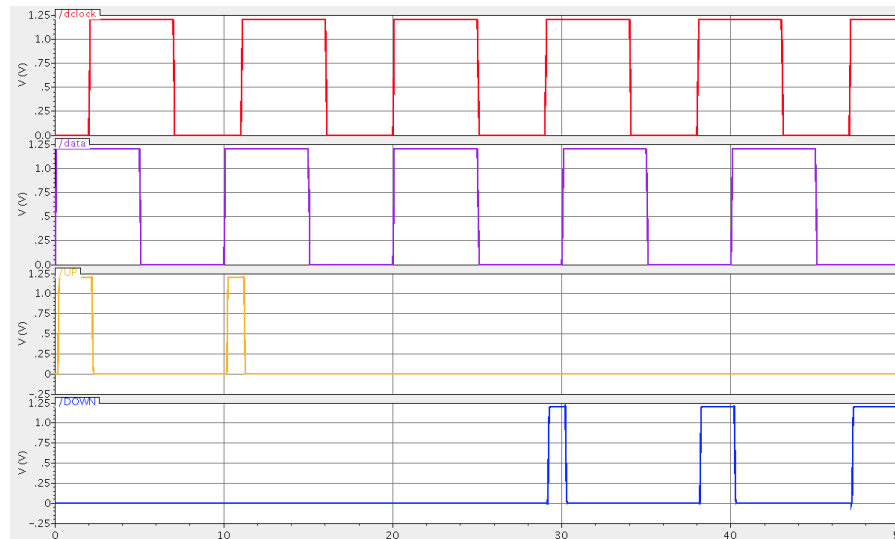
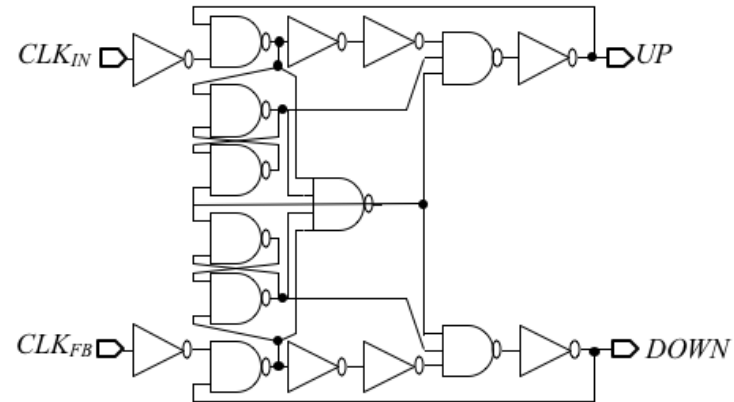
Appendix - PLL

- 65 nm test chip
 - PLL
 - PRBS generator



Appendix - PLL

- PLL
 - Phase detector

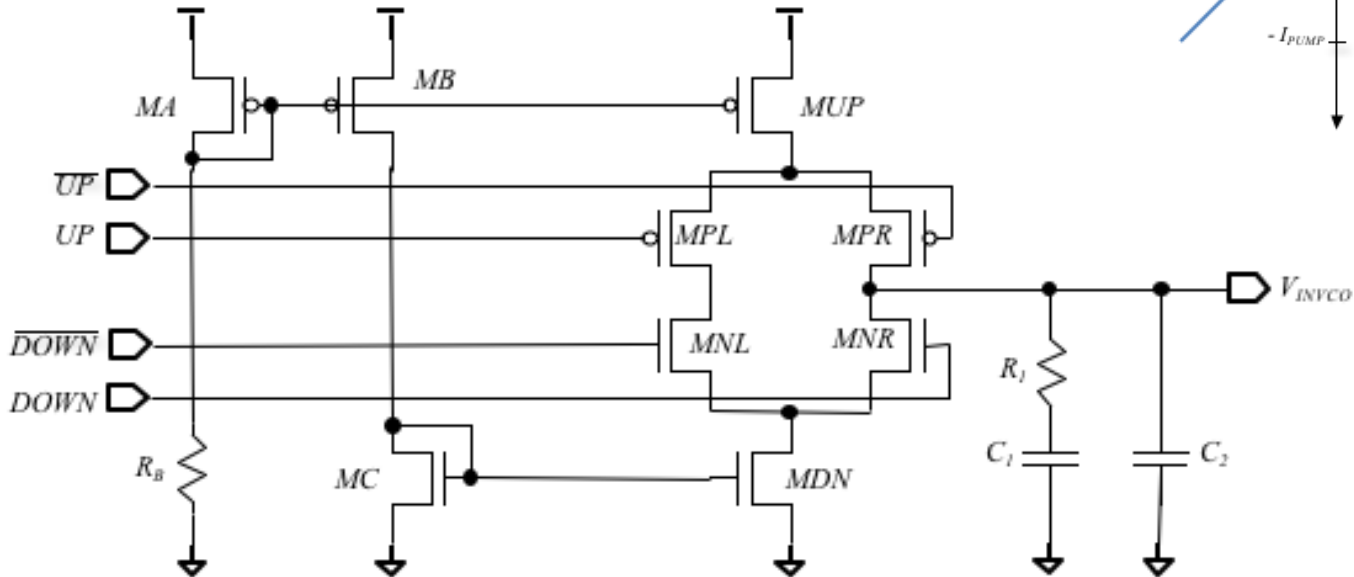
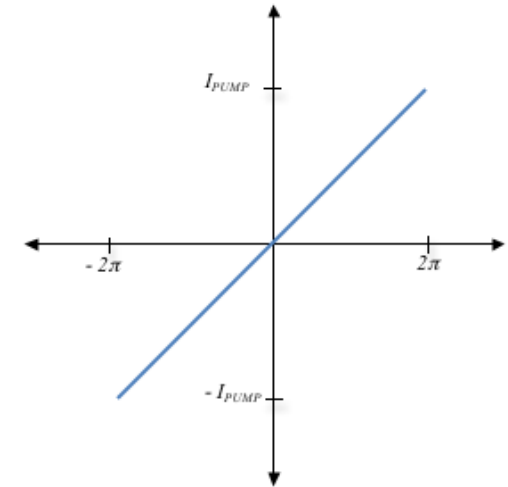


Appendix - PLL

- Charge pump

$$A_{PD} = \frac{I_{PUMP}}{2\pi}$$

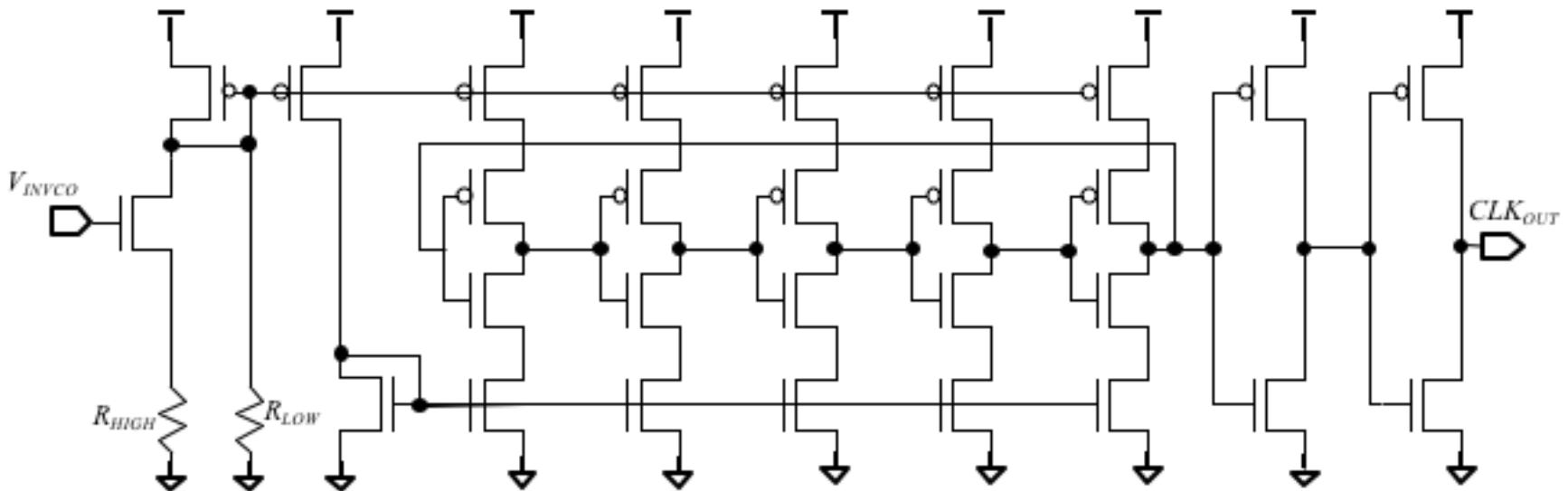
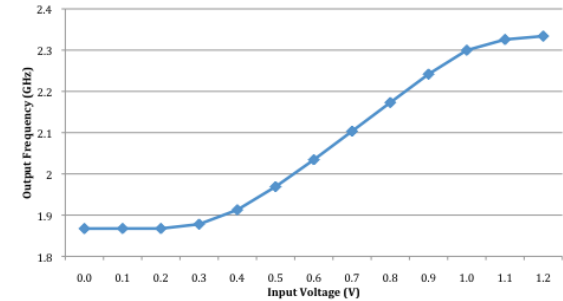
$$A_F = \frac{1 + sR_1C_1}{sC_1}$$



Appendix - PLL

- Voltage controlled oscillator

$$A_{VCO} = 2\pi \cdot \frac{f_{MAX} - f_{MIN}}{V_{MAX} - V_{MIN}}$$



Appendix - PLL

$$\phi_{CLK_{OUT}} = V_{INVCO} \cdot \frac{A_{VCO}}{s}$$

$$\phi_{CLK_D} = \frac{1}{N} \cdot \phi_{CLK_{OUT}} = \beta \cdot \phi_{CLK_{OUT}}$$

$$A_{OL} = A_{PD} A_F A_{VCO}$$

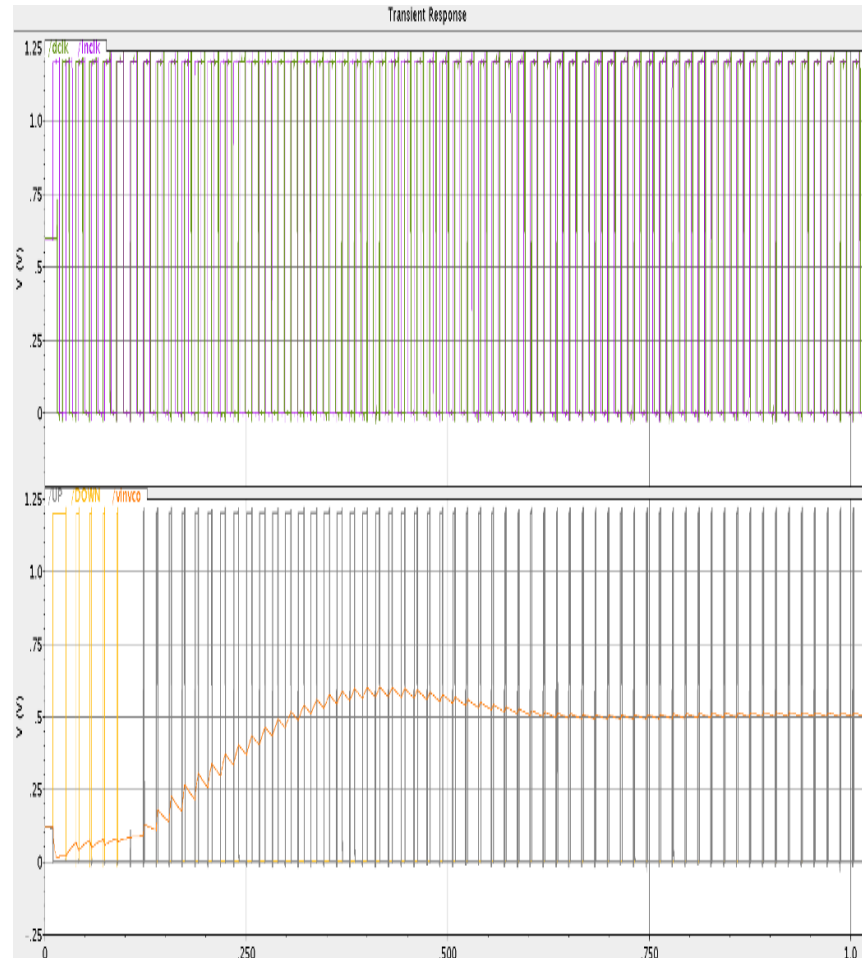
$$H(s) = \frac{\phi_{CLK_{IN}}}{\phi_{CLK_{OUT}}} = \frac{A_{PD} A_F \frac{A_{VCO}}{s}}{1 + \beta A_{PD} A_F \frac{A_{VCO}}{s}} = \frac{A_{PD} A_F A_{VCO}}{s + \beta A_{PD} A_F A_{VCO}}$$

$$A_F = \frac{1 + sR_1C_1}{sC_1}$$

$$H(s) = \frac{\phi_{CLK_{IN}}}{\phi_{CLK_{OUT}}} = \frac{A_{PD} A_{VCO} \left(\frac{1 + sR_1C_1}{C_1} \right)}{s^2 + s \left(\frac{A_{PD} A_{VCO} R_1}{N} \right) + \left(\frac{A_{PD} A_{VCO}}{NC_1} \right)}$$

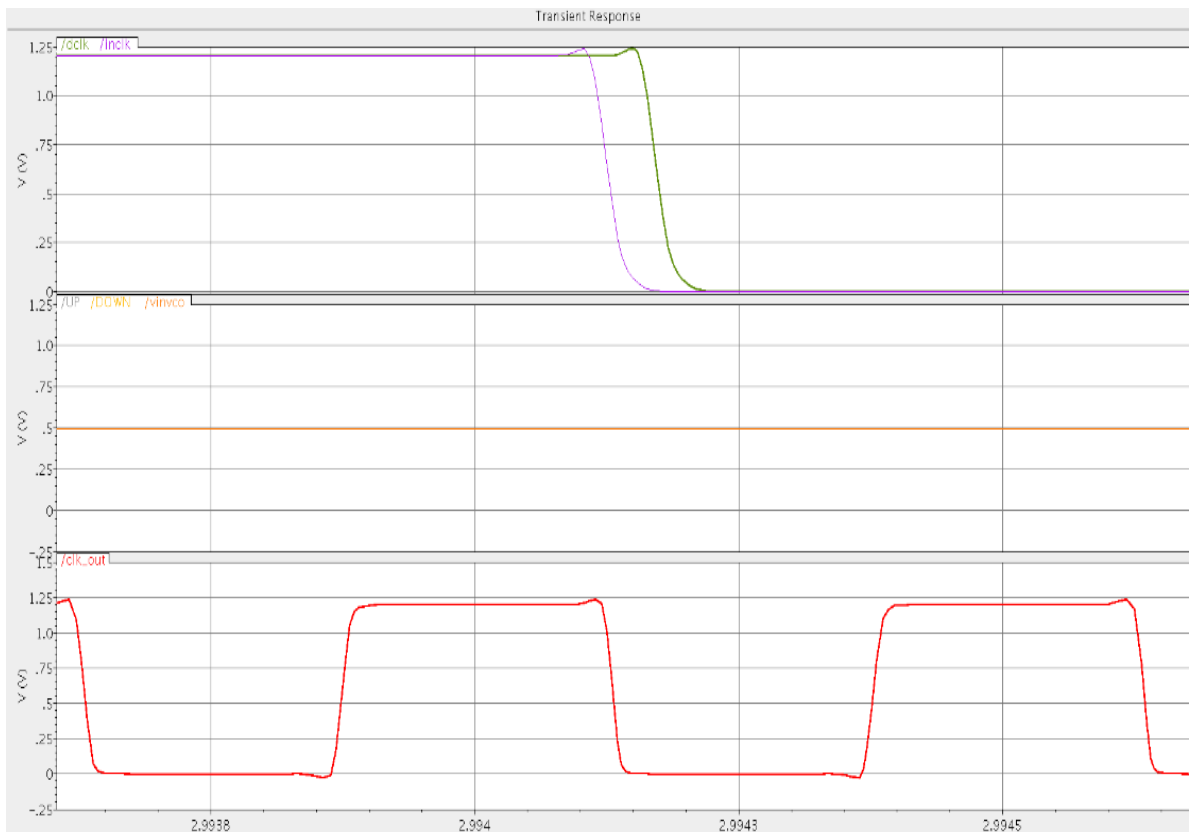
$$\omega_n = \sqrt{\frac{A_{PD} A_{VCO}}{NC_1}}$$

$$\zeta = \frac{\omega_n}{2} R_1 C_1$$



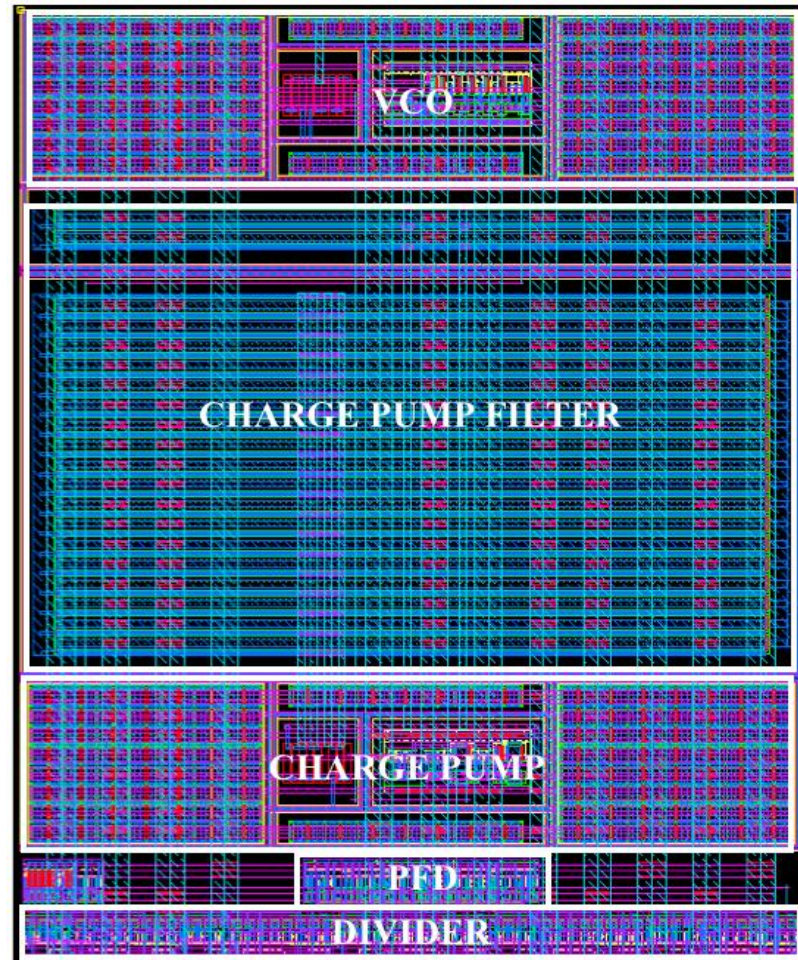
Appendix - PLL

- PLL at lock



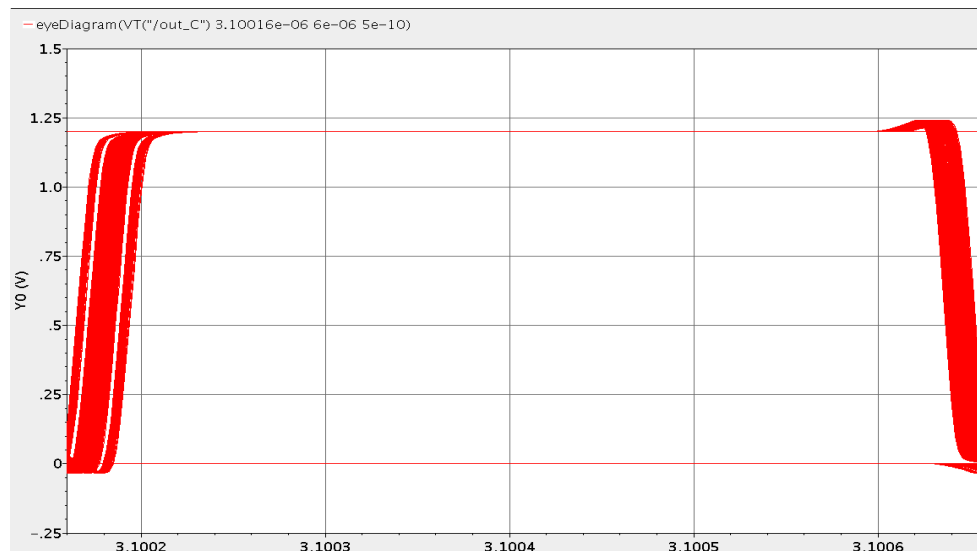
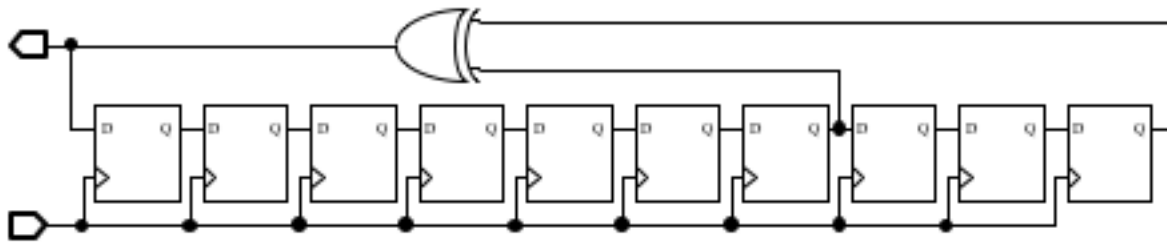
Appendix - PLL

- PLL layout



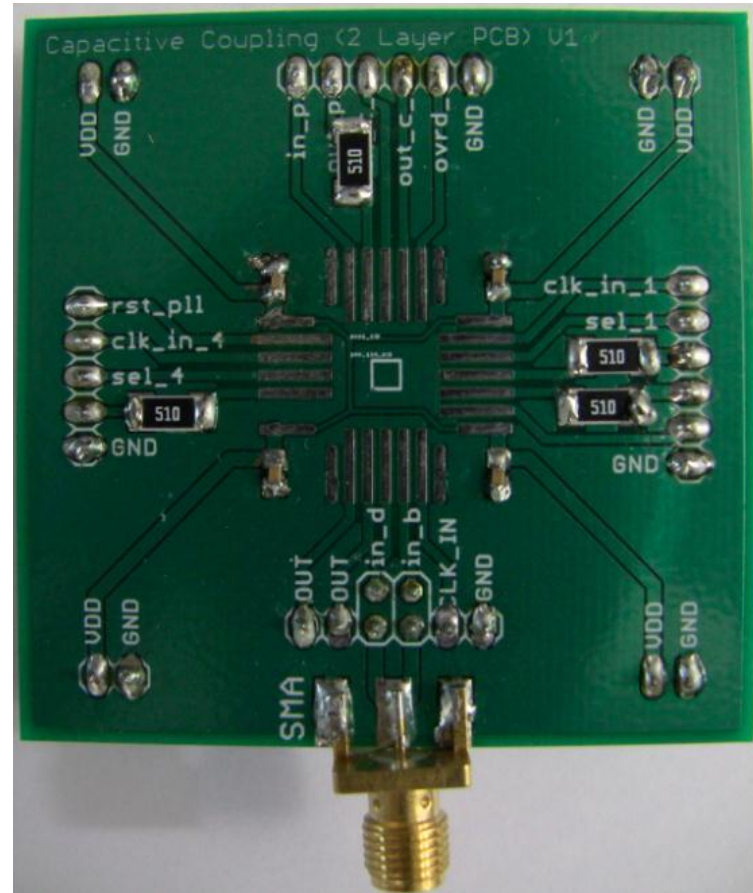
Appendix - PLL

- PRBS generator

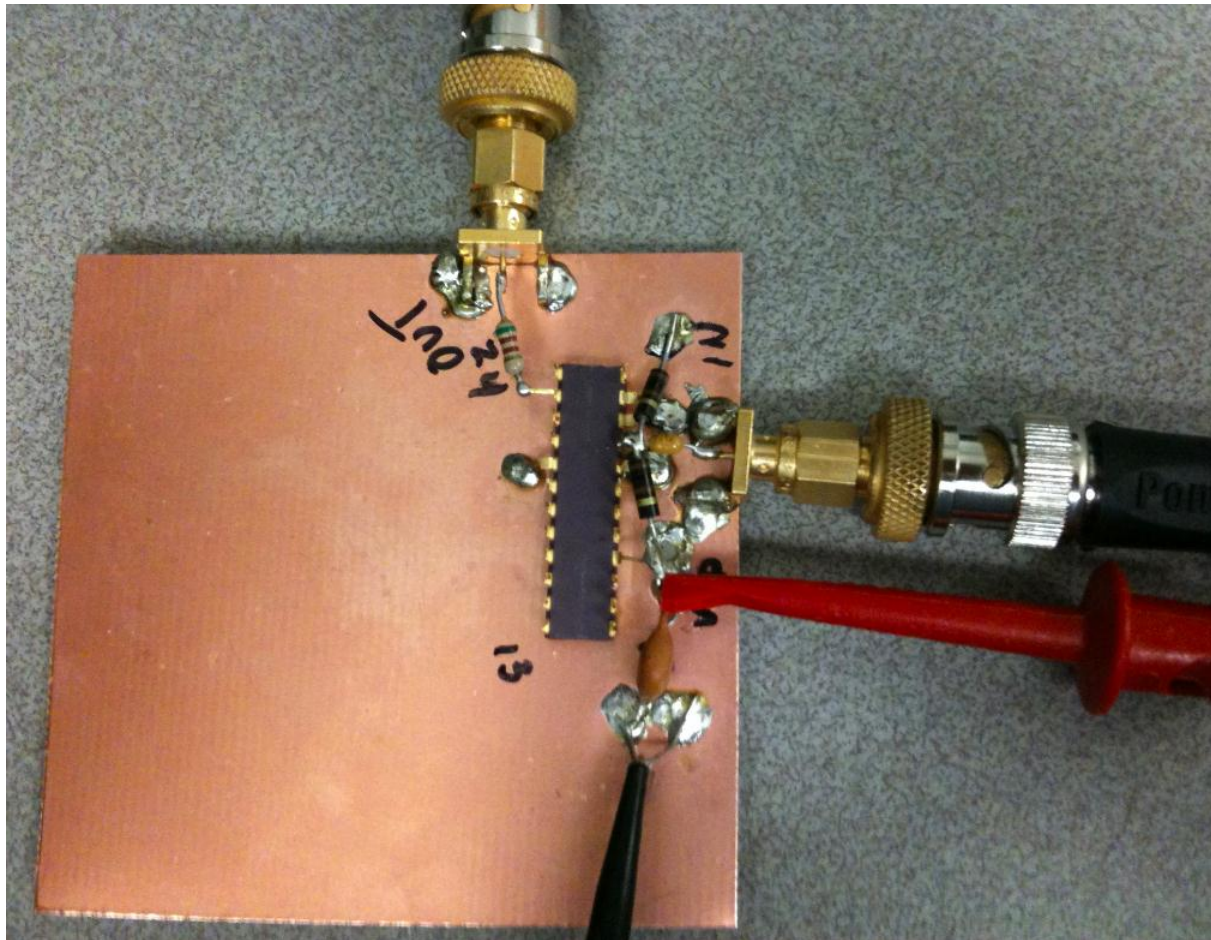


Appendix - PCB

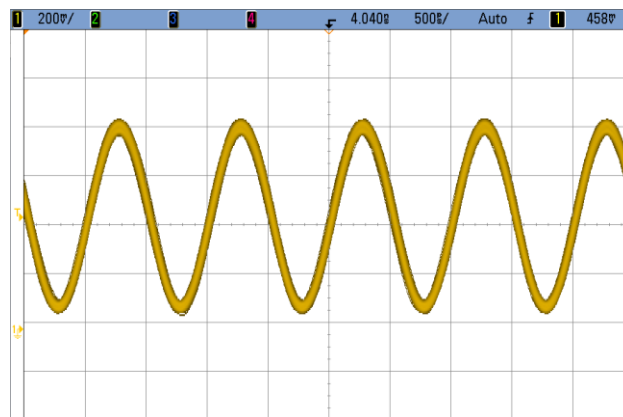
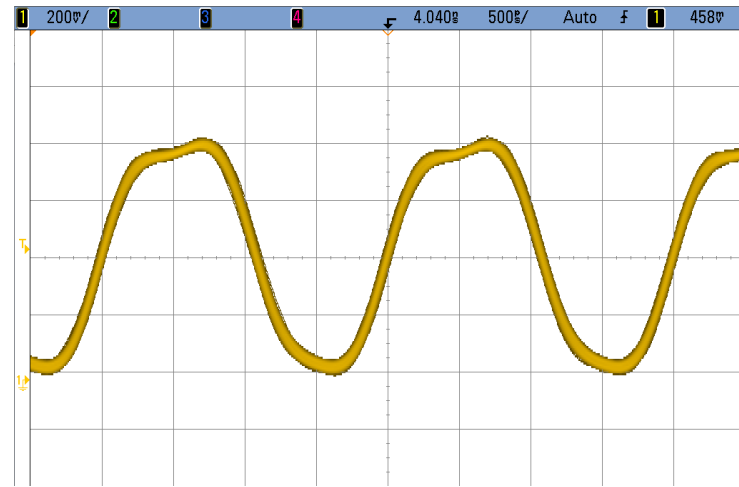
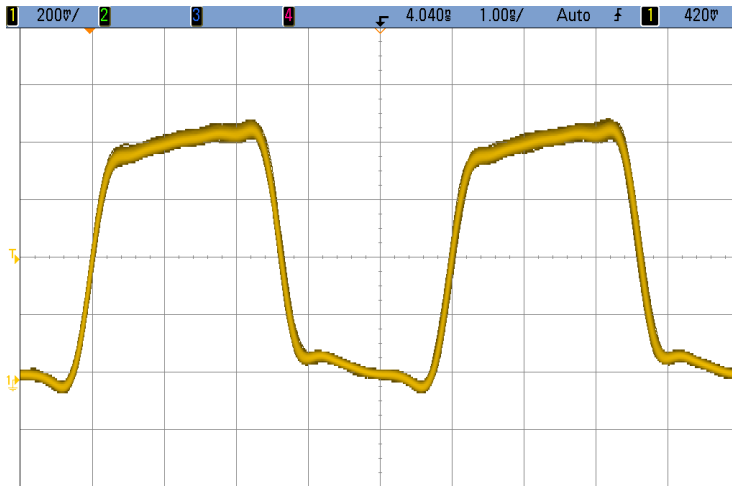
- PCB test board



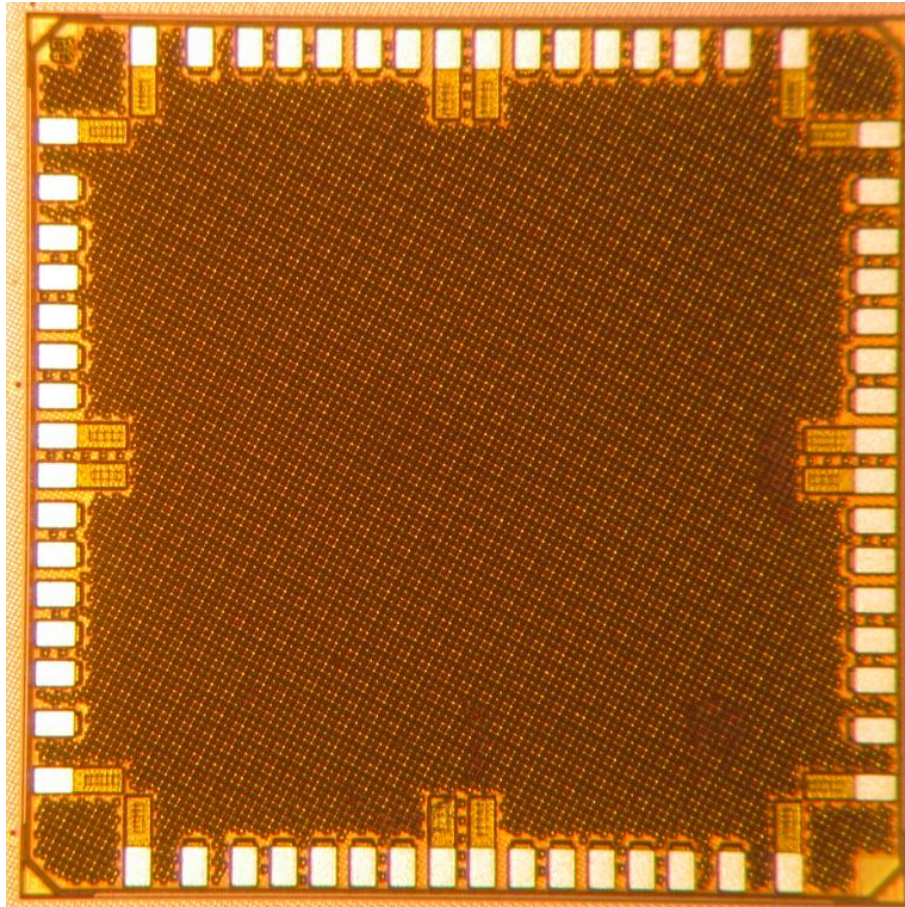
Appendix – Dead Bug



Appendix – Dead Bug



Appendix – 65 nm Chip



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